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(54) **METHOD AND SYSTEM FOR DRIVING LIGHT EMITTING DISPLAY**

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(56) **References Cited**

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U.S. PATENT DOCUMENTS

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3,506,851 A 4/1970 Polkinghorn et al.
3,750,987 A 8/1973 Gobel
3,774,055 A 11/1973 Bapat et al.
(Continued)

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FOREIGN PATENT DOCUMENTS

AU 729652 6/1997
AU 764896 12/2001
(Continued)

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OTHER PUBLICATIONS

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(Continued)

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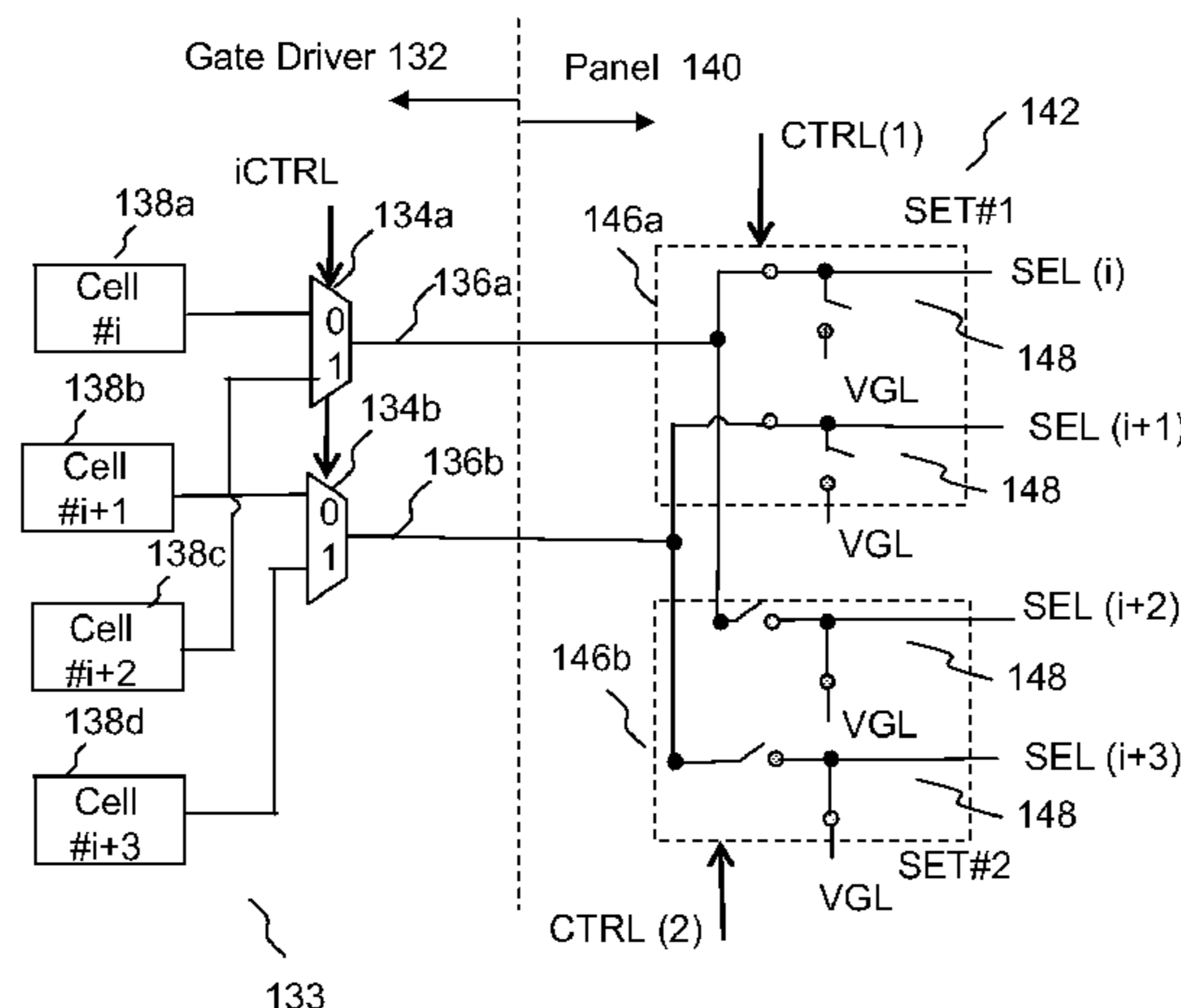
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(57) **ABSTRACT**

A display system includes a driver for operating a panel having a plurality of pixels arranged by a plurality of first lines and at least one second line. The driver includes a driver output unit for providing to the panel a single driver output for activating the plurality of first lines, the single driver output being demultiplexed on the panel to activate each first line.

4 Claims, 19 Drawing Sheets



(56)

References Cited

U.S. PATENT DOCUMENTS

4,090,096 A	5/1978	Nagami	6,753,655 B2	6/2004	Chen et al.
4,354,162 A	10/1982	Wright	6,753,834 B2	6/2004	Mikami et al.
4,996,523 A	2/1991	Bell et al.	6,756,741 B2	6/2004	Li
5,134,387 A *	7/1992	Smith et al. 345/83	6,777,888 B2	8/2004	Kondo
5,153,420 A	10/1992	Hack et al.	6,781,567 B2	8/2004	Kimura
5,170,158 A	12/1992	Shinya	6,788,231 B1	9/2004	Hsueh
5,204,661 A	4/1993	Hack et al.	6,809,706 B2	10/2004	Shimoda
5,266,515 A	11/1993	Robb et al.	6,828,950 B2	12/2004	Koyama
5,278,542 A *	1/1994	Smith et al. 345/690	6,858,991 B2	2/2005	Miyazawa
5,408,267 A	4/1995	Main	6,859,193 B1	2/2005	Yumoto
5,498,880 A	3/1996	Lee et al.	6,876,346 B2	4/2005	Anzai et al.
5,572,444 A	11/1996	Lentz et al.	6,900,485 B2	5/2005	Lee
5,589,847 A	12/1996	Lewis	6,903,734 B2	6/2005	Eu
5,619,033 A	4/1997	Weisfield	6,911,960 B1	6/2005	Yokoyama
5,648,276 A	7/1997	Hara et al.	6,911,964 B2	6/2005	Lee et al.
5,670,973 A	9/1997	Bassetti et al.	6,914,448 B2	7/2005	Jinno
5,691,783 A	11/1997	Numao et al.	6,919,871 B2	7/2005	Kwon
5,701,505 A	12/1997	Yamashita et al.	6,924,602 B2	8/2005	Komiya
5,714,968 A	2/1998	Ikeda	6,937,220 B2	8/2005	Kitaura et al.
5,744,824 A	4/1998	Kousai et al.	6,940,214 B1	9/2005	Komiya et al.
5,745,660 A	4/1998	Kolpatzik et al.	6,954,194 B2	10/2005	Matsumoto et al.
5,748,160 A	5/1998	Shieh et al.	6,970,149 B2	11/2005	Chung et al.
5,758,129 A	5/1998	Gray et al.	6,975,142 B2	12/2005	Azami et al.
5,835,376 A	11/1998	Smith et al.	6,975,332 B2	12/2005	Arnold et al.
5,870,071 A	2/1999	Kawahata	6,995,519 B2	2/2006	Arnold et al.
5,874,803 A	2/1999	Garbuzov et al.	7,027,015 B2	4/2006	Booth, Jr. et al.
5,880,582 A	3/1999	Sawada	7,034,793 B2	4/2006	Sekiya et al.
5,903,248 A	5/1999	Irwin	7,038,392 B2	5/2006	Libsch et al.
5,917,280 A	6/1999	Burrows et al.	7,057,588 B2	6/2006	Asano et al.
5,949,398 A	9/1999	Kim	7,061,451 B2	6/2006	Kimura
5,952,789 A	9/1999	Stewart et al.	7,071,932 B2	7/2006	Libsch et al.
5,990,629 A	11/1999	Yamada et al.	7,106,285 B2	9/2006	Naugler
6,023,259 A	2/2000	Howard et al.	7,112,820 B2	9/2006	Chang et al.
6,069,365 A	5/2000	Chow et al.	7,113,864 B2	9/2006	Smith et al.
6,091,203 A	7/2000	Kawashima et al.	7,122,835 B1	10/2006	Ikeda et al.
6,097,360 A	8/2000	Holloman	7,129,914 B2	10/2006	Knapp et al.
6,100,868 A	8/2000	Lee et al.	7,164,417 B2	1/2007	Cok
6,144,222 A	11/2000	Ho	7,224,332 B2	5/2007	Cok
6,229,506 B1	5/2001	Dawson et al.	7,248,236 B2	7/2007	Nathan et al.
6,229,508 B1	5/2001	Kane	7,259,737 B2	8/2007	Ono et al.
6,246,180 B1	6/2001	Nishigaki	7,262,753 B2	8/2007	Tanghe et al.
6,252,248 B1	6/2001	Sano et al.	7,274,363 B2	9/2007	Ishizuka et al.
6,268,841 B1	7/2001	Cairns et al.	7,310,092 B2	12/2007	Imamura
6,288,696 B1	9/2001	Holloman	7,315,295 B2	1/2008	Kimura
6,307,322 B1	10/2001	Dawson et al.	7,317,434 B2	1/2008	Lan et al.
6,310,962 B1	10/2001	Chung et al.	7,321,348 B2	1/2008	Cok et al.
6,323,631 B1	11/2001	Juang	7,327,357 B2	2/2008	Jeong
6,333,729 B1 *	12/2001	Ha 345/98	7,333,077 B2	2/2008	Koyama et al.
6,388,653 B1 *	5/2002	Goto et al. 345/98	7,343,243 B2	3/2008	Smith et al.
6,392,617 B1	5/2002	Gleason	7,414,600 B2	8/2008	Nathan et al.
6,396,469 B1	5/2002	Miwa et al.	7,466,166 B2	12/2008	Date et al.
6,414,661 B1	7/2002	Shen et al.	7,495,501 B2	2/2009	Iwabuchi et al.
6,417,825 B1	7/2002	Stewart et al.	7,502,000 B2	3/2009	Yuki et al.
6,430,496 B1	8/2002	Smith et al.	7,515,124 B2	4/2009	Yaguma et al.
6,433,488 B1	8/2002	Bu	7,535,449 B2	5/2009	Miyazawa
6,473,065 B1	10/2002	Fan	7,554,512 B2	6/2009	Steer
6,475,845 B2	11/2002	Kimura	7,569,849 B2	8/2009	Nathan et al.
6,501,098 B2	12/2002	Yamazaki	7,595,776 B2 *	9/2009	Hashimoto et al. 345/76
6,501,466 B1	12/2002	Yamagishi et al.	7,604,718 B2	10/2009	Zhang et al.
6,522,315 B2	2/2003	Ozawa et al.	7,609,239 B2	10/2009	Chang
6,535,185 B2	3/2003	Kim et al.	7,612,745 B2	11/2009	Yumoto et al.
6,542,138 B1	4/2003	Shannon et al.	7,619,594 B2	11/2009	Hu
6,559,839 B1	5/2003	Ueno et al.	7,619,597 B2	11/2009	Nathan et al.
6,580,408 B1	6/2003	Bae et al.	7,639,211 B2	12/2009	Miyazawa
6,583,398 B2	6/2003	Harkin	7,683,899 B2	3/2010	Hirakata et al.
6,618,030 B2	9/2003	Kane et al.	7,688,289 B2	3/2010	Abe et al.
6,639,244 B1	10/2003	Yamazaki et al.	7,692,641 B2 *	4/2010	Kudo G09G 3/20 315/169.3
6,680,580 B1	1/2004	Sung	7,760,162 B2	7/2010	Miyazawa
6,686,699 B2	2/2004	Yumoto	7,808,008 B2	10/2010	Miyake
6,690,000 B1	2/2004	Muramatsu et al.	7,859,520 B2	12/2010	Kimura
6,693,610 B2	2/2004	Shannon et al.	7,889,159 B2	2/2011	Nathan et al.
6,694,248 B2	2/2004	Smith et al.	7,903,127 B2 *	3/2011	Kwon 345/690
6,697,057 B2	2/2004	Koyama et al.	7,920,116 B2 *	4/2011	Woo et al. 345/89
6,724,151 B2	4/2004	Yoo	7,944,414 B2	5/2011	Shirasaki et al.
6,734,636 B2	5/2004	Sanford et al.	7,978,170 B2	7/2011	Park et al.
			7,989,392 B2	8/2011	Crockett et al.
			7,995,008 B2	8/2011	Miwa
			8,063,852 B2	11/2011	Kwak et al.

(56)

References Cited

U.S. PATENT DOCUMENTS

- | | | | | | | | | |
|--------------|-----|---------|------------------|--------------|-----|---------|------------------|-----------------------|
| 8,102,343 | B2 | 1/2012 | Yatabe | 2004/0150595 | A1 | 8/2004 | Kasai | |
| 8,144,081 | B2 | 3/2012 | Miyazawa | 2004/0155841 | A1 | 8/2004 | Kasai | |
| 8,159,007 | B2 | 4/2012 | Barna et al. | 2004/0171619 | A1 | 9/2004 | Libsch et al. | |
| 8,242,979 | B2 | 8/2012 | Anzai et al. | 2004/0174349 | A1 | 9/2004 | Libsch | |
| 8,253,665 | B2 | 8/2012 | Nathan et al. | 2004/0174354 | A1 | 9/2004 | Ono | |
| 8,319,712 | B2 | 11/2012 | Nathan et al. | 2004/0183759 | A1 | 9/2004 | Stevenson et al. | |
| 2001/0002703 | A1 | 6/2001 | Koyama | 2004/0189627 | A1 | 9/2004 | Shirasaki et al. | |
| 2001/0009283 | A1 | 7/2001 | Arao et al. | 2004/0196275 | A1 | 10/2004 | Hattori | |
| 2001/0026257 | A1 | 10/2001 | Kimura | 2004/0227697 | A1 | 11/2004 | Mori | |
| 2001/0030323 | A1 | 10/2001 | Ikeda | 2004/0239696 | A1 | 12/2004 | Okabe | |
| 2001/0040541 | A1 | 11/2001 | Yoneda et al. | 2004/0251844 | A1 | 12/2004 | Hashido et al. | |
| 2001/0043173 | A1 | 11/2001 | Troutman | 2004/0252085 | A1 | 12/2004 | Miyagawa | |
| 2001/0045929 | A1 | 11/2001 | Prache | 2004/0252089 | A1 | 12/2004 | Ono et al. | |
| 2001/0052940 | A1 | 12/2001 | Hagihara et al. | 2004/0256617 | A1 | 12/2004 | Yamada et al. | |
| 2002/0000576 | A1 | 1/2002 | Inukai | 2004/0257353 | A1 | 12/2004 | Imamura et al. | |
| 2002/0011796 | A1 | 1/2002 | Koyama | 2004/0257355 | A1 | 12/2004 | Naugler | |
| 2002/0011799 | A1 | 1/2002 | Kimura | 2004/0263437 | A1 | 12/2004 | Hattori | |
| 2002/0012057 | A1 | 1/2002 | Kimura | 2005/0007357 | A1 | 1/2005 | Yamashita et al. | |
| 2002/0030190 | A1 | 3/2002 | Ohtani et al. | 2005/0052379 | A1* | 3/2005 | Waterman | 345/87 |
| 2002/0047565 | A1 | 4/2002 | Nara et al. | 2005/0057459 | A1 | 3/2005 | Miyazawa | |
| 2002/0052086 | A1 | 5/2002 | Maeda | 2005/0067970 | A1 | 3/2005 | Libsch et al. | |
| 2002/0080108 | A1 | 6/2002 | Wang | 2005/0067971 | A1 | 3/2005 | Kane | |
| 2002/0084463 | A1 | 7/2002 | Sanford et al. | 2005/0083270 | A1 | 4/2005 | Miyazawa | |
| 2002/0101172 | A1 | 8/2002 | Bu | 2005/0110420 | A1 | 5/2005 | Arnold et al. | |
| 2002/0117722 | A1 | 8/2002 | Osada et al. | 2005/0110727 | A1* | 5/2005 | Shin | 345/76 |
| 2002/0140712 | A1* | 10/2002 | Ouchi et al. | 2005/0123193 | A1 | 6/2005 | Lamberg et al. | |
| 2002/0158587 | A1 | 10/2002 | Komiya | 2005/0140610 | A1 | 6/2005 | Smith et al. | |
| 2002/0158666 | A1 | 10/2002 | Azami et al. | 2005/0145891 | A1 | 7/2005 | Abe | |
| 2002/0158823 | A1 | 10/2002 | Zavracky et al. | 2005/0156831 | A1 | 7/2005 | Yamazaki et al. | |
| 2002/0171613 | A1* | 11/2002 | Goto et al. | 2005/0168416 | A1* | 8/2005 | Hashimoto et al. | 345/76 |
| 2002/0186214 | A1 | 12/2002 | Siwinski | 2005/0206590 | A1 | 9/2005 | Sasaki et al. | |
| 2002/0190971 | A1 | 12/2002 | Nakamura et al. | 2005/0219188 | A1* | 10/2005 | Kawabe et al. | 345/94 |
| 2002/0195967 | A1 | 12/2002 | Kim et al. | 2005/0243037 | A1 | 11/2005 | Eom et al. | |
| 2002/0195968 | A1 | 12/2002 | Sanford et al. | 2005/0248515 | A1 | 11/2005 | Naugler et al. | |
| 2003/0001828 | A1 | 1/2003 | Asano | 2005/0258867 | A1 | 11/2005 | Miyazawa | |
| 2003/0020413 | A1 | 1/2003 | Oomura | 2005/0285825 | A1 | 12/2005 | Eom et al. | |
| 2003/0030603 | A1 | 2/2003 | Shimoda | 2006/0012311 | A1 | 1/2006 | Ogawa | |
| 2003/0062524 | A1 | 4/2003 | Kimura | 2006/0038750 | A1 | 2/2006 | Inoue et al. | |
| 2003/0062844 | A1 | 4/2003 | Miyazawa | 2006/0038758 | A1 | 2/2006 | Routley et al. | |
| 2003/0076048 | A1 | 4/2003 | Rutherford | 2006/0038762 | A1 | 2/2006 | Chou | |
| 2003/0090445 | A1 | 5/2003 | Chen et al. | 2006/0044236 | A1* | 3/2006 | Kim | G09G 3/3233
345/82 |
| 2003/0090447 | A1 | 5/2003 | Kimura | 2006/0066533 | A1 | 3/2006 | Sato et al. | |
| 2003/0090481 | A1 | 5/2003 | Kimura | 2006/0077077 | A1* | 4/2006 | Kwon | 341/50 |
| 2003/0095087 | A1 | 5/2003 | Libsch | 2006/0092185 | A1 | 5/2006 | Jo et al. | |
| 2003/0098829 | A1 | 5/2003 | Chen et al. | 2006/0125408 | A1 | 6/2006 | Nathan et al. | |
| 2003/0107560 | A1 | 6/2003 | Yumoto et al. | 2006/0139253 | A1 | 6/2006 | Choi et al. | |
| 2003/0107561 | A1 | 6/2003 | Uchino et al. | 2006/0145964 | A1 | 7/2006 | Park et al. | |
| 2003/0111966 | A1 | 6/2003 | Mikami et al. | 2006/0191178 | A1 | 8/2006 | Sempel et al. | |
| 2003/0112205 | A1 | 6/2003 | Yamada | 2006/0209012 | A1* | 9/2006 | Hagood | 345/109 |
| 2003/0112208 | A1 | 6/2003 | Okabe et al. | 2006/0214888 | A1 | 9/2006 | Schneider et al. | |
| 2003/0117348 | A1 | 6/2003 | Knapp et al. | 2006/0221009 | A1 | 10/2006 | Miwa | |
| 2003/0122474 | A1 | 7/2003 | Lee | 2006/0227082 | A1 | 10/2006 | Ogata et al. | |
| 2003/0122747 | A1 | 7/2003 | Shannon et al. | 2006/0232522 | A1 | 10/2006 | Roy et al. | |
| 2003/0128199 | A1 | 7/2003 | Kimura | 2006/0244391 | A1 | 11/2006 | Shishido et al. | |
| 2003/0151569 | A1 | 8/2003 | Lee et al. | 2006/0244697 | A1 | 11/2006 | Lee et al. | |
| 2003/0156104 | A1 | 8/2003 | Morita | 2006/0261841 | A1 | 11/2006 | Fish | |
| 2003/0169241 | A1* | 9/2003 | LeChevalier | 2006/0290614 | A1 | 12/2006 | Nathan et al. | |
| 2003/0169247 | A1 | 9/2003 | Kawabe et al. | 2007/0001939 | A1* | 1/2007 | Hashimoto et al. | 345/76 |
| 2003/0179626 | A1 | 9/2003 | Sanford et al. | 2007/0001945 | A1 | 1/2007 | Yoshida et al. | |
| 2003/0185438 | A1 | 10/2003 | Osawa et al. | 2007/0008251 | A1 | 1/2007 | Kohno et al. | |
| 2003/0189535 | A1 | 10/2003 | Matsumoto et al. | 2007/0008297 | A1 | 1/2007 | Bassetti | |
| 2003/0197663 | A1 | 10/2003 | Lee et al. | 2007/0035489 | A1 | 2/2007 | Lee | |
| 2003/0214465 | A1 | 11/2003 | Kimura | 2007/0035707 | A1 | 2/2007 | Margulis | |
| 2003/0227262 | A1 | 12/2003 | Kwon | 2007/0040773 | A1* | 2/2007 | Lee et al. | 345/77 |
| 2003/0230141 | A1 | 12/2003 | Gilmour et al. | 2007/0040782 | A1 | 2/2007 | Woo et al. | |
| 2003/0230980 | A1 | 12/2003 | Forrest et al. | 2007/0063932 | A1 | 3/2007 | Nathan et al. | |
| 2004/0004589 | A1 | 1/2004 | Shih | 2007/0080908 | A1 | 4/2007 | Nathan et al. | |
| 2004/0032382 | A1 | 2/2004 | Cok et al. | 2007/0085801 | A1 | 4/2007 | Park et al. | |
| 2004/0041750 | A1 | 3/2004 | Abe | 2007/0109232 | A1 | 5/2007 | Yamamoto et al. | |
| 2004/0066357 | A1 | 4/2004 | Kawasaki | 2007/0128583 | A1 | 6/2007 | Miyazawa | |
| 2004/0070557 | A1 | 4/2004 | Asano et al. | 2007/0164941 | A1 | 7/2007 | Park et al. | |
| 2004/0129933 | A1 | 7/2004 | Nathan et al. | 2007/0182671 | A1 | 8/2007 | Nathan et al. | |
| 2004/0130516 | A1* | 7/2004 | Nathan et al. | 2007/0236430 | A1 | 10/2007 | Fish | |
| 2004/0135749 | A1 | 7/2004 | Kondakov et al. | 2007/0241999 | A1 | 10/2007 | Lin | |
| 2004/0145547 | A1 | 7/2004 | Oh | 2007/0242008 | A1 | 10/2007 | Cummings | |
| | | | | 2008/0001544 | A1 | 1/2008 | Murakami et al. | |
| | | | | 2008/0043044 | A1* | 2/2008 | Woo et al. | 345/690 |
| | | | | 2008/0048951 | A1 | 2/2008 | Naugler et al. | |

(56)

References Cited

U.S. PATENT DOCUMENTS

2008/0055134 A1 3/2008 Li et al.
 2008/0074360 A1 3/2008 Lu et al.
 2008/0088549 A1 4/2008 Nathan et al.
 2008/0094426 A1 4/2008 Kimpe
 2008/0122819 A1 5/2008 Cho et al.
 2008/0129906 A1 6/2008 Lin et al.
 2008/0228562 A1 9/2008 Smith et al.
 2008/0231641 A1 9/2008 Miyashita
 2008/0265786 A1 10/2008 Koyama
 2008/0290805 A1 11/2008 Yamada et al.
 2009/0009459 A1 1/2009 Miyashita
 2009/0015532 A1 1/2009 Katayama et al.
 2009/0058789 A1 3/2009 Hung et al.
 2009/0121988 A1 5/2009 Feng et al.
 2009/0146926 A1 6/2009 Sung et al.
 2009/0153448 A1 6/2009 Tomida et al.
 2009/0153459 A9 6/2009 Han et al.
 2009/0174628 A1 7/2009 Wang et al.
 2009/0201230 A1 8/2009 Smith
 2009/0201281 A1 8/2009 Routley et al.
 2009/0251486 A1 10/2009 Sakakibara et al.
 2009/0278777 A1 11/2009 Wang et al.
 2009/0289964 A1 11/2009 Miyachi
 2010/0039451 A1 2/2010 Jung
 2010/0039453 A1 2/2010 Nathan et al.
 2010/0207920 A1 8/2010 Chaji et al.
 2010/0225634 A1 9/2010 Levey et al.
 2010/0251295 A1 9/2010 Amento et al.
 2010/0269889 A1 10/2010 Reinhold et al.
 2010/0277400 A1 11/2010 Jeong
 2010/0315319 A1 12/2010 Cok et al.
 2011/0050741 A1 3/2011 Jeong
 2011/0069089 A1 3/2011 Kopf et al.
 2012/0299976 A1 11/2012 Chen et al.

FOREIGN PATENT DOCUMENTS

CA 1 294 034 1/1992
 CA 2 249 592 7/1998
 CA 2 303 302 3/1999
 CA 2303302 A1 3/1999
 CA 2 368 386 9/1999
 CA 2 242 720 1/2000
 CA 2 354 018 6/2000
 CA 2 432 530 7/2002
 CA 2 436 451 8/2002
 CA 2 507 276 8/2002
 CA 2 463 653 1/2004
 CA 2 498 136 3/2004
 CA 2 522 396 11/2004
 CA 2 438 363 2/2005
 CA 2 443 206 3/2005
 CA 2 519 097 3/2005
 CA 2 472 671 12/2005
 CA 2 523 841 1/2006
 CA 2 567 076 1/2006
 CA 2 495 726 7/2006
 CA 2 557 713 11/2006
 CA 2 526 782 C 8/2007
 CA 2 651 893 11/2007
 CA 2651893 A1 11/2007
 CA 2 672 590 10/2009
 CA 2672590 A1 10/2009
 CN 1601594 A 3/2005
 CN 1886774 12/2006
 DE 202006007613 9/2006
 EP 0 478 186 4/1992
 EP 1 028 471 A 8/2000
 EP 1 130 565 A1 9/2001
 EP 1 194 013 4/2002
 EP 1 321 922 6/2003
 EP 1 335 430 A1 8/2003
 EP 1 381 019 1/2004
 EP 1 429 312 A 6/2004
 EP 1 439 520 A2 7/2004

EP 1 465 143 A 10/2004
 EP 1 473 689 A 11/2004
 EP 1 517 290 A2 3/2005
 EP 1 521 203 A2 4/2005
 GB 2 399 935 9/2004
 GB 2 460 018 11/2009
 JP 09 090405 4/1997
 JP 10-254410 9/1998
 JP 11 231805 8/1999
 JP 2002-278513 9/2002
 JP 2003-076331 3/2003
 JP 2003-099000 4/2003
 JP 2003-173165 6/2003
 JP 2003-186439 7/2003
 JP 2003-195809 7/2003
 JP 2003-271095 9/2003
 JP 2003-308046 10/2003
 JP 2004-054188 2/2004
 JP 2004-226960 8/2004
 JP 2005-004147 1/2005
 JP 2005-099715 4/2005
 JP 2005-258326 9/2005
 JP 2005-338819 12/2005
 TW 569173 1/2004
 TW 200526065 8/2005
 TW 1239501 9/2005
 WO WO 98/11554 3/1998
 WO WO 99/48079 9/1999
 WO WO 01/27910 A1 4/2001
 WO WO 02/067327 A 8/2002
 WO WO 03/034389 4/2003
 WO WO 03/063124 7/2003
 WO WO 03/075256 9/2003
 WO WO 2004/003877 1/2004
 WO WO 2004/015668 A1 2/2004
 WO WO 2004/034364 4/2004
 WO WO 2005/022498 3/2005
 WO WO 2005/055185 6/2005
 WO WO 2005/055186 A1 6/2005
 WO WO 2005/069267 7/2005
 WO WO 2005/122121 12/2005
 WO WO 2006/063448 6/2006
 WO WO 2006/128069 11/2006
 WO WO 2008/057369 5/2008
 WO WO 2008/0290805 11/2008
 WO WO 2009/059028 5/2009
 WO WO 2009/127065 10/2009
 WO WO 2010/066030 6/2010
 WO WO 2010/120733 10/2010

OTHER PUBLICATIONS

Chaji et al.: "A fast settling current driver based on the CCII for AMOLED displays"; dated Dec. 2009 (6 pages).
 Chaji et al.: "Stable RGBW AMOLED display with OLED degradation compensation using electrical feedback"; dated Feb. 2010 (2 pages).
 Chapter 3: Color Spaces"Keith Jack:"Video Demystified: "A Handbook for the Digital Engineer" 2001 Referex ORD-0000-00-00 USA EP040425529 ISBN: 1-878707-56-6 pp. 32-33.
 Chapter 8: Alternative Flat Panel Display 1-25 Technologies; Willem den Boer: "Active Matrix Liquid Crystal Display: Fundamentals and Applications" 2005 Referex ORD-0000-00-00 U.K.; XP040426102 ISBN: 0/7506-7813-5 pp. 206-209 p. 208.
 European Partial Search Report Application No. 12 15 6251.6 European Patent Office dated May 30, 2012 (7 pages).
 European Patent Office Communication Application No. 05 82 1114 dated Jan. 11, 2013 (9 pages).
 European Patent Office Communication with Supplemental European Search Report for EP Application No. 07 70 1644.2 dated Aug. 18 2009 (12 pages).
 European Search Report Application No. 10 83 4294.0/1903 dated Apr. 8, 2013 (9 pages).
 European Search Report Application No. EP 05 80 7905 dated Apr. 2, 2009 (5 pages).

(56)

References Cited

OTHER PUBLICATIONS

European Search Report Application No. EP 05 82 1114 dated Mar. 27, 2009 (2 pages).

European Search Report Application No. EP 07 70 1644 dated Aug. 5, 2009.

European Search Report Application No. EP 10 17 5764 dated Oct. 18, 2010 (2 pages).

European Search Report Application No. EP 10 82 9593.2 European Patent Office dated May 17, 2013 (7 pages).

European Search Report Application No. EP 12 15 6251.6 European Patent Office dated Oct. 12, 2012 (18 pages).

European Search Report Application No. EP. 11 175 225.9 dated Nov. 4, 2011 (9 pages).

European Supplementary Search Report Application No. EP 09 80 2309 dated May 8, 2011 (14 pages).

European Supplementary Search Report Application No. EP 09 83 1339.8 dated Mar. 26, 2012 (11 pages).

Extended European Search Report Application No. EP 06 75 2777.0 dated Dec. 6, 2010 (21 pages).

Extended European Search Report Application No. EP 09 73 2338.0 dated May 24, 2011 (8 pages).

Extended European Search Report Application No. EP 11 17 5223., 4 mailed Nov. 8, 2011 (8 pages).

Extended European Search Report Application No. EP 12 17 4465.0 European Patent Office dated Sep. 7, 2012 (9 pages).

Fan et al. "LTIPS_TFT Pixel Circuit Compensation for TFT Threshold Voltage Shift and IR-Drop On the Power Line for Amoled Displays" 5 pages copyright 2012.

Goh et al. "A New a-Si:H Thin-Film Transistor Pixel Circuit for Active-Matrix Organic Light-Emitting Diodes" IEEE Electron Device Letters vol. 24 No. 9 Sep. 2003 pp. 583-585.

International Search Report Application No. PCT/CA2005/001844 dated Mar. 28, 2006 (2 pages).

International Search Report Application No. PCT/CA2006/000941 dated Oct. 3, 2006 (2 pages).

International Search Report Application No. PCT/CA2007/000013 dated May 7, 2007.

International Search Report Application No. PCT/CA2009/001049 mailed Dec. 7, 2009 (4 pages).

International Search Report Application No. PCT/CA2009/001769 dated Apr. 8, 2010.

International Search Report Application No. PCT/IB2010/002898 Canadian Intellectual Property Office dated Jul. 28, 2009 (5 pages).

International Search Report Application No. PCT/IB2010/055481 dated Apr. 7, 2011 (3 pages).

International Search Report Application No. PCT/IB2011/051103 dated Jul. 8, 2011 3 pages).

International Search Report Application No. PCT/IB2012/052651 5 pages dated Sep. 11, 2012.

International Searching Authority Written Opinion Application No. PCT/IB2010/055481 dated Apr. 7, 2011 (6 pages).

International Searching Authority Written Opinion Application No. PCT/IB2012/052651 6 pages dated Sep. 11, 2012.

International Searching Authority Written Opinion Application No. PCT/IB2011/051103 dated Jul. 8, 2011 6 pages.

International Searching Authority Written Opinion Application No. PCT/IB2010/002898 Canadian Intellectual Property Office dated Mar. 30, 2011 (8 pages).

International Searching Authority Written Opinion Application No. PCT/CA2009/001769 dated Apr. 8, 2010.

Ma e y et al: "Organic Light-Emitting Diode/Thin Film Transistor Integration for foldable Displays" Conference record of the 1997 International display research conference and international workshops on LCD technology and emissive technology. Toronto Sep. 15-19, 1997 (6 pages).

Nathan et al. "Amorphous Silicon Thin Film Transistor Circuit Integration for Organic LED Displays on Glass and Plastic" IEEE Journal of Solid-State Circuits vol. 39 No. 9 Sep. 2004 pp. 1477-1486.

Nathan et al.: "Call for papers second international workshop on compact thin-film transistor (TFT) modeling for circuit simulation"; dated Sep. 2009 (1 page).

Nathan et al.: "Thin film imaging technology on glass and plastic"; dated Oct. 31-Nov. 2, 2000 (4 pages).

Ono et al. "Shared Pixel Compensation Circuit for AM-OLED Displays" Proceedings of the 9th Asian Symposium on Information Display (ASID) pp. 462-465 New Delhi dated Oct. 8-12 2006 (4 pages).

Smith, Lindsay I., "A tutorial on Principal Components Analysis," dated Feb. 26, 2001 (27 pages).

Stewart M. et al. "Polysilicon TFT technology for active matrix OLED displays" IEEE transactions on electron devices vol. 48 No. 5 May 2001 (7 pages).

Yi He et al. "Current-Source a-Si:H Thin Film Transistor Circuit for Active-Matrix Organic Light-Emitting Displays" IEEE Electron Device Letters vol. 21 No. 12 Dec. 2000 pp. 590-592.

International Search Report Application No. PCT/IB2013/059074, dated Dec. 18, 2013 (5 pages).

International Searching Authority Written Opinion Application No. PCT/IB2013/059074, dated Dec. 18, 2013 (8 pages).

Supplementary European Search Report for EP 09 80 2309, dated May 8, 2011 (14 pages).

Ahnood et al.: "Effect of threshold voltage instability on field effect mobility in thin film transistors deduced from constant current measurements"; dated Aug. 2009.

Alexander et al.: "Pixel circuits and drive schemes for glass and elastic AMOLED displays"; dated Jul. 2005 (9 pages).

Ashtiani et al.: "AMOLED Pixel Circuit With Electronic Compensation of Luminance Degradation"; dated Mar. 2007 (4 pages).

Chaji et al.: "A Current-Mode Comparator for Digital Calibration of Amorphous Silicon AMOLED Displays"; dated Jul. 2008 (5 pages).

Chaji et al.: "A Low-Cost Stable Amorphous Silicon AMOLED Display with Full V~T- and V~O~L~E~D Shift Compensation"; dated May 2007 (4 pages).

Chaji et al.: "A low-power driving scheme for a-Si:H active-matrix organic light-emitting diode displays"; dated Jun. 2005 (4 pages).

Chaji et al.: "A low-power high-performance digital circuit for deep submicron technologies"; dated Jun. 2005 (4 pages).

Chaji et al.: "A novel a-Si:H AMOLED pixel circuit based on short-term stress stability of a-Si:H TFTs"; dated Oct. 2005 (3 pages).

Chaji et al.: "A Novel Driving Scheme and Pixel Circuit for AMOLED Displays"; dated Jun. 2006 (4 pages).

Chaji et al.: "A novel driving scheme for high-resolution large-area a-Si:H AMOLED displays"; dated Aug. 2005 (4 pages).

Chaji et al.: "A Stable Voltage-Programmed Pixel Circuit for a-Si:H AMOLED Displays"; dated Dec. 2006 (12 pages).

Chaji et al.: "A Sub- μ A fast-settling current-programmed pixel circuit for AMOLED displays"; dated Sep. 2007.

Chaji et al.: "An Enhanced and Simplified Optical Feedback Pixel Circuit for AMOLED Displays"; dated Oct. 2006.

Chaji et al.: "Compensation technique for DC and transient instability of thin film transistor circuits for large-area devices"; dated Aug. 2008.

Chaji et al.: "Driving scheme for stable operation of 2-TFT a-Si AMOLED pixel"; dated Apr. 2005 (2 pages).

Chaji et al.: "Dynamic-effect compensating technique for stable a-Si:H AMOLED displays"; dated Aug. 2005 (4 pages).

Chaji et al.: "Electrical Compensation of OLED Luminance Degradation"; dated Dec. 2007 (3 pages).

Chaji et al.: "eUTDSP: a design study of a new VLIW-based DSP architecture"; dated May 2003 (4 pages).

Chaji et al.: "Fast and Offset-Leakage Insensitive Current-Mode Line Driver for Active Matrix Displays and Sensors"; dated Feb. 2009 (8 pages).

Chaji et al.: "High Speed Low Power Adder Design With A New Logic Style: Pseudo Dynamic Logic (SDL)"; dated Oct. 2001 (4 pages).

Chaji et al.: "High-precision, fast current source for large-area current-programmed a-Si flat panels"; dated Sep. 2006 (4 pages).

Chaji et al.: "Low-Cost AMOLED Television with IGNIS Compensating Technology"; dated May 2008 (4 pages).

(56)

References Cited

OTHER PUBLICATIONS

Chaji et al.: "Low-Cost Stable a-Si:H AMOLED Display for Portable Applications"; dated Jun. 2006 (4 pages).

Chaji et al.: "Low-Power Low-Cost Voltage-Programmed a-Si:H AMOLED Display"; dated Jun. 2008 (5 pages).

Chaji et al.: "Merged phototransistor pixel with enhanced near infrared response and flicker noise reduction for biomolecular imaging"; dated Nov. 2008 (3 pages).

Chaji et al.: "Parallel Addressing Scheme for Voltage-Programmed Active-Matrix OLED Displays"; dated May 2007 (6 pages).

Chaji et al.: "Pseudo dynamic logic (SDL): a high-speed and low-power dynamic logic family"; dated 2002 (4 pages).

Chaji et al.: "Stable a-Si:H circuits based on short-term stress stability of amorphous silicon thin film transistors"; dated May 2006 (4 pages).

Chaji et al.: "Stable Pixel Circuit for Small-Area High-Resolution a-Si:H AMOLED Displays"; dated Oct. 2008 (6 pages).

Chaji et al.: "Thin-Film Transistor Integration for Biomedical Imaging and AMOLED Displays"; dated May 2008 (177 pages).

Jafarabadiashtiani et al.: "A New Driving Method for a-Si AMOLED Displays Based on Voltage Feedback"; dated May 2005 (4 pages).

Lee et al.: "Ambipolar Thin-Film Transistors Fabricated by PECVD Nanocrystalline Silicon"; dated May 2006 (6 pages).

Matsueda et al.: "35.1: 2.5-in. AMOLED with Integrated 6-bit Gamma Compensated Digital Data Driver"; dated May 2004 (4 pages).

Nathan et al.: "Backplane Requirements for Active Matrix Organic Light Emitting Diode Displays"; dated Sep. 2006 (16 pages).

Nathan et al.: "Driving schemes for a-Si and LTPS AMOLED displays"; dated Dec. 2005 (11 pages).

Nathan et al.: "Invited Paper: a-Si for AMOLED—Meeting the Performance and Cost Demands of Display Applications (Cell Phone to HDTV)"; dated Jun. 2006 (4 pages).

Philipp: "Charge transfer sensing" *Sensor Review*, vol. 19, No. 2, Dec. 31, 1999 (Dec. 31, 1999) 10 pages.

Rafati et al.: "Comparison of a 17 b multiplier in Dual-rail domino and in Dual-rail D L (D L) logic styles"; dated 2002 (4 pages).

Safavaian et al.: "Three-TFT image sensor for real-time digital X-ray imaging"; dated Feb. 2, 2006 (2 pages).

Safavian et al.: "3-TFT active pixel sensor with correlated double sampling readout circuit for real-time medical x-ray imaging"; dated Jun. 2006 (4 pages).

Safavian et al.: "A novel current scaling active pixel sensor with correlated double sampling readout circuit for real time medical x-ray imaging"; dated May 2007 (7 pages).

Safavian et al.: "A novel hybrid active-passive pixel with correlated double sampling CMOS readout circuit for medical x-ray imaging"; dated May 2008 (4 pages).

Safavian et al.: "Self-compensated a-Si:H detector with current-mode readout circuit for digital X-ray fluoroscopy"; dated Aug. 2005 (4 pages).

Safavian et al.: "TFT active image sensor with current-mode readout circuit for digital x-ray fluoroscopy [5969D-82]"; dated Sep. 2005 (9 pages).

Vygranenko et al.: "Stability of indium-oxide thin-film transistors by reactive ion beam assisted deposition"; dated Feb. 2009.

Wang et al.: "Indium oxides by reactive ion beam assisted evaporation: From material study to device application"; dated Mar. 2009 (6 pages).

* cited by examiner

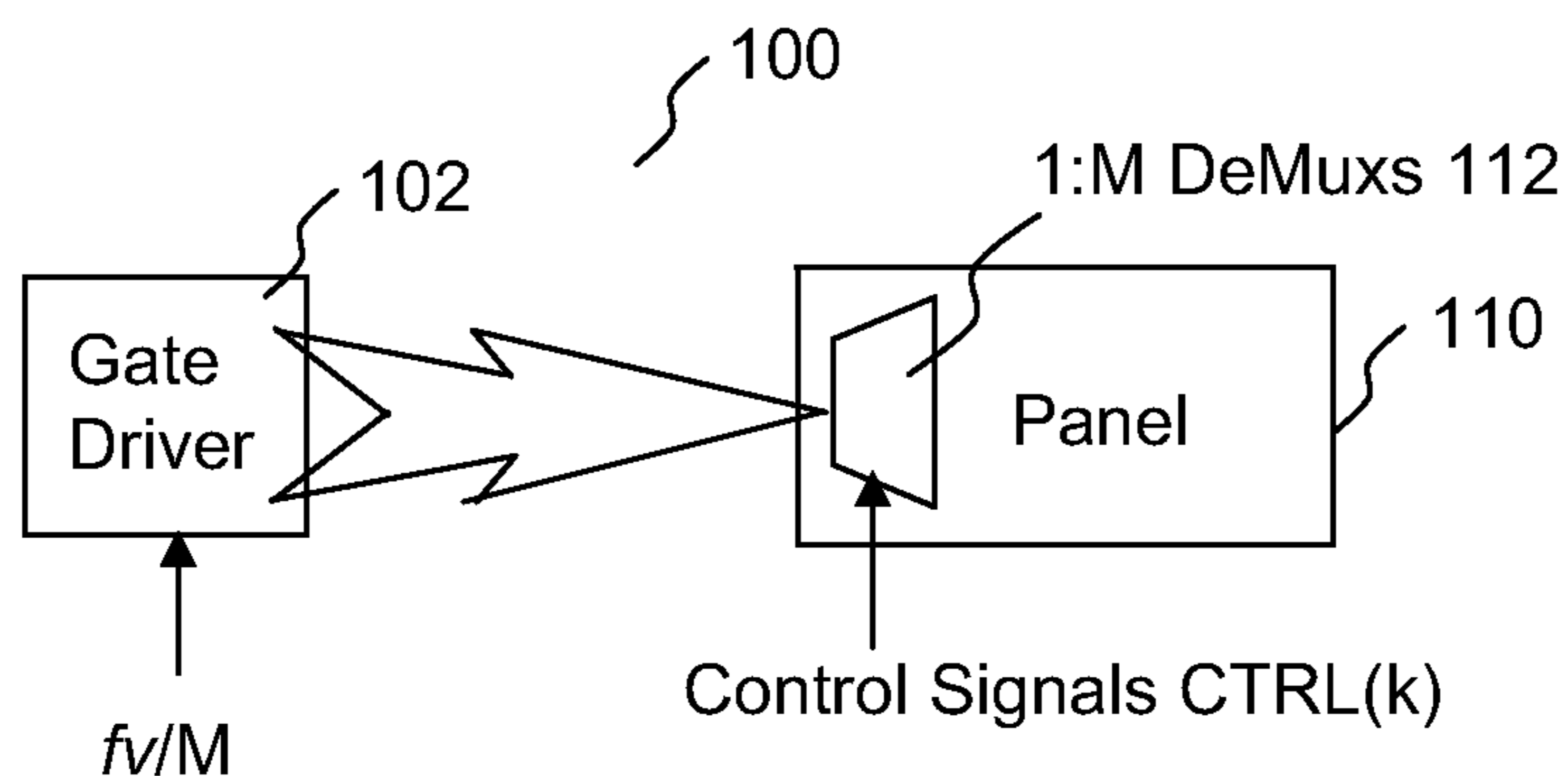


FIG. 1A

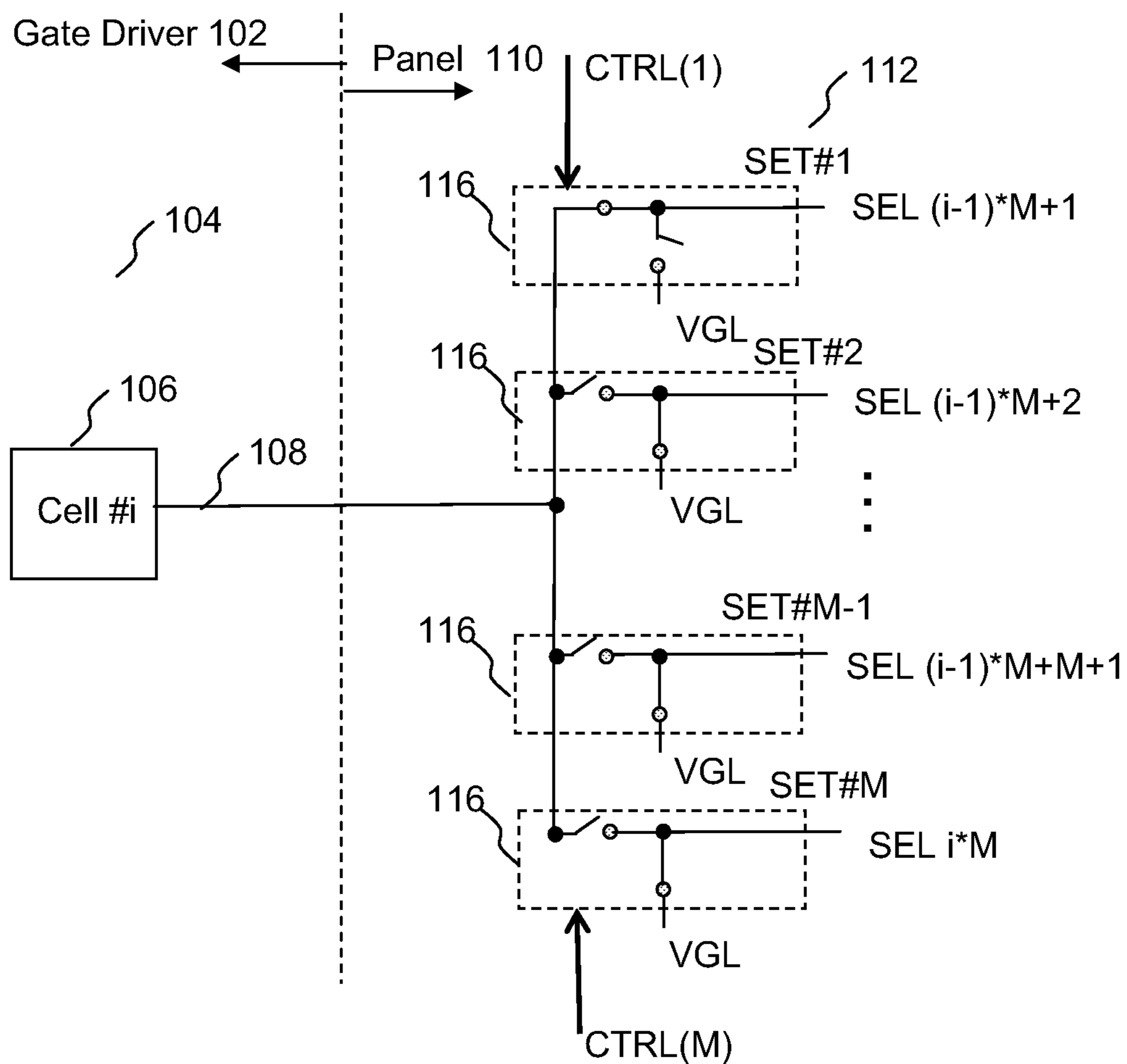


FIG. 1B

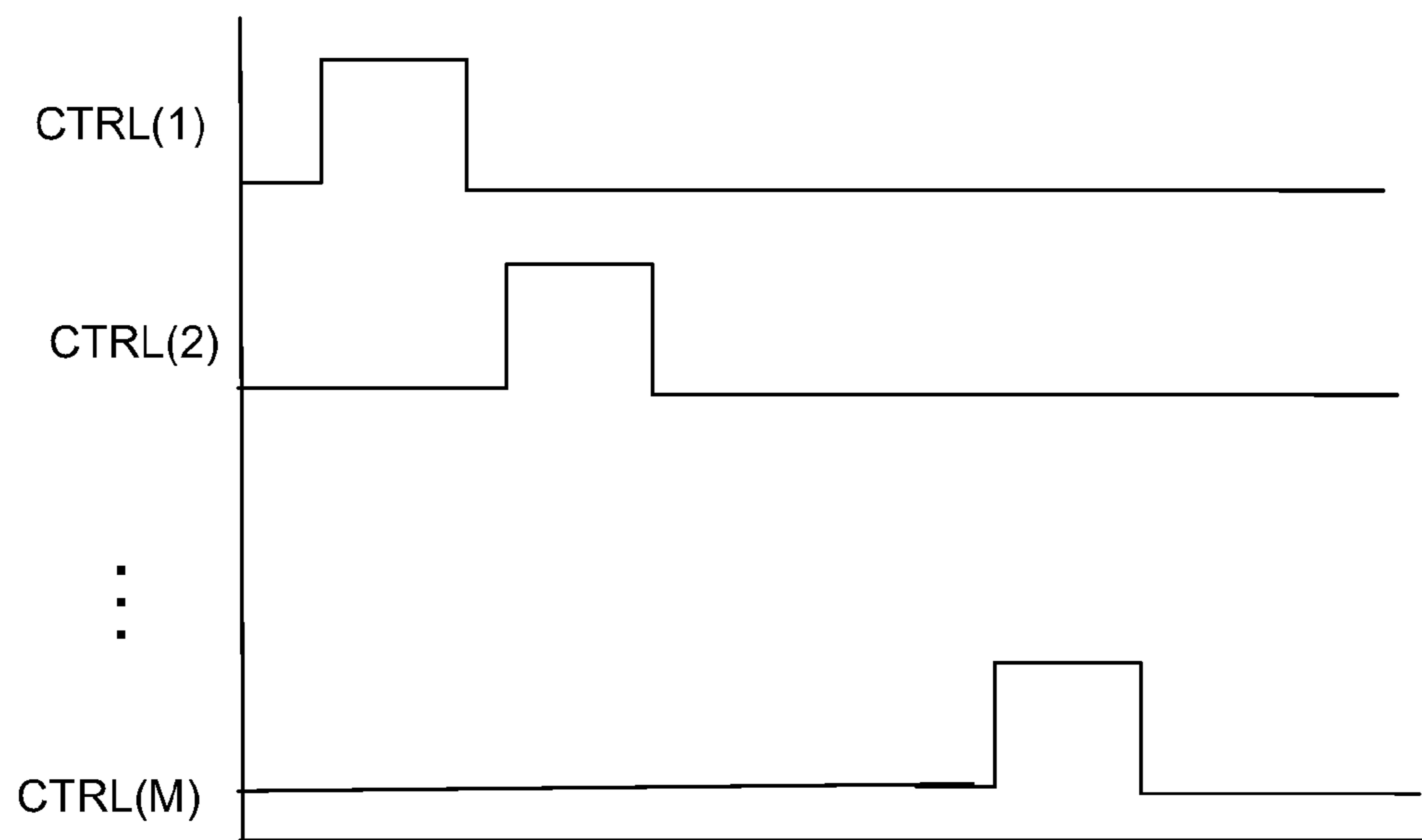


FIG. 2

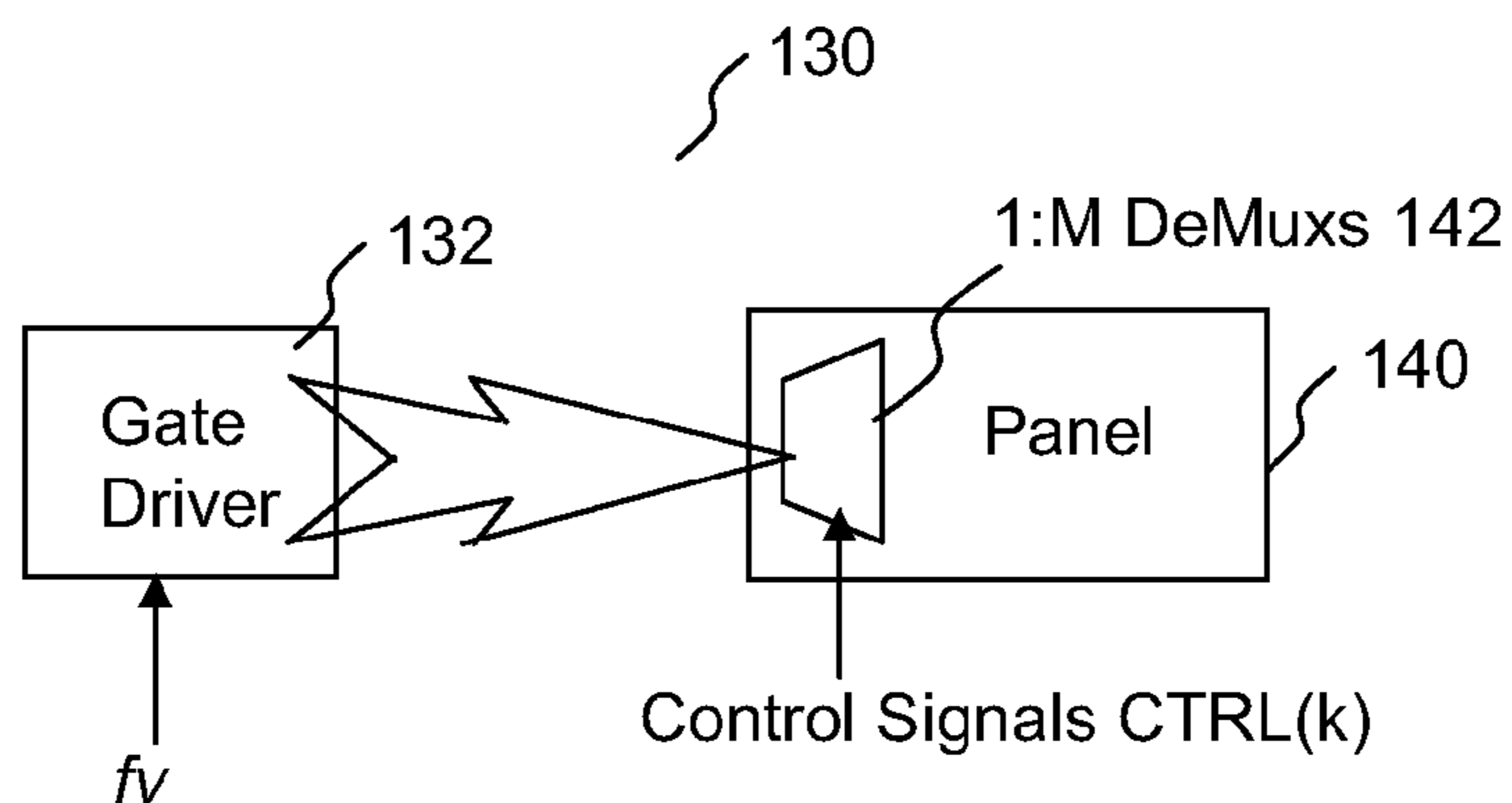


FIG. 3A

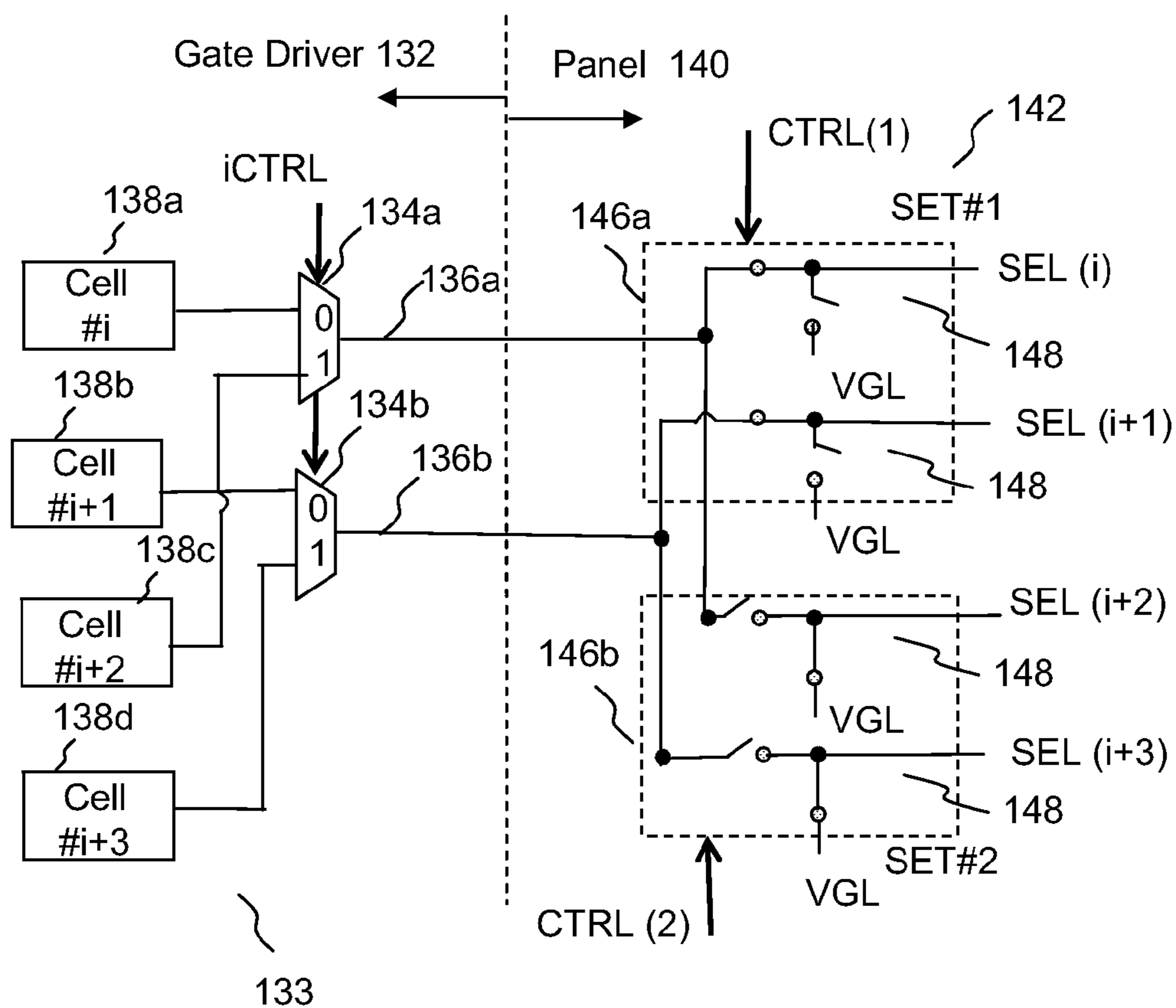


FIG. 3B

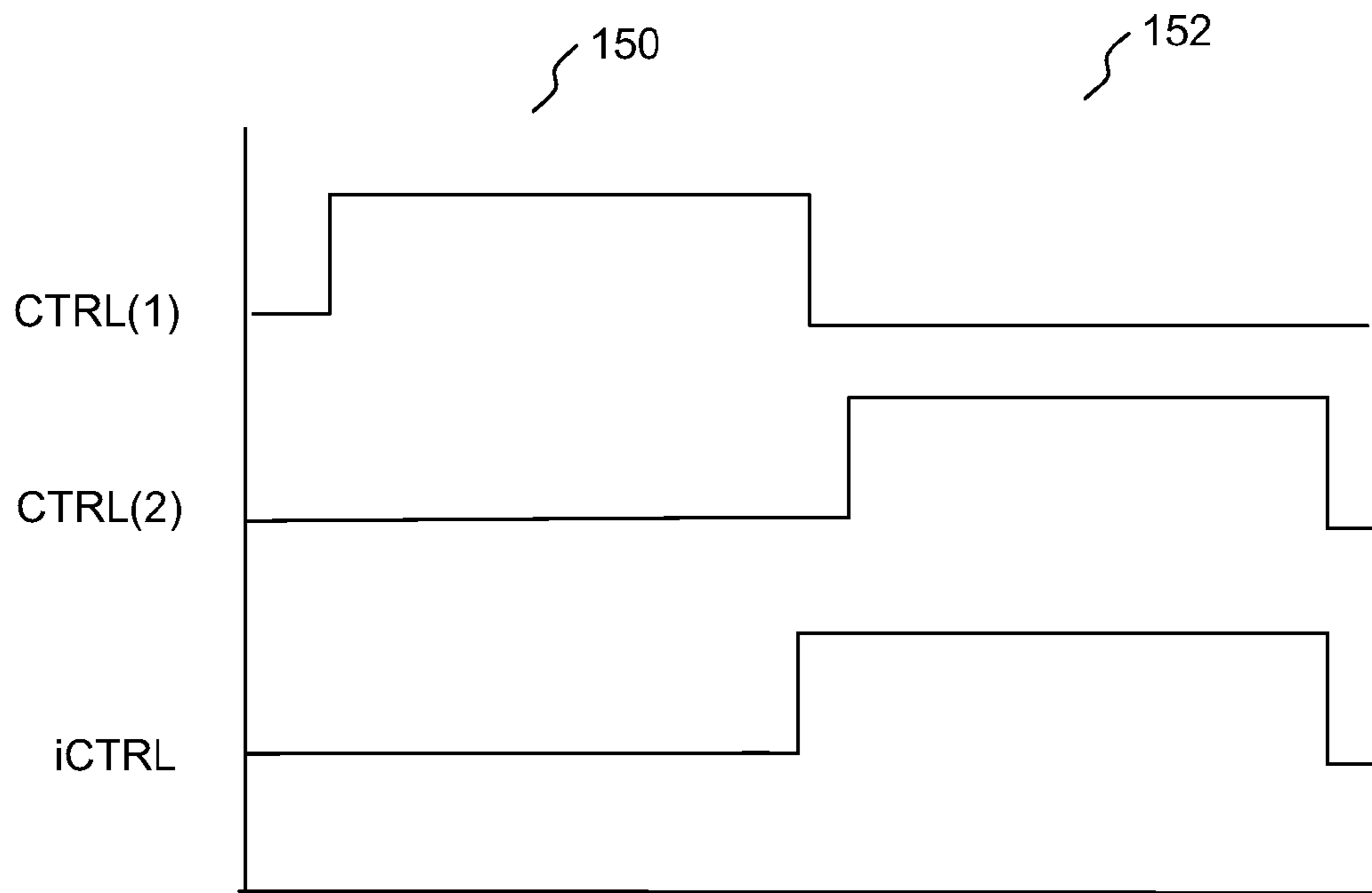


FIG. 4

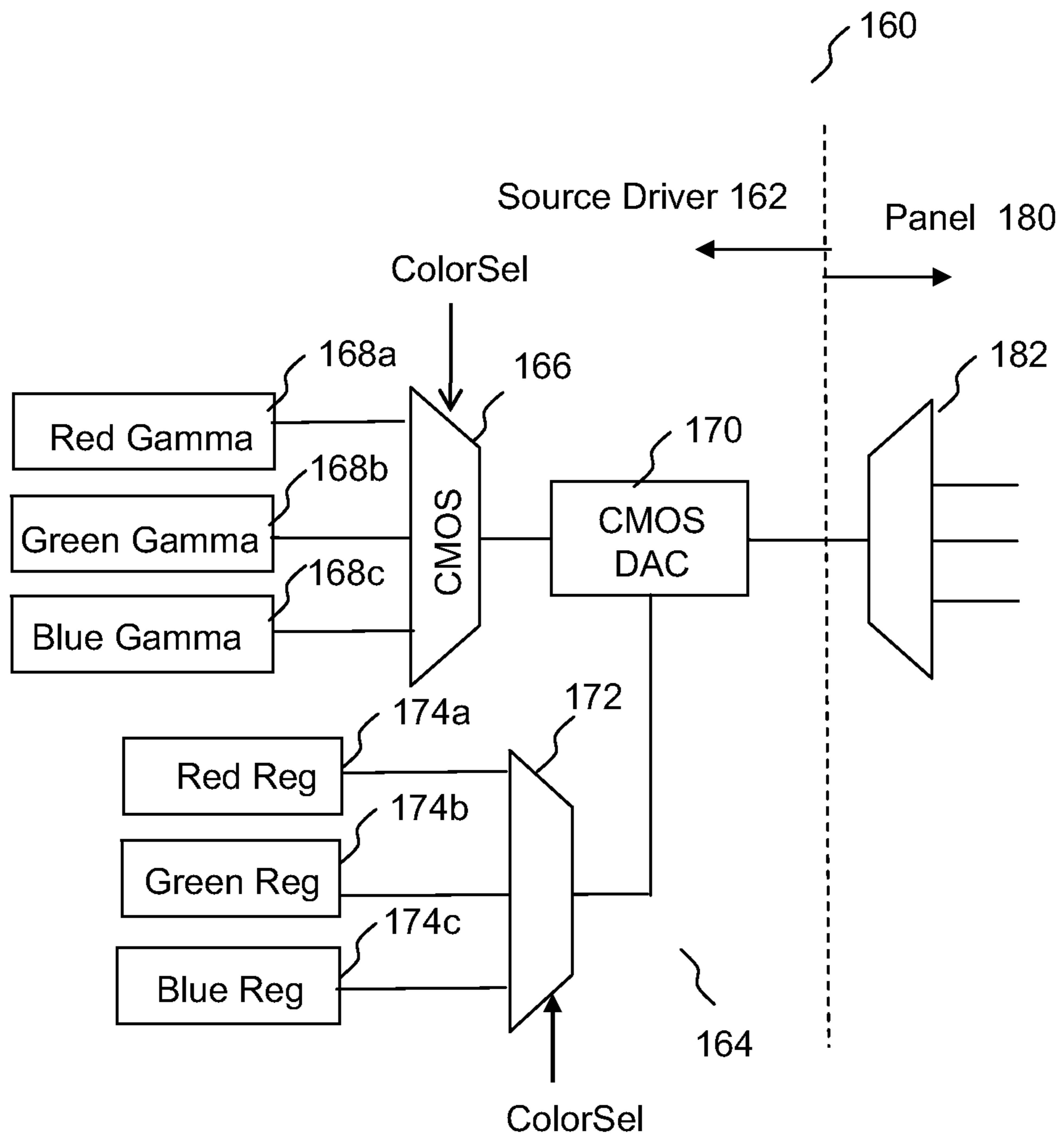


FIG. 5

Red	Green	Blue	Red	Green	Blue
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FIG. 6

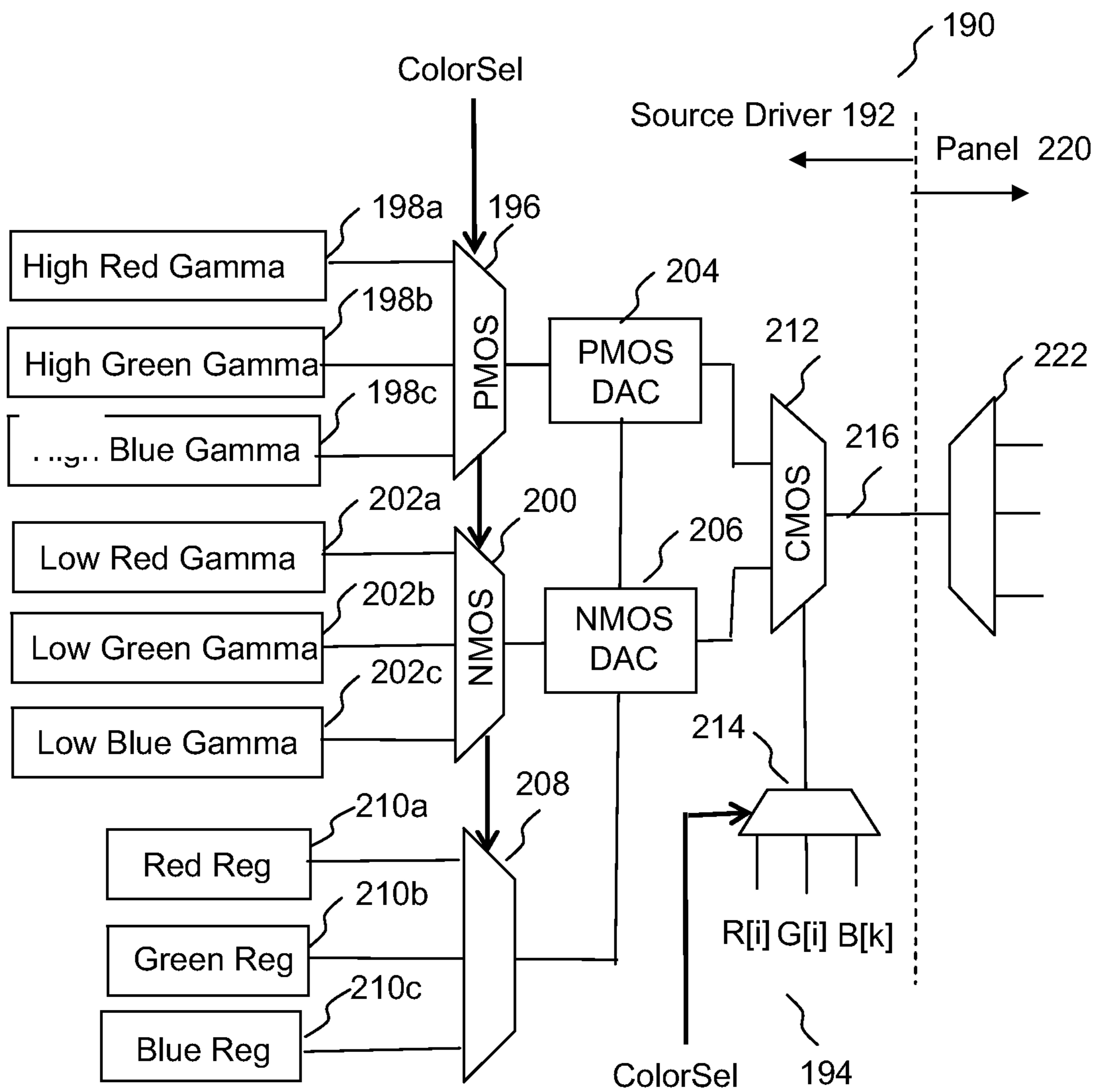


FIG. 7

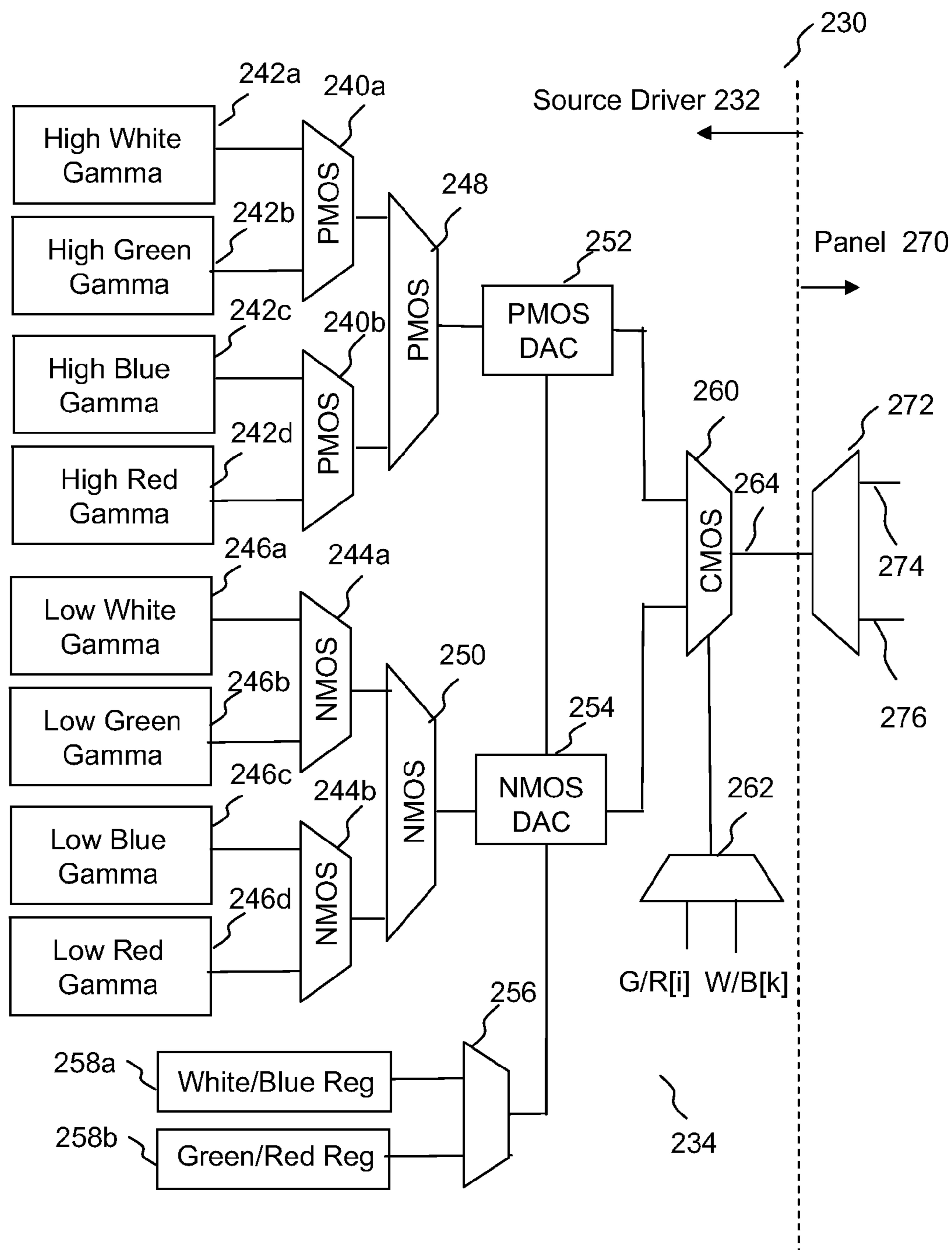


FIG. 8

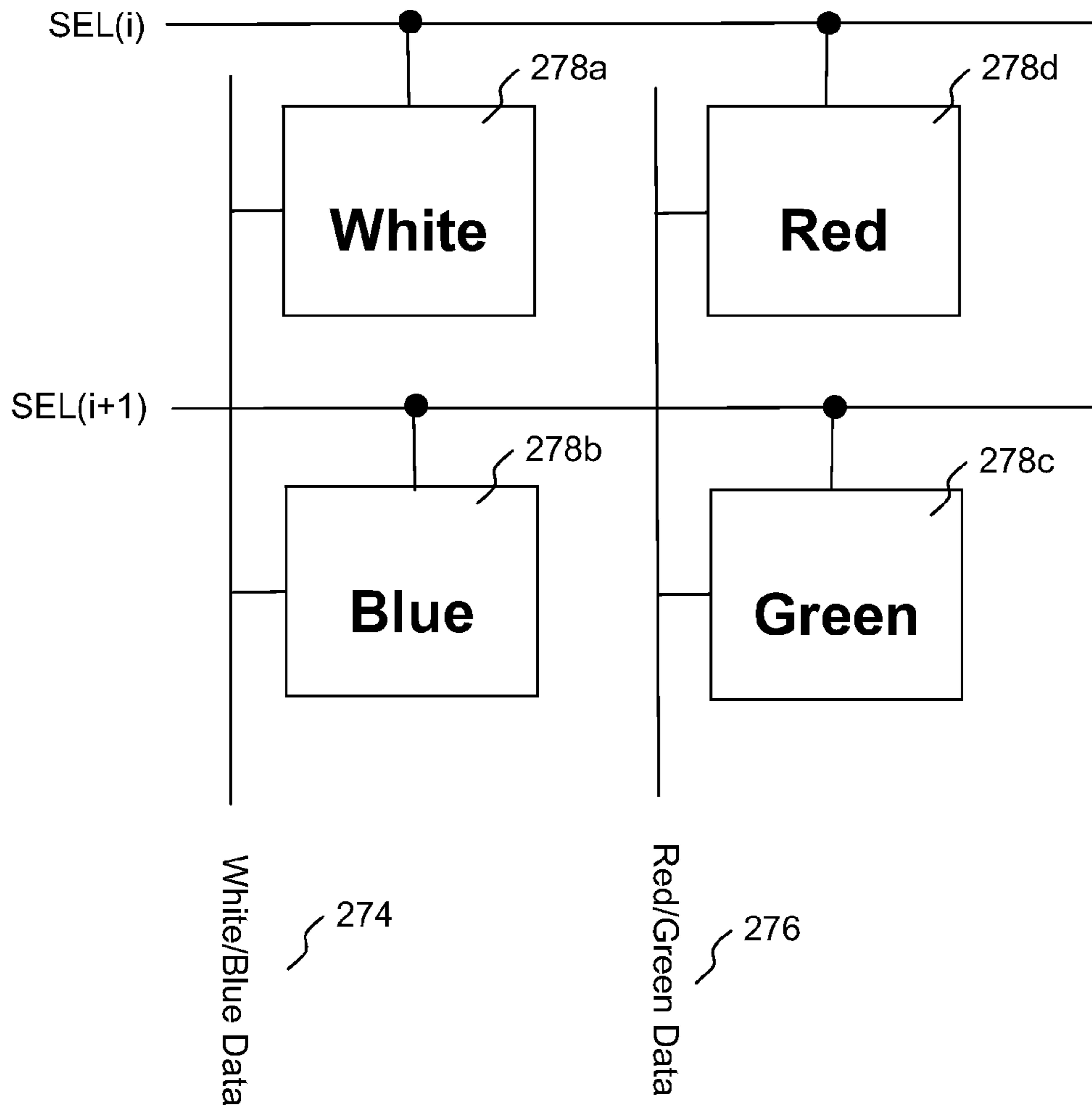


FIG. 9

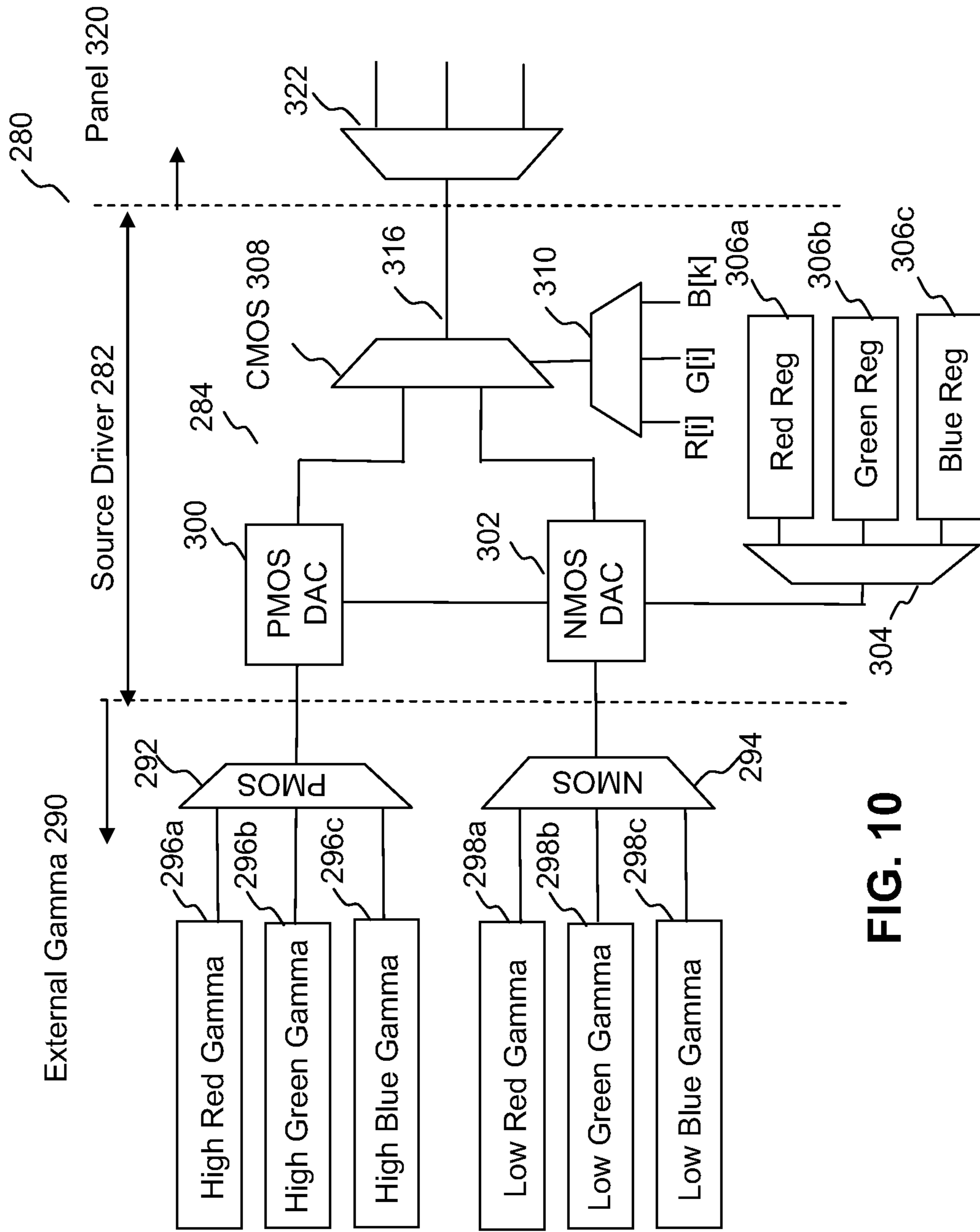


FIG. 10

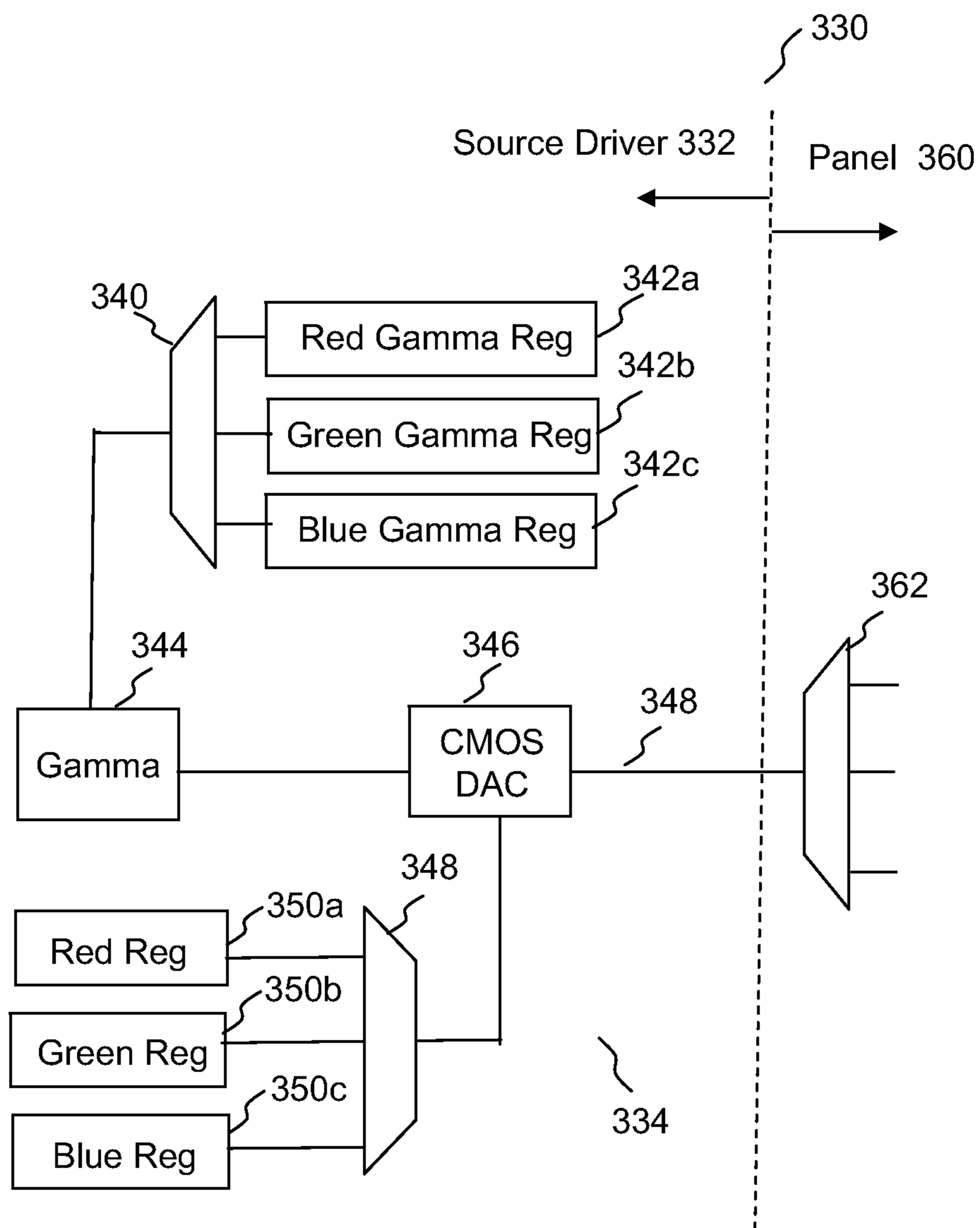


FIG. 11

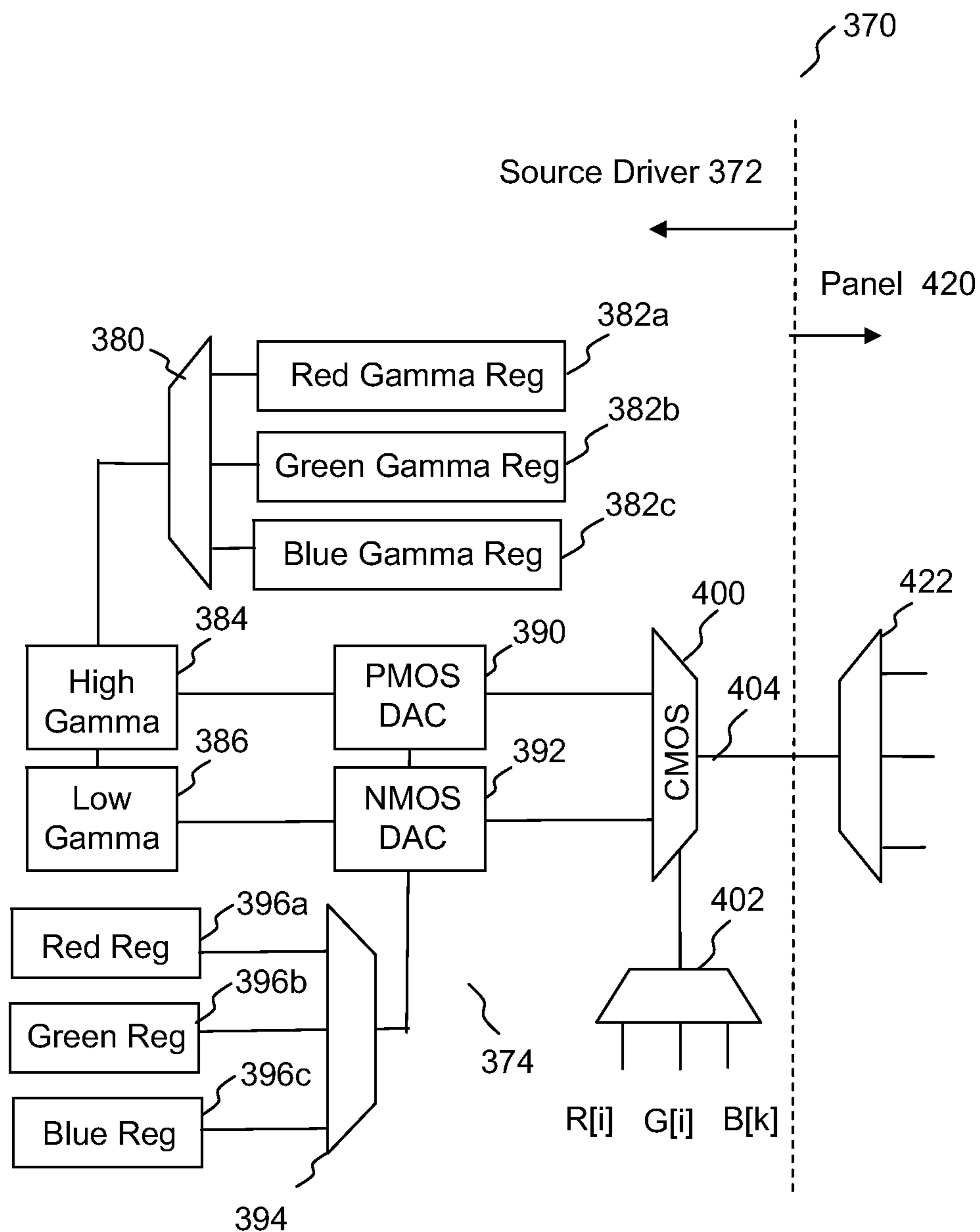


FIG. 12

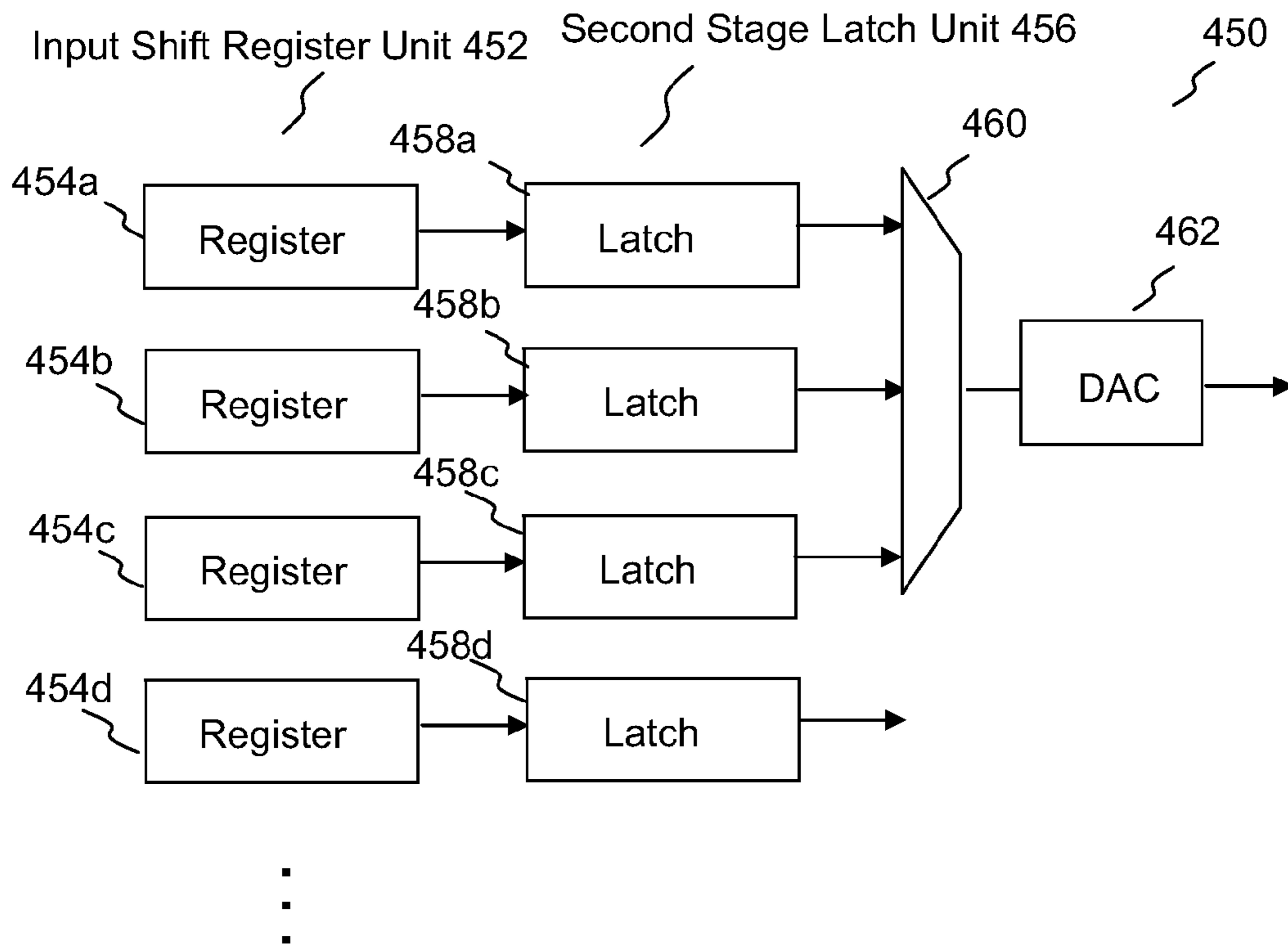


FIG. 13 (Prior Art)

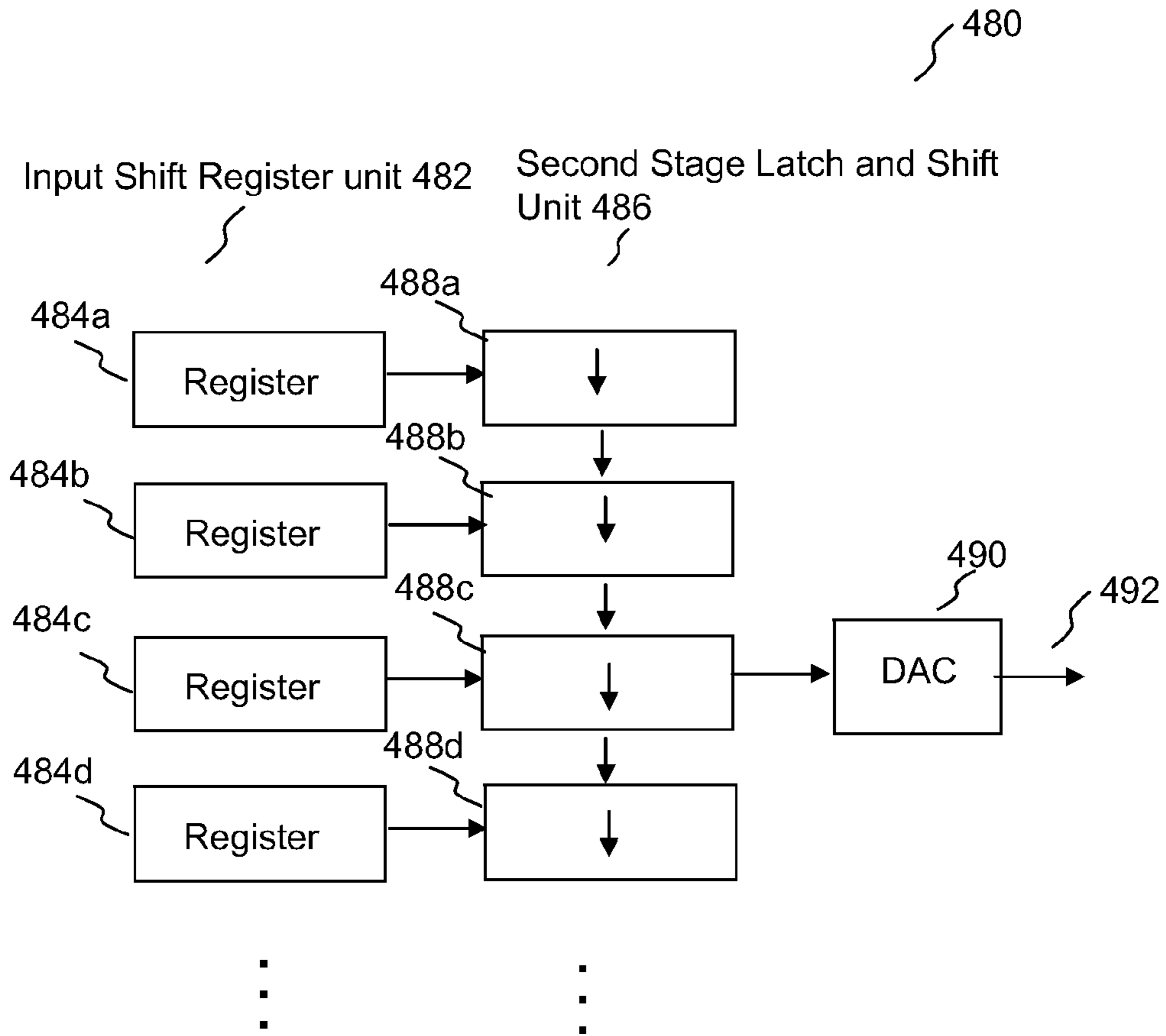


FIG. 14

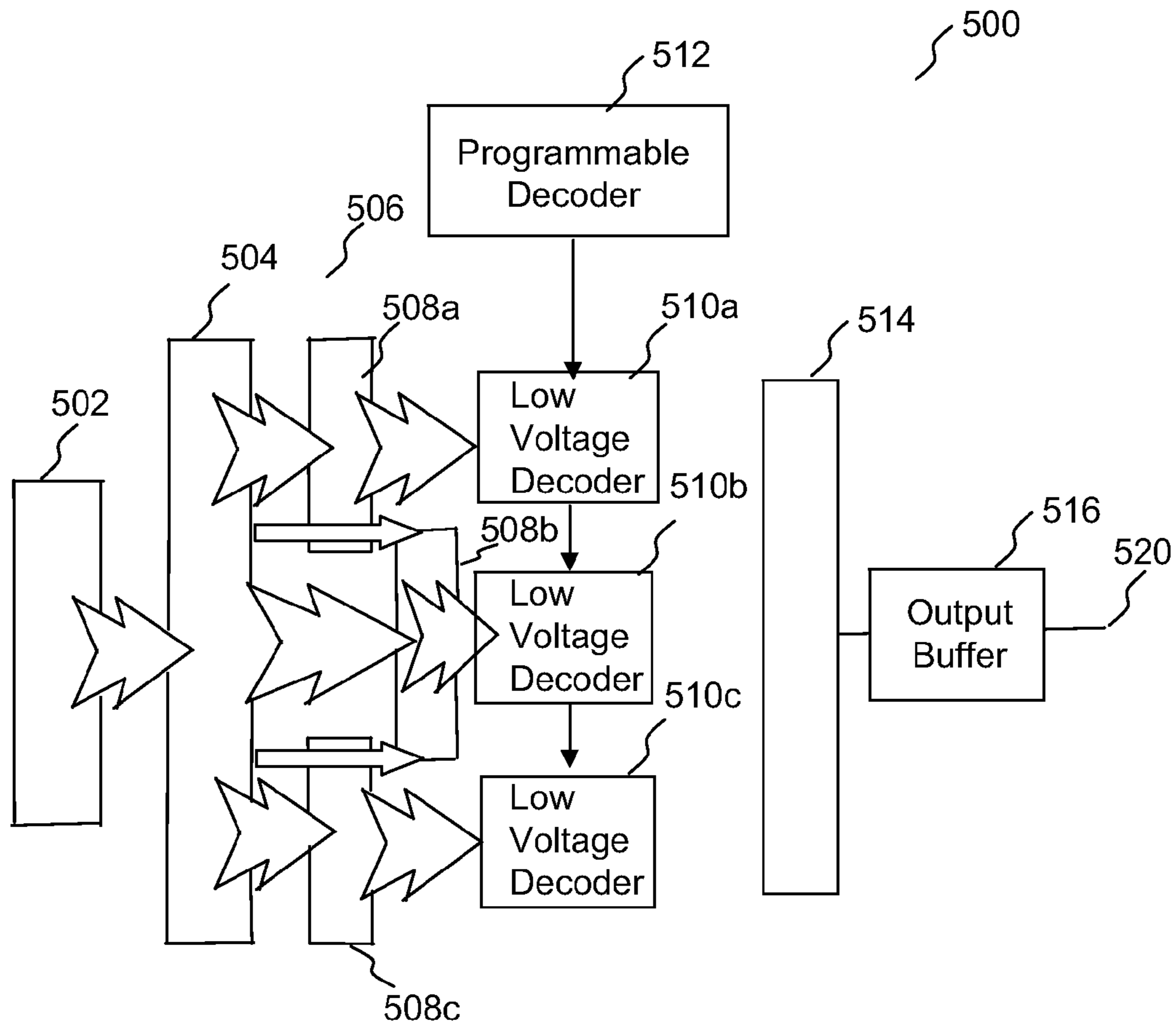


FIG. 15

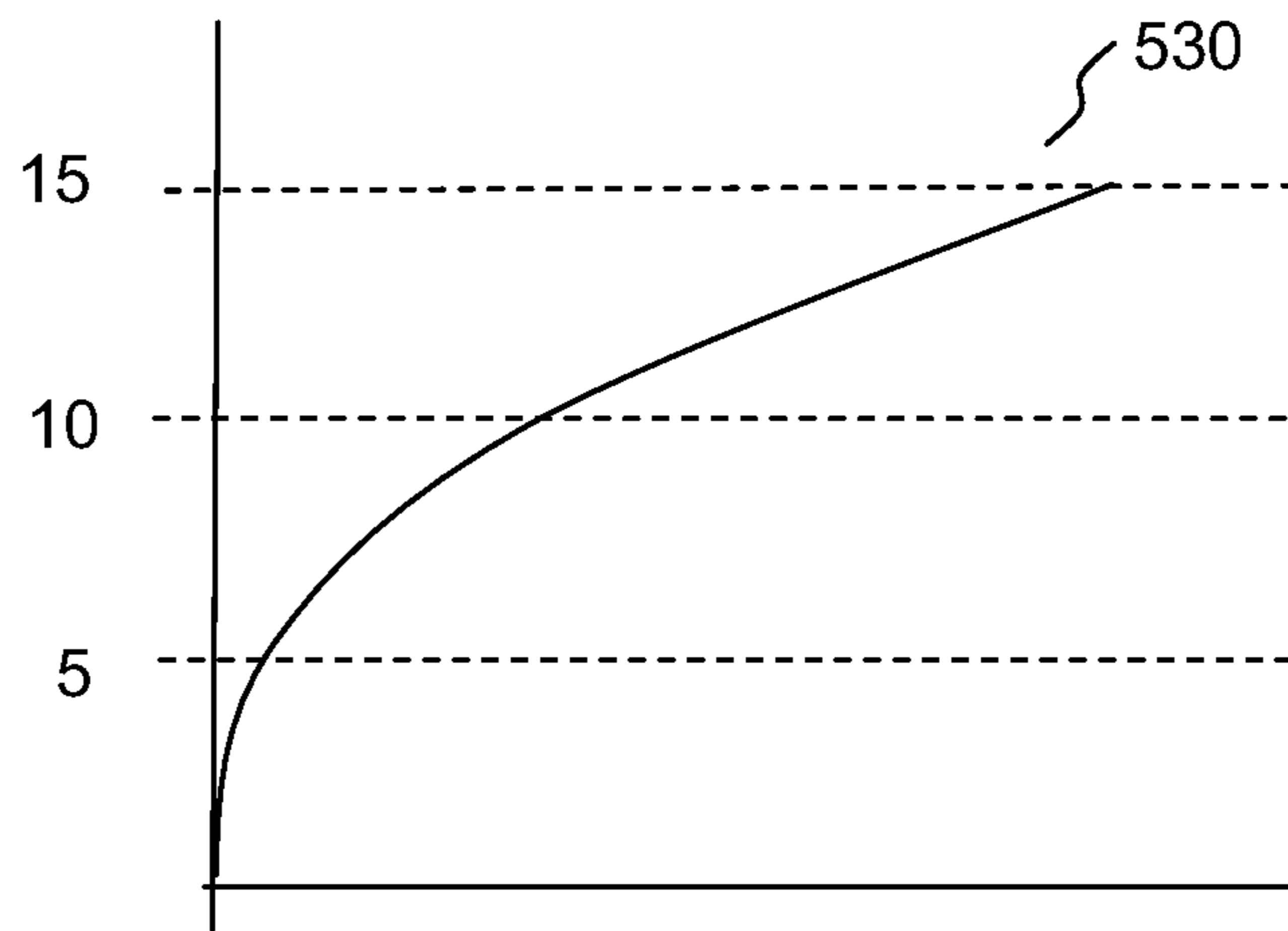


FIG. 16A

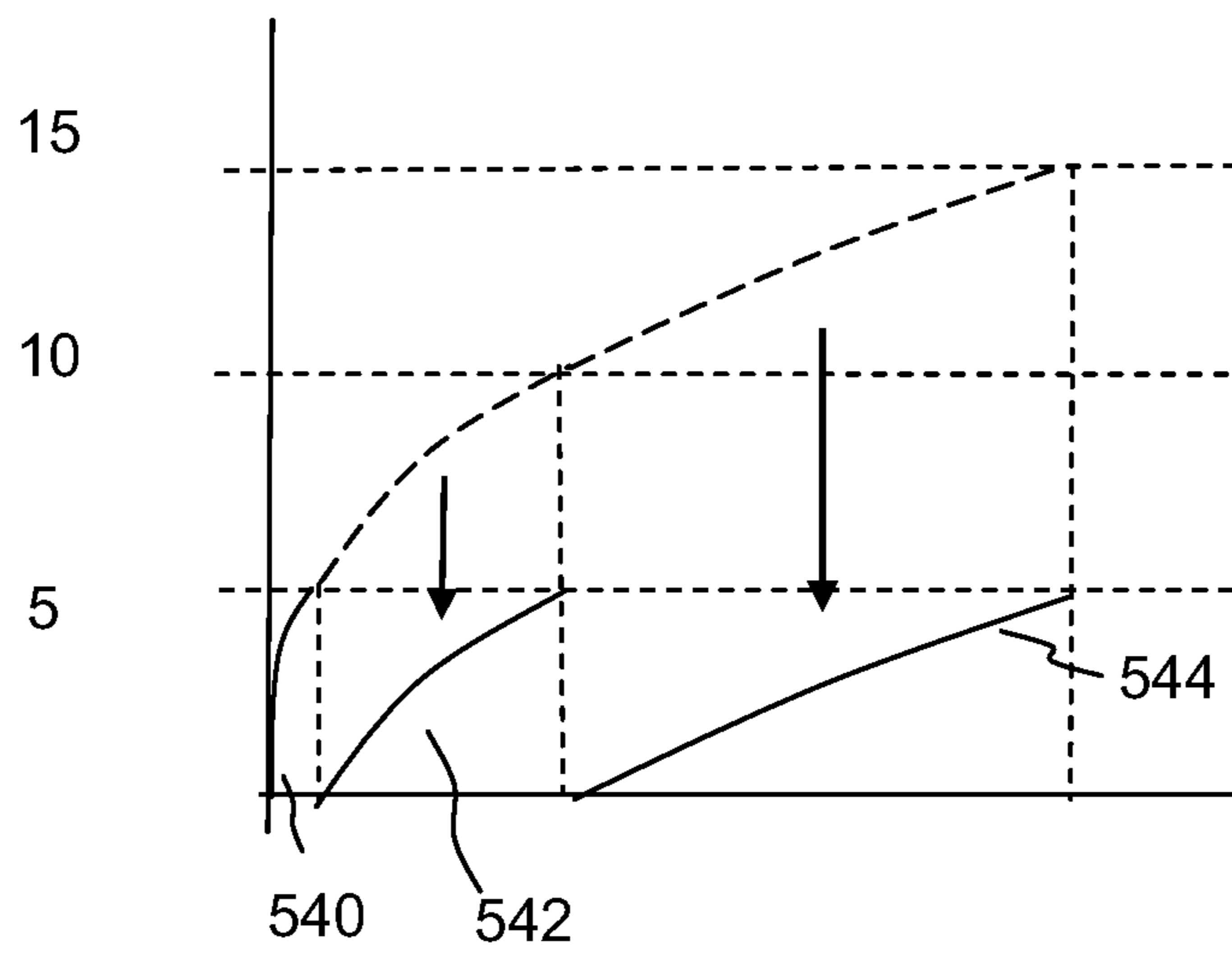


FIG. 16B

600

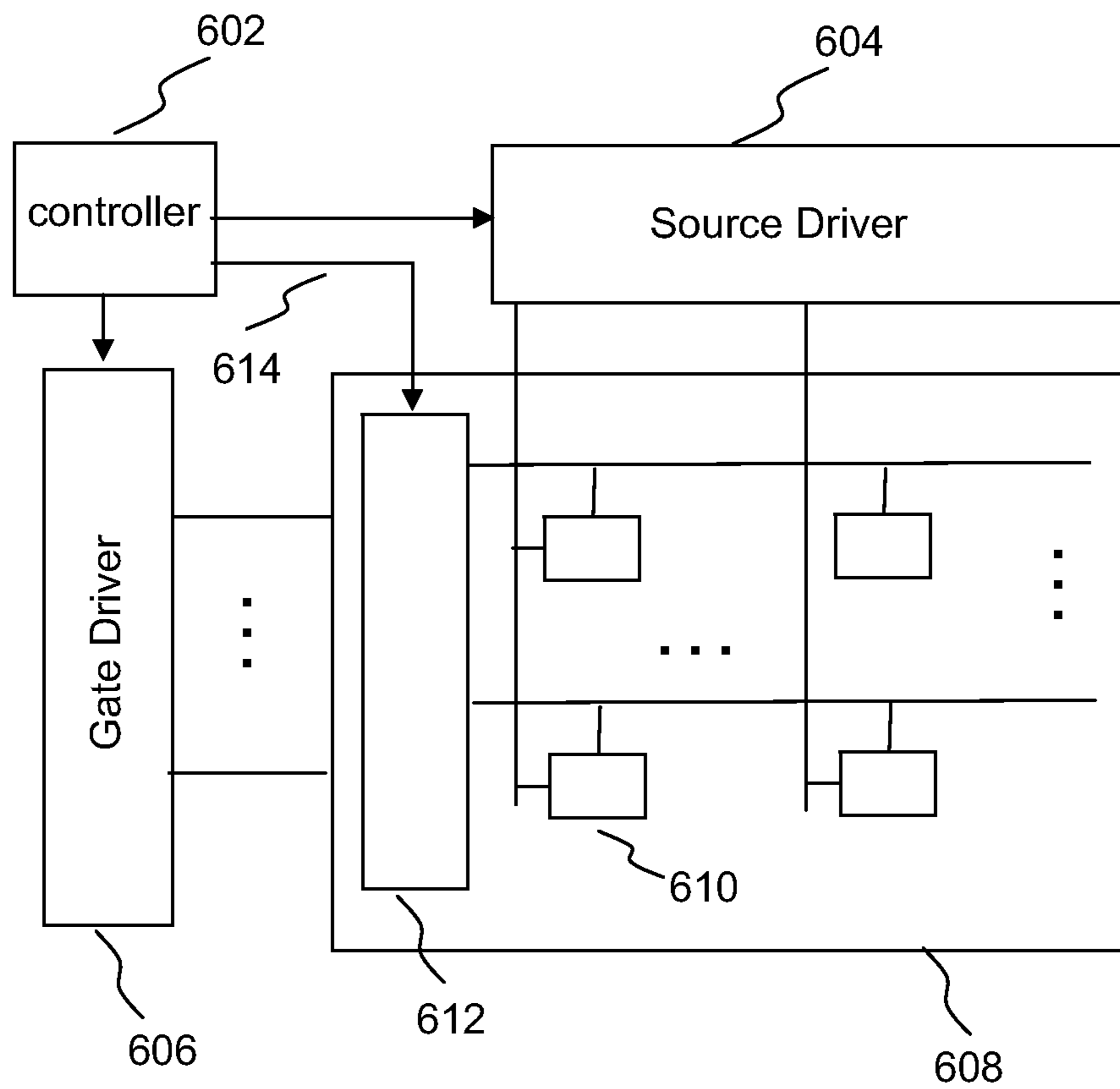


FIG. 17

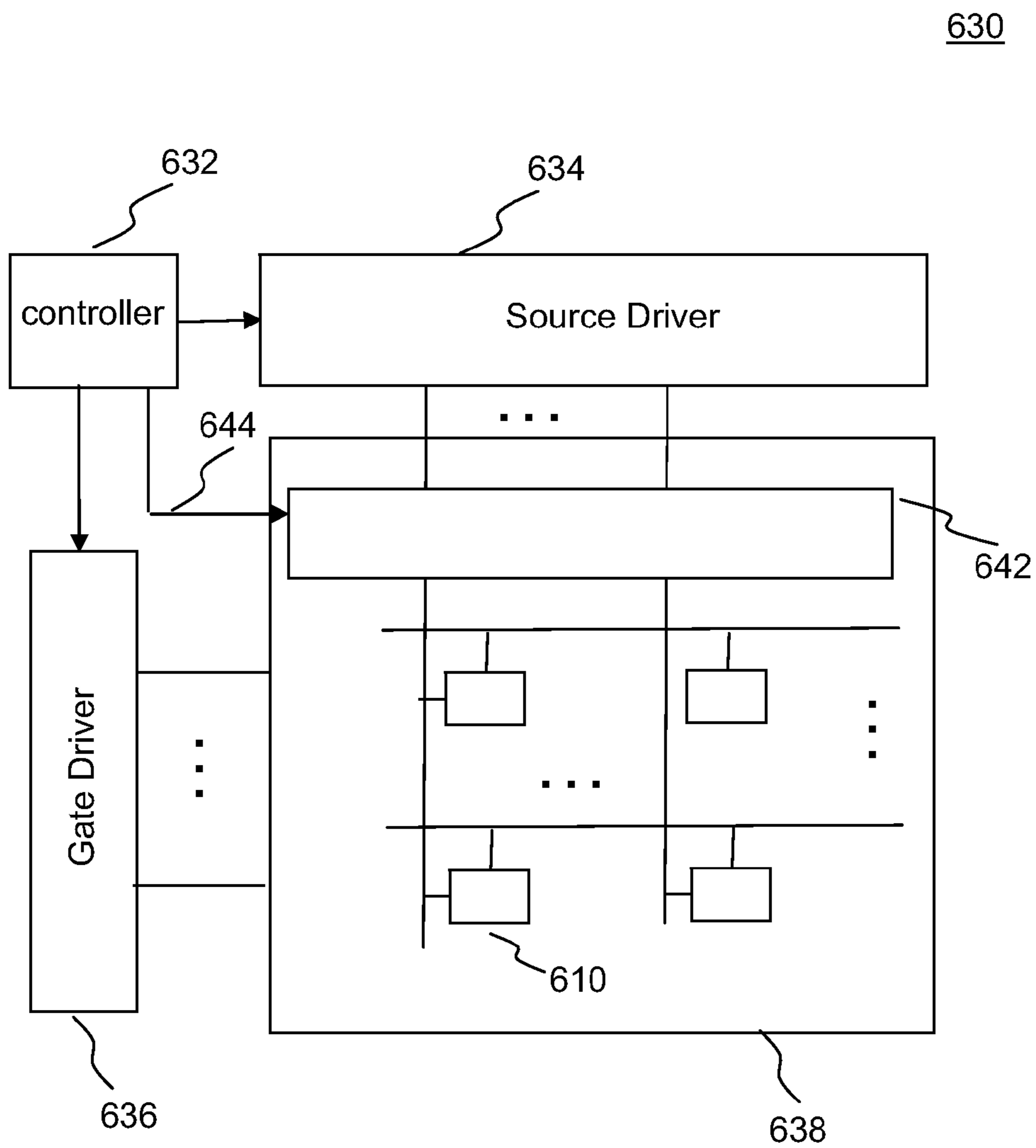


FIG. 18

660

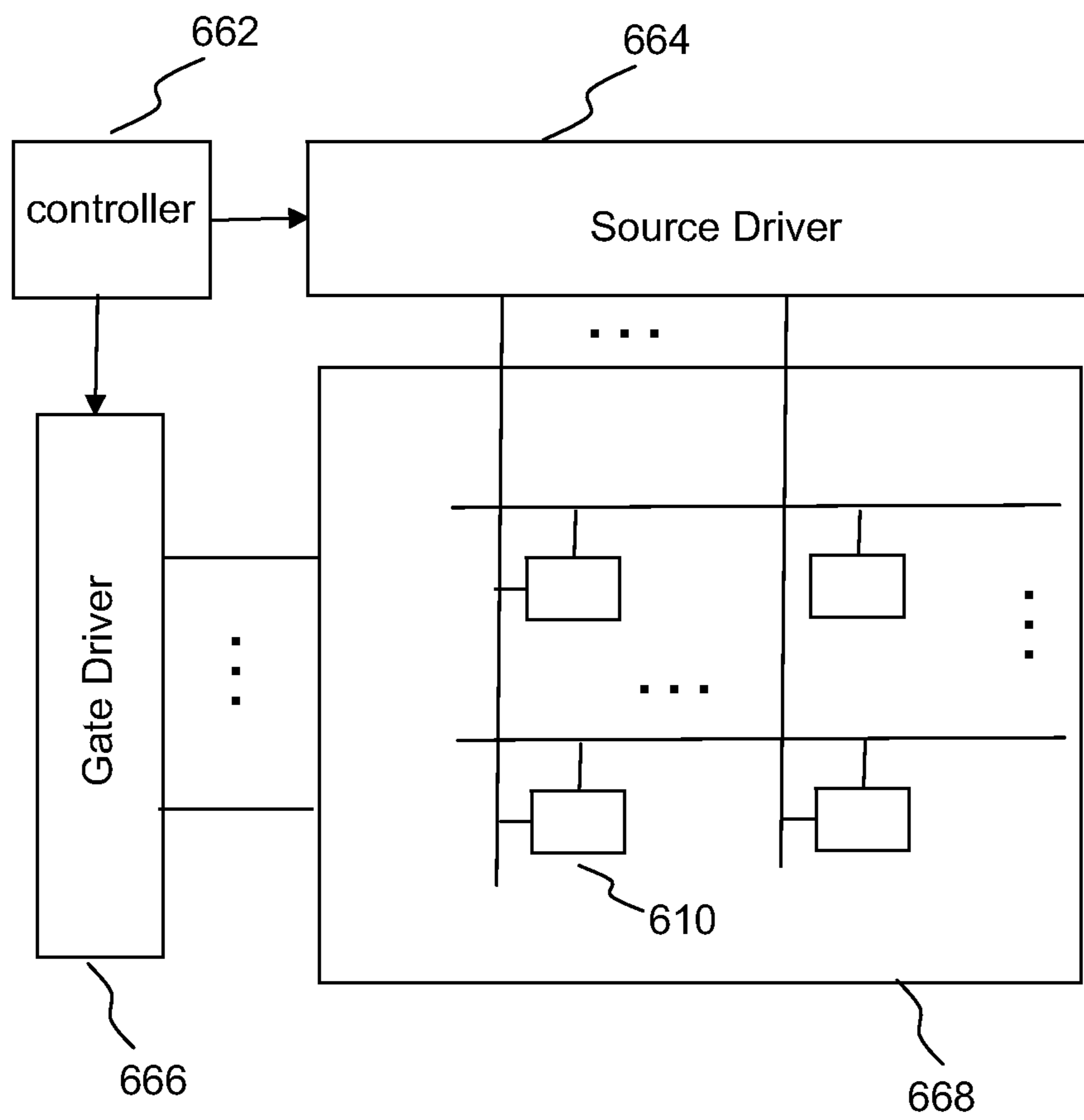


FIG. 19

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METHOD AND SYSTEM FOR DRIVING LIGHT EMITTING DISPLAY

Matter enclosed in heavy brackets [] appears in the original patent but forms no part of this reissue specification; matter printed in italics indicates the additions made by reissue; a claim printed with strikethrough indicates that the claim was canceled, disclaimed, or held invalid by a prior post-patent action or proceeding.

FIELD OF INVENTION

The present invention relates to a display system, more specifically to a method and system for driving light emitting displays.

BACKGROUND OF THE INVENTION

A display device having a plurality of pixels (or subpixels) arranged in a matrix has been widely used in various applications. Such a display device includes a panel having the pixels and peripheral circuits for controlling the panels. Typically, the pixels are defined by the intersections of scan lines and data lines, and the peripheral circuits include a gate driver for scanning the scan lines and a source driver for supplying image data to the data lines. The source driver may include gamma corrections for controlling gray scale of each pixel. In order to display a frame, the source driver and the gate driver respectively provide a data signal and a scan signal to the corresponding data line and the corresponding scan line. As a result, each pixel will display a predetermined brightness and color.

In recent years, the matrix display has been widely employed in small electronic devices, such as handheld devices, cellular phones, personal digital assistants (PDAs), and cameras. However, the conversional scheme and structure of the source driver and the gate driver demands the large number of elements (e.g., resistors, switchers, and operational amplifiers), resulting that the layout area of the peripheral circuits is still large and expensive.

Therefore there is a need to provide a display driver that can reduce a driver die area and thus cost, without reducing the driver performance.

SUMMARY OF THE INVENTION

It is an object of the invention to provide a method and system that obviates or mitigates at least one of the disadvantages of existing systems.

According to an embodiment of this disclosure, there is provided a display system, which includes: a driver for operating a panel having a plurality of pixels arranged by a plurality of first lines and at least one second line, the driver having: a driver output unit for providing to the panel a single driver output for activating the plurality of first lines, the single driver output being demultiplexed on the panel to activate each first line.

According to an embodiment of this disclosure, there is provided a display system, which includes: a driver for operating a panel having a plurality of pixels arranged by a plurality of data lines and at least one scan line, the driver having: a shift register unit including a plurality of shift registers; a latch and shift register unit including a plurality of latch and shift circuits for the plurality of shift registers, each storing an image signal from the corresponding shift register or shifting the image signal to a next latch and shift

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circuit; and a decoder unit including at least one decoder coupled to one of the latch and shift circuits, for decoding the image signal latched in the one of the latch and shift circuit to provide a driver output.

According to an embodiment of this disclosure, there is provided a display system, which includes: a driver for operating a panel having a plurality of pixels, the driver having: a plurality of multiplexers for a plurality of offset gamma curve sections, each offset gamma curve section having a first range less than a second range of a main gamma curve, at least one of offset gamma curve sections being offset by a predetermined voltage from a corresponding section of the main gamma curve; a plurality of decoders for the plurality of multiplexers; and an output buffer for providing a driver output based on the output from the decoder and the predetermined voltage.

BRIEF DESCRIPTION OF THE DRAWINGS

These and other features of the invention will become more apparent from the following description in which reference is made to the appended drawings wherein:

FIG. 1A illustrates a gate driver and a panel for a display system;

FIG. 1B illustrates an example of the gate driver and the panel of FIG. 1A;

FIG. 2 illustrates a timing chart for operating the display system of FIGS. 1A-1B;

FIG. 3A illustrates another example of a gate driver and a panel for a display system;

FIG. 3B illustrates an example of the gate driver and the panel of FIG. 3A;

FIG. 4 illustrates a timing chart for operating the display system of FIGS. 3A-3B;

FIG. 5 illustrates an example of a source driver and a panel for a display system;

FIG. 6 illustrates an example of operation for the display system having RGB pixel structure;

FIG. 7 illustrates a further example of a source driver and a panel for a display system;

FIG. 8 illustrates a further example of a source driver and a panel for a display system having RGBW pixel structure;

FIG. 9 illustrates an example of subpixel configuration for RGBW pixel structure;

FIG. 10 illustrates a further example of a source driver, external gamma and a panel for a display system;

FIG. 11 illustrates a further example of a source driver and a panel for a display system;

FIG. 12 illustrates a further example of a source driver and a panel for a display system;

FIG. 13 illustrates a source driver for a conventional display system;

FIG. 14 illustrates a further example of a source driver for a display system;

FIG. 15 illustrates a further example of a source driver for a display system;

FIG. 16 illustrates an example of a gamma curve and segmented offset gamma curves;

FIG. 17 illustrates an example of a display system having the gate driver of FIG. 1A or 3A;

FIG. 18 illustrates an example of a display system having the source driver of FIGS. 5-12; and

FIG. 19 illustrates an example of a display system having the source driver of FIGS. 14-15.

DETAILED DESCRIPTION

One or more currently preferred embodiments have been described by way of example. It will be apparent to persons

skilled in the art that a number of variations and modifications can be made without departing from the scope of the invention as defined in the claims.

Embodiments in this disclosure are described using a panel having pixels that are coupled to at least first line and at least one second line (e.g., scan lines and data lines) and being operated by a driver. The driver may be a driver IC having a plurality of pins, e.g., source driver ICs, gate driver ICs. The panel may be, for example, but not limited to, a LCD or LED panel. The panel may be a color panel or a monochrome panel.

In the description below, the terms “source driver” and “data driver” are used interchangeably, and the terms “gate driver” and “address driver” are used interchangeably. In the description below, the terms “row”, “scan line” and “address line” may be used interchangeably. In the description below, the terms “column”, “data line” and “source line” may be used interchangeably. In the description below, the terms “pixel” and “subpixel” may be used interchangeably.

Referring to FIGS. 1A-1B, there is illustrated a system **100** having a gate driver **102** and a panel **110** having pixels arranged in rows and columns. The system **100** includes a mechanism for multiplexing (muxing) gate driver outputs based on frequency reduction. In FIG. 1A, “fv” represents the vertical frequency of the display (or row frequency), and “M” is the number of muxing blocks. In FIG. 1B, “Cell #i” represents an address cell **106**, and “SEL k” ($k=(i-1)*M+1, (i-1)*M+2, \dots, (i-1)*M+M+1, i*M$) represents a row or a scan line coupled to the row of the panel **110**. A pixel in the row is selected by the scan line. The address cell **106** may be a logic or a flip-flop in a shift register chain to output a gate output.

The gate driver **102** includes a driver output unit **104** having at least one address cell **106** (Cell #i). The address cell **106** provides a single gate driver output **108** which is shared by M rows. An individual gate driver output **108** from the gate driver **102** is active for M rows. On the panel side **110**, a demultiplexer **112** (“1:M Demuxs” in FIG. 1A) is employed for M rows. The input of the demultiplexer **112** is coupled to the gate driver output **108**, and the outputs of the demultiplexer **112** are coupled to M rows. In this example, the demultiplexer **112** is coupled to scan lines SEL $(i-1)*M+1, SEL (i-1)*M+2, \dots, \text{and } SEL i*M$. The activated gate driver output **108** from the address cell **106** (Cell #i) is assigned to each individual row in sequence, via the demultiplexer **112**.

The demultiplexer **112** is implemented using, for example, thin film transistors, on the panel **110**. The demultiplexer **112** includes a plurality of switch blocks for activating M rows. In FIG. 1B, switches **116** (SET #1, SET #2, . . . , SET #M) are shown as an example of the components of the demultiplexer **112**. The switch block **116** (SET #k: $k=1, 2, \dots, M$) is employed for the scan line SEL $(i-1)*M+k$. Each switch block **116** includes a pair of switches, one being capable of connecting the gate driver output **108** to the corresponding scan line and the other being capable of connecting VGL to the corresponding scan line. VGL may be a ground level voltage. Each scan line SEL $(i-1)*M+k$ turns to be on the VGL level or the activated gate driver output **108** via the corresponding switch block **116** (SET #k). Each switch block **116** (SET #k) is controlled by the corresponding control signal CTRL (k). In FIG. 3B, the scan line SEL $(i-1)*M+k$ is selected (becomes active) by the control signal CTRL (k). By operating the demultiplexer **112** with the control signals CTRL (1)-CTRL (M), the number of the gate driver outputs and address cells is reduced by a factor of M.

In FIG. 1B, one address cell **116** is shown as an element of the driver output unit **104**; however, the number of the address cells may vary. In FIG. 1B, M rows (scan lines) are shown; however, the panel **110** may include a plurality of groups of rows where the ith group has M rows and is operated by the ith address cell (Cell #i). One of ordinary skill in the art would appreciate that the gate driver **102** and the panel **110** may include components not shown in the FIGS. 1A-1B.

Referring to FIGS. 1A, 1B and 2, the operation of a display having the gate driver **102** and the panel **110** is described. Each of the controlling signals CTRL (1)-CTRL (M) for controlling the demultiplexing on the panel **110** works at the normal gate frequency. When the display programming reaches the row SEL $(i-1)*M+1$, the control signal CTRL (1) for that row is high, resulting that the address cell **106** for the ith block (Cell #i) of rows is connected to SEL $(i-1)*M+1$. Thus, that row SEL $(i-1)*M+1$ is selected and the image data can be written in the pixels of the row.

After the programming of the row SEL $(i-1)*M+1$, the next control signal CTRL (2) is high, resulting that the next row SEL $(i-1)*M+2$ becomes active. This continues till the entire display is programmed (end of a frame).

If a row is not active, the control signal related to that row is low or the address cell related to that row is not active. Thus, the row is connected to VGL which will disconnect the pixels in that row from the gate driver **102**.

Referring to FIGS. 3A-3B, there is illustrated a system **130** having a gate driver **132** and a panel **140** having pixels arranged in rows and columns. The system **130** has a mechanism for reducing the number of gate driver outputs and reducing the operation frequency of demultiplexing control signals on the panel side. In FIG. 3A, “fv” represents the vertical frequency of the display (or row frequency). In FIG. 3B, “Cell #j” ($j=i, i+1, i+2, i+3$) represents an address cell, and “SEL k” ($k=i, i+1, i+2, i+3$) represents a row or a scan line coupled to the row of the panel **140**. A pixel in the row is selected by the scan line. The address cell may be a logic or a flip-flop in a shift register chain to output a gate output.

In the system **130**, gate driver output signals are multiplexed on the gate driver **132** side, and the outputs from the gate driver **132** are demultiplexed on the panel **140** side.

The gate driver **132** includes a driver output unit **133** having a plurality of multiplexers for a plurality of address cells. Each address cell provides a gate driver signal, and each multiplexer multiplexing the gate driver signals and outputs a single gate driver output. In FIG. 3B, four address cells **138a-138d** (Cell #i, Cell #i+1, Cell #i+2, and Cell #i+3) are shown as an example of the address cells in the gate driver **132**. In FIG. 3B, two multiplexers **134a** and **134b** are shown as an example of multiplexing the gate driver signals. The multiplexers **134a** and **134b** are controlled by a control signal iCTRL. The multiplexer **134a** is coupled to the address cells **138a** and **138c** (Cell #i and Cell #i+2) and outputs a gate output signal **136a** that corresponds to either address cell **138a** or **138c** (Cell #i or Cell #i+2). The multiplexer **134b** is coupled to the address cells **138b** and **138d** (Cell #i+1 and Cell #i+3) and outputs a gate output signal **136b** that corresponds to either address cell **138b** or **138d** (Cell #i+1 or Cell #i+3).

The panel **140** includes a multiplexer **142** (“1:M Demuxs” in FIG. 3A) coupling to the gate driver outputs and a plurality of rows. The demultiplexer **142** is implemented using, for example, thin film transistors, on the panel **140**. The demultiplexer **142** includes a plurality of switch group

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blocks, each coupling to the gate driver multiplexers. In FIG. 3B, two switch group blocks 146a and 146b (SET #1 and SET #2) are shown as an example of the components of the demultiplexer 142. On the panel side 140, the activated gate driver outputs 136a and 136b are assigned of the switch group blocks 146a and 146b.

Each switch group block in the panel 140 includes a plurality of switch blocks 148. In FIG. 3B, each of the switch group blocks 146a and 146b includes two switch blocks 148, one being capable of coupling one gate driver output 136a to one scan line and the other being capable of coupling the other gate driver output 136b to the other scan line. The switch block 148 includes a pair of switches, one being capable of coupling the gate driver output to the corresponding scan line and the other being capable of coupling VGL to the corresponding scan line. VGL may be a ground level voltage. The switch block 148 in the switch group block (SET #k: k=1, 2, . . .) is controlled by the corresponding control signal CTRL (k). Each scan line turns to be on the VGL level or the corresponding activated gate driver output 136a or 136b via the corresponding switch block 148. In FIG. 3B, the scan lines SEL (i) and SEL (i+1) are selected (become active) by the control signal CTRL (1), and the scan lines SEL (i+2) and SEL (i+3) are selected (become active) by the control signal CTRL (2).

In FIG. 3B, the multiplexing (muxing) and demultiplexing (demuxing) operations are executed for two rows, however, the multiplexing and demultiplexing operations may be executed for more than two rows. In FIG. 3B, four address cells are shown as an element of the driver output unit 133; however, the number of the address cells is not limited to four and may vary. In FIG. 3B, rows (scan lines) are divided into two groups, each having two rows; however, the number of groups and the number of rows in each group are not limited to two and may vary. One of ordinary skill in the art would appreciate that the gate driver 132 and the panel 140 may include components not shown in the FIGS. 3A-3B.

In this structure, the physical multiplexing is used at the gate driver side 132. As a result, the number of address cells remains the same while the number of gate driver outputs is reduced by a factor of multiplexing blocks. The number of rows in each set (SET #k) can be increased for further reduction in output of the gate driver and the frequency of the control signals. Since multiple gate driver outputs can be active, the operation frequency of the demultiplexing control signals is reduced.

Referring to FIGS. 3A, 3B and 4, the operation of a display having the gate driver 132 and the panel 140 is described. When the display programming reaches the rows SEL (i) and SEL (i+1), the control signal CTRL (1) for those rows is high (150), resulting that the gate driver output 136a is coupled to the row SEL (i) and the gate driver output 136b is coupled to the row SEL (i+1). At this period (150), the control signal iCTRL is in one state (e.g., low). The gate driver output 136a corresponds to the output from the address cell 138a (Cell #i) and the gate driver output 136b corresponds to the output from the address cell 138b (Cell #i+1). The image data can be written in the pixels of the selected rows SEL (i) and SEL (i+1).

After the programming of the rows SEL (i) and SEL (i+1), the next control signal CTRL (2) is high (152), resulting that the next rows SEL (i+2) and SEL (i+3) become active. At this period (152), the control signal iCTRL is in the other state (e.g., high). The gate driver output 136a corresponds to the output from the address cell 138c (Cell #i+2) and the gate driver output 136b corresponds to the output from the address cell 138d (Cell #i+3). The image data can be written

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in the pixels of the selected rows SEL (i+2) and SEL (i+3). This continues till the entire display is programmed (end of a frame).

If a row is not active, the control signal related to that row is low or the address cell related to that row is not active. Thus, the row is connected to VGL which will disconnect the pixels in that row from the gate driver 132.

Referring to FIG. 5, there is illustrated a system 160 having a source driver 162 and a panel 180 having subpixels for RGB. Most of light emitting displays employ different gammas (or gamma corrections) for different subpixels, which use different decoders for different outputs. In the system 160, gammas (gamma corrections, gamma voltages) are multiplexed on the source driver 162 side. In the description, the terms "gamma", "gamma correction" and "gamma voltages" may be used interchangeably. One of ordinary skill in the art would appreciate that the source driver 162 and the panel 180 may include components not shown in FIG. 5.

The source driver 162 includes a driver output unit 164 having a CMOS multiplexer 166 and a CMOS digital to analog converter (DAC) 170. The multiplexer 166 multiplexes a Red gamma correction 168a, a Green gamma correction 168b and a Blue gamma correction 168c. The DAC 170 includes a decoder. In the description, the terms "DAC" and "DAC decoder" may be used interchangeably.

Each of the gamma corrections 168a, 168b and 168c provides a reference voltage to the DAC 170. The reference voltage is selected based on the dynamic range of the DAC decoder 170. The reference voltage at the gamma correction block may be generated using, for example, resistors, or be stored using, for example, registers.

The output from the multiplexer 166 is provided to the DAC 170. The multiple gammas share one decoder in the DAC 170. The DAC decoder 170 operates on an output from a multiplexer 172. The multiplexer 172 multiplexes a Red register (reg) 174a for storing image data for Red, a Green register (reg) 174b for storing image data for Green, and a Blue register (reg) 174c for storing image data for Blue. The CMOS DAC 170 provides a single source driver output 174.

A demultiplexer 182 is employed on the panel 180 side to demultiplex the driver output 174 from the source driver 162. The demultiplexer 182 is implemented using, for example, thin film transistors, on the panel 180. The outputs from the demultiplexer 182 are couples to three data lines. The driver output 174 is demultiplexed 182 on the panel 180 side and goes to different subpixels (i.e., Red subpixel, Blue subpixel and Green subpixel).

In the system 160, the output of the source driver 162 is multiplexed to reduce the number of driver pins and demultiplexed at the panel 180. To further improve the size of the driver area, the multiplexing is executed at few stage earlier at the gamma selection and DAC inputs. For example, when, the Red pixels are being programmed at the panel 180, the Red data (Red register 174a) and the red gamma 168a are assigned to the DAC 170.

The multiplexers 166 and 172 may be controlled by a color selection control signal ColorSel. The demultiplexer 182 may be controlled by the control signal ColorSel or a control signal associated with the multiplexing control signal ColorSel.

As shown in FIG. 6, the Red pixels, Green pixels and Blue pixels may be programmed sequentially. It will be appreciated by one of ordinary skill in the art that the programming sequence is not limited to that of FIG. 6, and is changeable by using the color selection control signal.

Generally, the output range of the voltage required for the light emitting displays is high and thus source drivers are to be a rail-to-rail design for the power. Currently, this results in using multiple CMOS decoders, leading to a larger area source driver. Referring to FIG. 7, there is illustrated a system 190 having a source driver 192 and a panel 220 having subpixels for RGB. In this system 190, multiple gammas (gamma corrections, gamma voltages) are multiplexed and a DAC is divided into separate NMOS and PMOS components, resulting in that the source driver 192 area is reduced. One of ordinary skill in the art would appreciate that the source driver 192 and the panel 220 may include components not shown in FIG. 7.

The source driver 192 includes gamma corrections for Red, Blue and Green, each providing a reference voltage to a DAC decoder. The reference voltage is selected based on the dynamic range of the decoder. The reference voltage may be generated using, for example, resistors, or be stored using, for example, registers. Each gamma correction has a high voltage level gamma correction (high voltage level of gamma corrections) and a low voltage level gamma correction (low voltage level of gamma corrections). The high voltage level of gamma corrections is a level from a predefined reference voltage to the high point of the driver output, and the low voltage level of gamma corrections is a level from the predetermined reference voltage to the beginning of the gamma voltage. The predetermined reference voltage may be at the middle for the driver output range. For example, if the driver range is 10V, the predetermined reference voltage is 5V; the high voltage level of gamma corrections is 5 to 10V; and the low voltage level of gamma corrections is 0 to 5V.

The source driver 192 includes a driver output unit 194 having a PMOS multiplexer 196 for the high voltage level of gamma corrections, and a NMOS multiplexer 200 for the low voltage level of gamma corrections. In FIG. 7, the multiplexer 196 multiplexes a high Red gamma correction 198a, a high Green gamma correction 198b and a high Blue gamma correction 198c, and the multiplexer 200 multiplexes a low Red gamma correction 202a, a low Green gamma correction 202b and a low Blue gamma correction 202c.

The driver output unit 194 includes a DAC that is divided into separate components: a PMOS component 204 ("PMOS DAC" in FIG. 7) and a NMOS component 206 ("NMOS DAC" in FIG. 7). The PMOS component 204 includes a PMOS decoder and receives the output from the multiplexer 196. The NMOS component 206 includes a NMOS decoder and receives the output from the multiplexer 200. The reference voltage from the gamma correction is selected based on the dynamic range of the NMOS and PMOS decoders in the components 204 and 206. The PMOS and NMOS decoders in the components 204 and 206 operate on an output from a multiplexer 208 for multiplexing a Red register 210a, a Green register 210b, and a Blue register 210c. The registers 210a, 210b and 210c correspond to the resistors 174a, 174b and 174c of FIG. 5, respectively. The multiplexers 196, 200 and 208 are controlled by a color selection control signal ColorSel.

The driver output unit 194 includes a CMOS multiplexer 212 for multiplexing the outputs from the PMOS and NMOS components 204 and 206. The multiplexer 212 is operated by an output from a multiplexer 214. The multiplexer 214 multiplexes bit signals R[j], G[i], and B[k], based on the color selection control signal ColorSel. R[j] (G[i], B[k]) is a bit that defines when to use which part of the gamma for Red (Green, Blue). The bit R[j] (G[i], B[k]) is generated based on

the Red register 210a (210b, 210c) and predefined data about the gamma curve for Red (Green, Blue), e.g., gamma values. The multiplexer 212 outputs a single source driver output 216.

When the bit signal R[j] is active and the other signals are not active, the source driver 192 outputs the driver output 216 based on either the high Red gamma correction or the low Red gamma correction.

A demultiplexer 222 is employed on the panel 220 side to demultiplex the source driver output 216. The demultiplexer 222 corresponds to the demultiplexer 182 of FIG. 5. The demultiplexer 222 is implemented using, for example, thin film transistors, on the panel 220. The outputs from the demultiplexer 222 are couples to three data lines. The demultiplexer 222 may be controlled by the control signal ColorSel or a control signal associated with the multiplexing control signal ColorSel. Based on the output from the demultiplexer 222, one of three data lines is active. The driver output 216 is demultiplexed 222 on the panel 220 side and goes to different subpixels (i.e., Red subpixel, Blue subpixel, Green subpixel).

Based on the image data, one of the low gamma correction and the high gamma correction is selected. For example, if the high voltage level of gamma corrections is 5 to 10V, the low voltage level of gamma corrections is 0 to 5V, and the image data requires 6 V, the high end of gamma correction will be selected.

Based on the color selection control signal ColorSel, the Red pixels, Green pixels and Blue pixels may be programmed sequentially, similar to that of FIG. 6. It will be appreciated by one of ordinary skill in the art that the programming sequence is not limited to that of FIG. 6, and is changeable by using the color selection control signal.

Instead of using a CMOS decoder that has twice as many transistors as a PMOS or NMOS decoder for the entire range the output voltage, the PMOS decoder 204 is used for the higher range and the NOMS decoder 206 for the lower range of the voltage. Thus, the area will be reduced by using twice less transistors.

Referring to FIG. 8, there is illustrated a system 230 having a source driver 232 and a panel 270 having subpixels. The system 230 is applied to quad RGBW pixel structure. Multiple gamma corrections for White, Green, Blue and Red are multiplexed in the source driver 232. In the source driver 232, four different gamma corrections are generated (White, Green Blue and Low) for each of high voltage level and low voltage level. One of ordinary skill in the art would appreciate that the source driver 232 and the panel 270 may include components not shown in FIG. 8.

The source driver 232 includes gamma corrections for White, Green, Blue and Red, each providing a reference voltage to a DAC decoder. The gamma correction may be generated using, for example, resistors, or be stored using, for example, registers. Each gamma correction has a high voltage level gamma correction (high voltage level of gamma corrections) and a low voltage level gamma correction (low voltage level of gamma corrections). As described above, the high voltage level of gamma corrections is a level from the reference voltage to the reference voltage to the high point of the driver output, and the low voltage level of gamma corrections is a level from the reference voltage to the beginning of the gamma voltage.

The source driver 232 includes a driver output unit 270 having PMOS multiplexers 240a and 240b for high voltage level of gamma corrections, and NMOS multiplexers 244a and 244b for low voltage level of gamma corrections. The multiplexer 240a multiplexes a high White gamma correc-

tion 242a and a high Green gamma correction 242b, and the multiplexer 240b multiplexes a high Blue gamma correction 242c and a high RED gamma correction 242d. The multiplexer 244a multiplexes a low White gamma correction 246a and a low Green gamma correction 246b, and the multiplexer 244b multiplexes a low Blue gamma correction 246c and a low RED gamma correction 246d.

The driver output unit 270 includes a PMOS multiplexer 248 for multiplexing the outputs from the PMOS multiplexers 240a and 240b, and a NMOS multiplexer 250 for multiplexing the outputs from the NMOS multiplexers 244a and 244b. Based on the image data and a color selection, one of the low gamma correction and the high gamma correction for the selected color is selected.

The driver output unit 270 includes a DAC that is divided into separate components; a PMOS component 252 ("PMOS DAC" in FIG. 8) for the high voltage level of the gamma corrections and a NMOS component 254 ("NMOS DAC" in FIG. 8) for the low voltage level of the gamma corrections. The PMOS component 252 includes a PMOS decoder and receives the output from the multiplexer 248. The NMOS component 254 includes a NMOS decoder and receives the output from the multiplexer 250. The reference voltage from the gamma correction is selected based on the dynamic range of the NMOS and PMOS decoders in the components 252 and 254.

The PMOS and NMOS decoders in the components 252 and 254 operate on an output from a multiplexer 256 for multiplexing a White/Blue register 258a and a Green/Red register 258b. The White/Blue register 258a stores image data for White/Blue. The Green/Red register 258b stores image data for Green/Red. In the RGBW structure, each data line carries data for two different colors. In this example, one data line carries data for White and Blue, and the other data line carries data for Green and Red. In one row, a data line is connected, for example, to White pixels (Green pixels) while during the next row it is connected to Blue pixels (Red pixels). As a result, the register 258a used for White and Blue data is shared, and the register 258b used for Green and Red is shared.

The driver output unit 270 includes a CMOS multiplexer 260 for multiplexing the outputs from the PMOS and NMOS decoders in the components 252 and 254. The multiplexer 260 is operated by a multiplexer 262 for multiplexing bit signals $G/R[i]$ and $W/B[k]$. $W/B[k]$ ($G/R[j]$) is a bit that defines when to use which part of the gamma for White or Blue (Green or Red). The bit $W/B[k]$ ($G/R[j]$) is generated based on the White/Blue register 258a (Green/Red register 258b) and predefined gamma values for White and Blue (Green and Red). The multiplexer 260 provides a source driver output 264.

When the bit signal $W/B[k]$ is active, the source driver 192 outputs the source driver output 264 based on the high White gamma correction, the low White gamma correction, the high Blue gamma correction, the low White gamma correction or the low Blue gamma correction.

A demultiplexer 272 is employed in the panel 270 side to demultiplex the driver output 264 from the source driver 232. The demultiplexer 272 is implemented using, for example, thin film transistors, on the panel 270. The outputs from the demultiplexer 272 are couples to two data lines 274 and 276. The demultiplexer 272 is controlled by a control signal associated with the color selection. Based on the output from the demultiplexer 272, one of two data lines 274 and 276 is active. The driver output 264 is demultiplexed

272 on the panel 270 side and goes to different subpixels (i.e., White subpixel, Blue subpixel, Green subpixel, Red subpixel).

In the source driver 232, one PMOS decoder 254 is used for the higher range and one NOMS decoder 254 for the lower range of the voltage. Thus, the area will be reduced by using twice less transistors than a CMOS decoder.

In the panel 270, instead of having four Red subpixel, Green subpixel, Blue subpixel, and White subpixel side by side, they are configured in a quad arrangement where two subpixels for two colors are in one row and the other two colors are in the other row. In this example, one data line 274 carries data for White and Blue subpixels 278a and 278b, and the other data line 276 carries data for Green and Red subpixels 278c and 278d, as shown in FIG. 9. The subpixels are divided into two rows and two columns. Thus the source driver provides data for two subpixels at a time.

Referring to FIG. 10, there is illustrated a system 280 having a source driver 282, a panel 320 having pixels, and external gamma buffer area 290. The system 280 is applied to RGB pixel structure. Multiple gamma corrections for Red, Green and Blue are multiplexed in the external buffer area 290. The external gamma buffer area 290 is located external to the source driver area 282 (e.g., external to the source driver IC). The gamma voltages are generated externally and applied to the source driver 282 through buffers in the external gamma buffer area 290. On the display side 320, a demultiplexing is used to provide data for each color. One of ordinary skill in the art would appreciate that the source driver 282, the external gamma buffer area 290 and the panel 320 may include components not shown in FIG. 10.

A PMOS multiplexer 292 is employed in the external gamma buffer area 290 for high voltage level of gamma corrections, and a NMOS multiplexer 294 is employed in the external gamma buffer area 290 for low voltage level of gamma corrections. The multiplexer 292 multiplexes a high Red gamma correction 296a, a high Green gamma correction 296b and a high Blue gamma correction 296c, and the multiplexer 294 multiplexes a low Red gamma correction 298a, a low Green gamma correction 298b and a low Blue gamma correction 298c. The gamma corrections 296a, 296b and 296c correspond to the gamma corrections 198a, 198b and 198c of FIG. 7, respectively and are located outside the source driver 282. The gamma corrections 298a, 298b and 298c correspond to the gamma corrections 202a, 202b and 202c of FIG. 7, respectively and are located outside the source driver 282. The PMOS and NMOS multiplexers 292 and 294 correspond to the multiplexers 196 and 200 of FIG. 7, respectively and are located outside the source driver 282. The outputs from the PMOS and NMOS multiplexers 292 and 294 are provided to the source driver 282.

The source driver 282 includes a driver output unit 284. The driver output unit 284 includes a DAC that is divided into separate components: a PMOS component 300 ("PMOS DAC" in FIG. 10) and a NMOS component 302 ("NMOS DAC" in FIG. 10). The PMOS and NMOS components 300 and 302 correspond to the PMOS and NMOS components 204 and 206 of FIG. 7, respectively. The PMOS component 300 includes a PMOS decoder and receives the output from the multiplexer 292. The NMOS component 302 includes a NMOS decoder and receives the output from the multiplexer 294. The PMOS and NMOS decoders in the components 300 and 302 operate on an output from a multiplexer 304 for multiplexing a Red register 306a, Green register 306b and Blue register 306c. The registers 306a, 306b and 306c correspond to the registers 210a, 210b and 210c of FIG. 7, respectively.

The driver output unit **284** includes a CMOS multiplexer **308** for multiplexing the outputs from the PMOS and NMOS components **300** and **302**. The multiplexer **308** is operated by a multiplexer **310** for multiplexing bit signals $R[j]$, $G[i]$ and $B[k]$. The multiplexers **308** and **310** correspond to the multiplexers **212** and **214** of FIG. 7, respectively. The multiplexer **308** outputs a single source driver output **316**.

A demultiplexer **322** is employed on the panel **320** side to demultiplex the driver output **264** from the source driver **282**. The demultiplexer **322** corresponds to the demultiplexer **182** of FIG. 5. The demultiplexer **322** is implemented using, for example, thin film transistors, on the panel **320**. The outputs from the demultiplexer **322** are couples to three data lines. The demultiplexer **322** is controlled by a control signal associated with the color selection. Based on the output from the demultiplexer **322**, one of three data lines is active. The driver output **316** is demultiplexed **322** on the panel **320** side and goes to different subpixels (i.e., Red subpixel, Blue subpixel, Green subpixel).

In this example, the PMOS decoder component **300** is used for the higher range and the NMOS decoder component **302** for the lower range of the voltage. Thus, the source area will be reduced by using twice less transistors than that of a CMOS decoder. In addition, the gammas are multiplexed and provided from the outside of the source driver **282** area, thus the number of inputs required for the gamma correction is reduced as well.

For small displays, the gamma correction is internally programmable. The data for gamma correction is stored in internal registers. To reduce the number of gamma registers, DAC resistive ladders and DAC decoders, the gamma registers are multiplexed, as shown in FIG. 11. For programming each color, the corresponding gamma color is assigned to the gamma block. Referring to FIG. 11, there is illustrated a system **330** having a source driver **332** and a panel **360** having pixels. The system is applied to quad RGB pixel structure. Multiple gamma corrections for Red, Green and Blue are multiplexed in the source driver **332**. One of ordinary skill in the art would appreciate that the source driver **332** and the panel **360** may include components not shown in FIG. 11.

The source driver **332** includes a driver output unit **334** having a multiplexer **340** for multiplexing a Red gamma register **342a**, a Green gamma register **342b** and a Blue gamma register **342c**, each for storing the corresponding gamma correction data. The gamma correction is internally programmed (configurable), and the data for the gamma correction is stored in the register. The driver output unit **334** includes a gamma circuit **344** for generating the gamma voltage based on its input signals from the multiplexer **340** (i.e., data from the gamma register **342a**, **342b**, **342c**). The gamma circuit **344** may be, for example, but not limited to, a digital potentiometer or a DAC.

The driver output unit **334** includes a CMOS DAC **346** that has a decoder and receives the output from the gamma correction **344**. The DAC decoder in the DAC **346** operates on an output from a multiplexer **348** for multiplexing a Red register **350a**, a Green register **350b** and a Blue register **350c**. The registers **350a**, **350b** and **350c** correspond to the registers **174a**, **174b** and **174c** of FIG. 5, respectively. The driver output **348** from the DAC decoder **346** is demultiplexed at a demultiplexer **362** in the panel **360** and goes to different subpixels (e.g., Red subpixel, Green subpixel and Blue subpixel). The demultiplexer **362** is implemented using, for example, thin film transistors, on the panel **360**.

For further improving the source driver area, the DAC is divided into NMOS and PMOS decoders as shown in FIG.

12. Referring to FIG. 12, there is illustrated a system **370** having a source driver **372** and a panel **420** having pixels. The system **370** is applied to RGB pixel structure. Multiple gamma corrections for Red, Green and Blue are multiplexed in the source driver **372**. One of ordinary skill in the art would appreciate that the source driver **372** and the panel **420** may include components not shown in FIG. 12.

The source driver **372** includes a driver output unit **374** having a multiplexer **380** for multiplexing a Red gamma register **382a**, a Green gamma register **382b** and a Blue gamma register **382c**. The gamma registers **382a**, **382b** and **382c** correspond to the gamma registers **342a**, **342b** and **342c** of FIG. 11, respectively. The driver output unit **374** includes a high gamma circuit **384** and a low gamma circuit **386**. The high gamma circuit **384** generates a high gamma voltage based on its input signals from the multiplexer **380** (i.e., data from the gamma register **382a**, **382b**, **382c**). The low gamma circuit **386** generates a low gamma voltage based on its input signals from the multiplexer **380** (i.e., data from the gamma register **382a**, **382b**, **382c**). Each of the gamma circuits **384** and **386** may be, for example, but not limited to, a digital potentiometer or a DAC.

The driver output unit **374** includes PMOS and NMOS components **390** and **392**. The PMOS component **390** includes a PMOS decoder and is provided for the high gamma **384**. The NMOS component **392** includes a NMOA decoder and is provided for the low gamma **386**. The PMOS and NMOS components **390** and **392** correspond to the PMOS and NMOS components **204** and **206** of FIG. 7. The PMOS and NMOS decoders in the components **390** and **392** operate on an output from a multiplexer **394** for multiplexing a Red register **396a**, a Green register **396b** and a Blue register **396c**. The registers **396a**, **396b** and **396c** correspond to the registers **174a**, **174b** and **174c** of FIG. 5 (**210a**, **210b** and **210c** of FIG. 7), respectively.

The driver output unit **374** includes a CMOS multiplexer **400** for multiplexing the outputs from the PMOS and NMOS decoders in the components **390** and **392**. The multiplexer **400** is operated by a multiplexer **402** for multiplexing bit signals $R[j]$, $G[i]$ and $B[k]$. The bit signals $R[j]$, $G[i]$ and $B[k]$ correspond to the bit signals $R[j]$, $G[i]$ and $B[k]$ of FIG. 8. The multiplexer **400** outputs a source driver output **404**.

A demultiplexer **422** is employed on the panel **420** side to demultiplex the driver output **404** from the source driver **372**. The demultiplexer **422** corresponds to the demultiplexer **182** of FIG. 5. The demultiplexer **422** is implemented using, for example, thin film transistors, on the panel **420**. The outputs from the demultiplexer **422** are couples to three data lines. The demultiplexer **422** is controlled by a control signal associated with the color selection. Based on the output from the demultiplexer **422**, one of three data lines is active. The driver output **404** is demultiplexed **422** on the panel **420** side and goes to different subpixels (i.e., Red subpixel, Blue subpixel, Green subpixel).

To develop muxing in a source driver, data for each color is multiplexed as shown in FIG. 13. FIG. 13 illustrates a source driver **450** for scanning a panel for a conventional display system. The source driver **450** includes a shift register unit **452** and a latch unit **456**. The shift register unit **452** includes a plurality of shift registers **454a-454d**, and receives a latch signal. The latch unit **456** includes a plurality of latch circuits **458a-458d** that are employed for the shift registers **454a-454b**, respectively. Each latch circuit **458a**, **458b**, **458c**, **458d** latches a digital image signal in response to the latch signal from the corresponding shift register. The outputs from three latch circuits **458a**, **458b** and **458c** are multiplexed by a multiplexer **460** to output R, G, B image

signals. The data for each color is multiplexed **460**. A DAC **462** includes a decoder for decoding the output from the multiplexer **460** to output analog image signals.

To further reduce the source area, the latch unit **456** is replaced with shift registers as shown in FIG. **14**. Referring to FIG. **14**, there is illustrated a source driver **480** for a display system. The source driver **480** includes a first stage shift register unit **482**, a second stage latch and shift unit **486**, and a DAC unit. The multiplexer **460** of FIG. **13** is not implemented in the source driver **480** side. The shift register unit **482** includes a plurality of shift registers, and each receives a latch signal. The latch and shift unit **486** includes a plurality of latch and shift registers that are employed for the shift registers in the shift register unit **482**, respectively. In FIG. **14**, four shift registers **484a-484d** are shown as an example of the components of the shift register unit **482**. In FIG. **14**, four latch and shift registers **488a-488d** are shown as an example of the components of the latch and shift unit **486**. In FIG. **14**, one DAC **490** is shown as an element of the DAC unit. The DAC **490** has a decoder. The DAC **490** is coupled to the latch and shift register **488c**, which decodes its input and outputs a source driver output **492**.

It will be appreciated by one of ordinary skill in the art that the number of the shift registers and the number of the latch and shift registers are not limited to four and may vary. It will be appreciated by one of ordinary skill in the art that the source driver **480** may include components not illustrated in FIG. **14**. It will be appreciated by one of ordinary skill in the art that the DAC unit of the source driver **480** may include more than one DAC. In one example, the DAC unit includes a plurality of DACs connected in M intervals.

Each latch and shift register in the second stage latch and shift unit **486** can copy its input signal and keep it intact till the next activation signal. The input signal to the latch and shift register may come from the corresponding first stage shift register or the previous latch and shift register in the chain. As a result, the latch and shift register can store the data for a row from the first stage shift register or it can shift its own data to the next units. For example, the latch and shift register **488a** latches a digital image signal in response to an activation signal from the corresponding shift register **484a**. The latched signal is shifted to the next latch and shift register **488b**.

After the input signal for a row is stored in the shift register unit **482**, the second stage latch unit **486** is activated and copies the signals from the shift register unit **482**. After that, the second stage latch unit **486** shifts the data one by one to the DACs connected in M intervals connect to the latch unit where M defines the muxing order.

After the first color data is programmed, the latch data is shifted by the number of required bits so that the second data is stored in the latch **488c** connected to the DAC **490**. This operation is executed for other colors as well until all the colors are programmed. This implementation results in a simpler routing and smaller die area. It will be appreciated by one of ordinary skill in the art that a panel side may have a demultiplexer for demultiplexing the source driver **480** output associated with the M multiplexing operation. It will be appreciated by one of ordinary skill in the art that the source driver **480** is applicable to monochrome displays.

Referring to FIG. **15**, there is illustrated a source driver **500** for a display system. To develop DAC decoders, high voltage fabrication process is used, which results in large die area. Instead of a having one gamma curve that covers the entire output voltage range (e.g. 0 to 15), the source driver **500** uses a plurality of smaller offset gamma curve segments

(sections) at lower voltage range, which are extracted from different part of the complete gamma curve.

The source driver **500** includes a gamma block **502** for changing the color (gray scale) mapping for a display, a resistive ladder **504** for generating reference voltages, and an overlapping multiplexer block **506** for the offset gamma curve sections.

The overlapping multiplexer block **506** includes a plurality of multiplexers, each for multiplexing reference voltages for different colors. In FIG. **15**, three multiplexers **508a**, **508b** and **508c** are shown as an example of components of the overlapping multiplexer block **506**. The adjacent multiplexer covers different range of the output voltage, having the beginning and the end of the range. However, the end of one range in one multiplexer and the beginning of the other range in the adjacent multiplexer overlap each other. The overlapping provides flexibility in achieving different gamma curve. The same inputs are being used for both multiplexers.

The source driver **500** includes a DAC decoder section that is segmented into a plurality of low voltage decoders for the offset gamma curve sections. In FIG. **15**, the three low voltage decoders **510a**, **510b** and **510c** are shown as the elements of the DAC decoder, each operating at low voltage. The two adjacent decoders share a small portion of their dynamic range. A programmable decoder **512** defines the border of each decoder **510a-510c** according to the gamma curves. This allows for having different gamma curves for different applications.

In FIG. **16A**, an example of a main gamma curve is illustrated. The main gamma curve **530** of FIG. **16A** has a range from 0 to 10V. In FIG. **16B**, the main gamma curve **530** of FIG. **16A** is segmented into a plurality of offset gamma curve sections **540**, **542** and **544**. Each offset gamma curve section has a shape corresponding to that of the same section of the main gamma curve **530**, and has a voltage range 0 to 5V. The gamma curve section **542** is offset by -5V. The gamma curve section **542** is offset by -10V. Using the offset gamma curve sections, the internal circuits associated with the gamma corrections are offset to lower voltage. The gamma curve section may be internally programmed or input from an external area or device. The display system may include a module for programming/defining offset gamma curve sections. This module may be integrated or operate in conjunction with the programmable decoder **512**.

Referring to FIGS. **15** and **16B**, the multiplexer **508a** is allocated for one offset gamma curve section (e.g., **540** of FIG. **16B**) and the low voltage decoder **510a** uses that offset gamma curve section. The multiplexer **508b** is allocated for another offset gamma curve section (e.g., **542** of FIG. **16B**) and the low voltage decoder **510b** uses that offset gamma curve section. The multiplexer **508c** is allocated for the other offset gamma curve section (e.g., **544** of FIG. **16B**) and the low voltage decoder **510c** uses that offset gamma curve section. The low voltage decoders **510a**, **510b** and **510c** are programmable.

The source driver **500** includes an output buffer **516**. The output buffer **516** outputs a source driver output **520** based on the output from the decoder and the offset voltage.

Based on the pixel circuit data, one offset gamma curve section with its corresponding decoder is being selected. Then the data is passed to the output buffer **516**. In order to create the required voltage, the created voltage is being shifted up at the output buffer **516**. If a voltage is selected

from the second gamma curve section 542 of FIG. 16B, it will be offset by 5 V at the output buffer 516 to cover for the original offset.

Each segment is in its own well so that the body bias can be adjusted accordingly. The decoder can be implemented in low voltage process, leading to smaller die area (over three times saving).

Referring to FIG. 17, there is illustrated an example of a display system 600. The system 600 includes a controller 602, a source driver IC 604, a gate driver IC 606, and a panel 608. The gate driver 606 may include the gate driver 102 of FIGS. 1A-1B or the gate driver 132 of FIGS. 3A-3B. The panel 608 includes a pixel array having a plurality of pixels (or subpixels) 610 and a demultiplexer 612. The demultiplexer 612 may include the demultiplexer 112 of FIGS. 1A-1B or the demultiplexer 142 of FIGS. 3A-3B. The controller 602 controls the source driver 604 and the gate driver 606. The controller 602 also generates control signals 614 to operate the demultiplexer 612, which may correspond to the control signals CTRL(k) of FIGS. 1A or 3A. The demultiplexer 612 is implemented using, for example, thin film transistors, on the panel 608.

Referring to FIG. 18, there is illustrated an example of a display system 630. The system 630 includes a controller 632, a source driver IC 634, a gate driver IC 636, and a panel 638. The source driver 632 may include the source driver 162 of FIG. 5, 192 of FIG. 7, 232 of FIG. 8, 282 of FIG. 10, 332 of FIG. 11 or 372 of FIG. 12. The panel 638 includes a pixel array having a plurality of pixels (or subpixels) 610 and a demultiplexer 642. The demultiplexer 642 may include the demultiplexer 182 of FIG. 5, 222 of FIG. 7, 272 of FIG. 8, 322 of FIG. 10, 362 of FIG. 11 or 422 of FIG. 12. The controller 632 controls the source driver 634 and the gate driver 636. The controller 632 also generates control signals 644 to operate the demultiplexer 632. The demultiplexer 642 is implemented using, for example, thin film transistors, on the panel 638. The system 630 may include the external gamma 290 of FIG. 10.

Referring to FIG. 19, there is illustrated an example of a display system 660 having the source driver elements of FIG. 14 or FIG. 15. The system 660 includes a controller 662, a source driver IC 664, a gate driver IC 666, and a panel 668. The panel 668 includes a pixel array having a plurality of pixels (or subpixels) 610. The controller 662 controls the source driver 664 and the gate driver 666. The controller 662 controls, for example, the shift register unit 482 and the latch and shift unit 486 of FIG. 14 or the overlapping multiplexer block 506 and the low voltage decoders 510a-510b of FIG. 15.

In the above example, the gate drivers and the source drivers are described separately. However, one of ordinary skill in the art would appreciate that any of the gate drivers of FIGS. 1A and 3B can be used with the source drivers of FIGS. 6-15.

What is claimed is:

1. A drive system for [an] a light emitting device (LED) display panel having a multiplicity of LED pixels arranged in rows and columns, each of said LED pixels having a drive transistor that includes a gate, a source and a drain and each of said LED pixels further having an LED coupled to said drive transistor, comprising:

a gate driver having at least one address cell providing a single gate driver output for multiple rows of pixels of said LED display panel,

a gate driver multiplexer and a demultiplexer that includes multiple switch blocks coupled to the gate driver and controllably coupling said single gate driver output to said multiple rows of pixels in sequence [so that whenever] such that when a selected one of said multiple rows is connected to said single gate driver output, all the other said multiple rows are disconnected from said [single-gate] single-gate driver output.

2. A display system according to claim 1, wherein the gate driver [output unit] comprises:

at least one multiplexer, the multiplexer for multiplexing driver signals to provide the single gate driver output.

3. A display system according to claim 2, wherein the panel comprises:

a demultiplexer having a plurality of switch blocks for activating the [first lines] multiple rows, each switch block receiving outputs from the at least one multiplexer.

4. A method of driving a display panel having a multiplicity of pixels arranged in rows and columns, each of said pixels having a drive transistor that includes a gate, a source and a drain, said method comprising:

multiplexing driver signals to generate a single gate driver signal;

providing a single gate driver signal for multiple rows of said pixels of said display panel,

demultiplexing the single gate driver signal by controllably coupling said single gate driver signal to selected ones of said multiple rows of pixels in sequence, and controllably disconnecting said gate driver signal from said multiple rows of pixels to which said single gate driver signal is not coupled.

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