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Lee et al.

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(54) **BEAMFORMING WITH PARTIAL CHANNEL KNOWLEDGE**

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(75) Inventors: **Jungwon Lee**, San Diego, CA (US);
Rohit U. Nabar, Santa Clara, CA (US)

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(73) Assignee: **Marvell International Ltd.**, Hamilton (BM)

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Primary Examiner — Thomas H Tarcza

Assistant Examiner — Harry Liu

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(57) **ABSTRACT**

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Related U.S. Application Data

(63) Continuation of application No. 12/192,264, filed on Aug. 15, 2008, now Pat. No. 7,893,871.

A method in a transmitter for selecting steering vectors for simultaneously transmitting a plurality of streams (N_S) between the transmitter and a receiver, where the receiver has N_R receive antennas, where the transmitter knows respective channels associated with M receive antennas of the receiver, and where M is less than N_R , includes constructing a partial channel matrix that describes a multiple input, multiple output (MIMO) channel between the transmitter and the M receive antennas, generating L independent vectors using the partial channel matrix, wherein L is a rank of the partial channel matrix, selecting a respective steering vector for each of the plurality of streams to be transmitted to the receiver, including, if N_S is less than or equal to L , selecting N_S of the L independent vectors as the steering vectors, and, if N_S is greater than L , (i) selecting the L independent vectors as steering vectors to steer L of the plurality of streams; and (ii) selecting $N_S - L$ orthogonal vectors in a null space of the L independent vectors.

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(51) **Int. Cl.**
H01Q 3/00 (2006.01)

(52) **U.S. Cl.** **342/377**

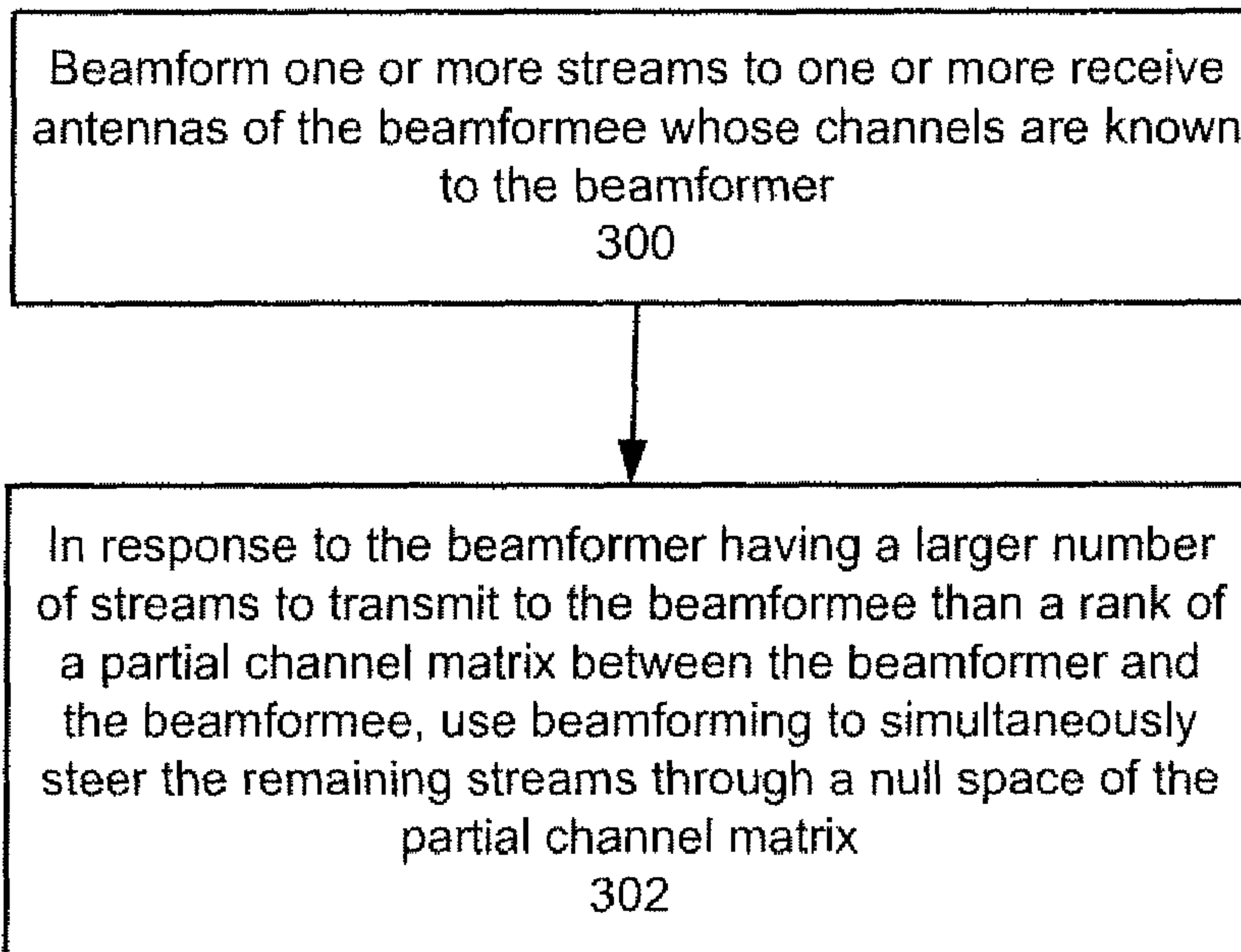
(58) **Field of Classification Search** **342/377**
See application file for complete search history.

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20 Claims, 5 Drawing Sheets



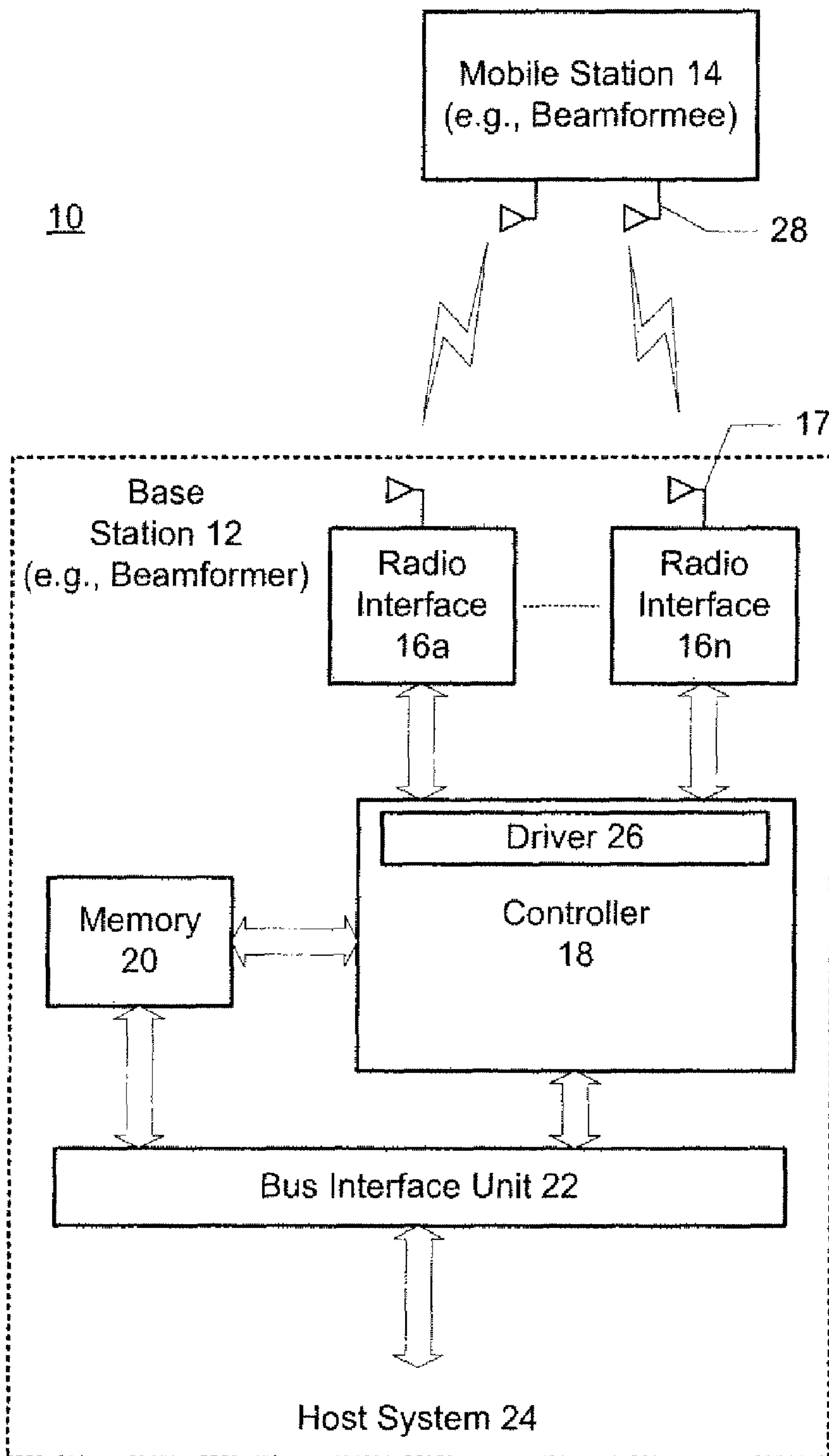


FIG. 1

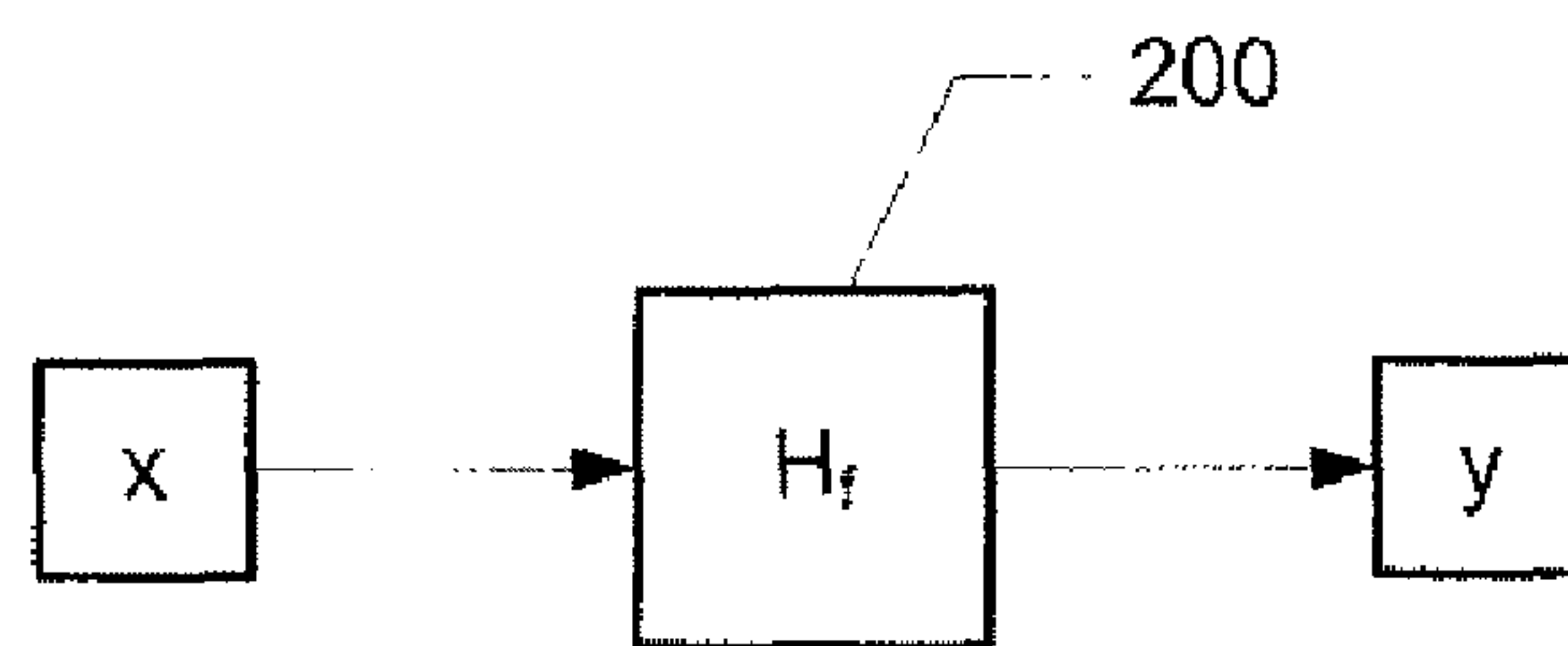


FIG. 2

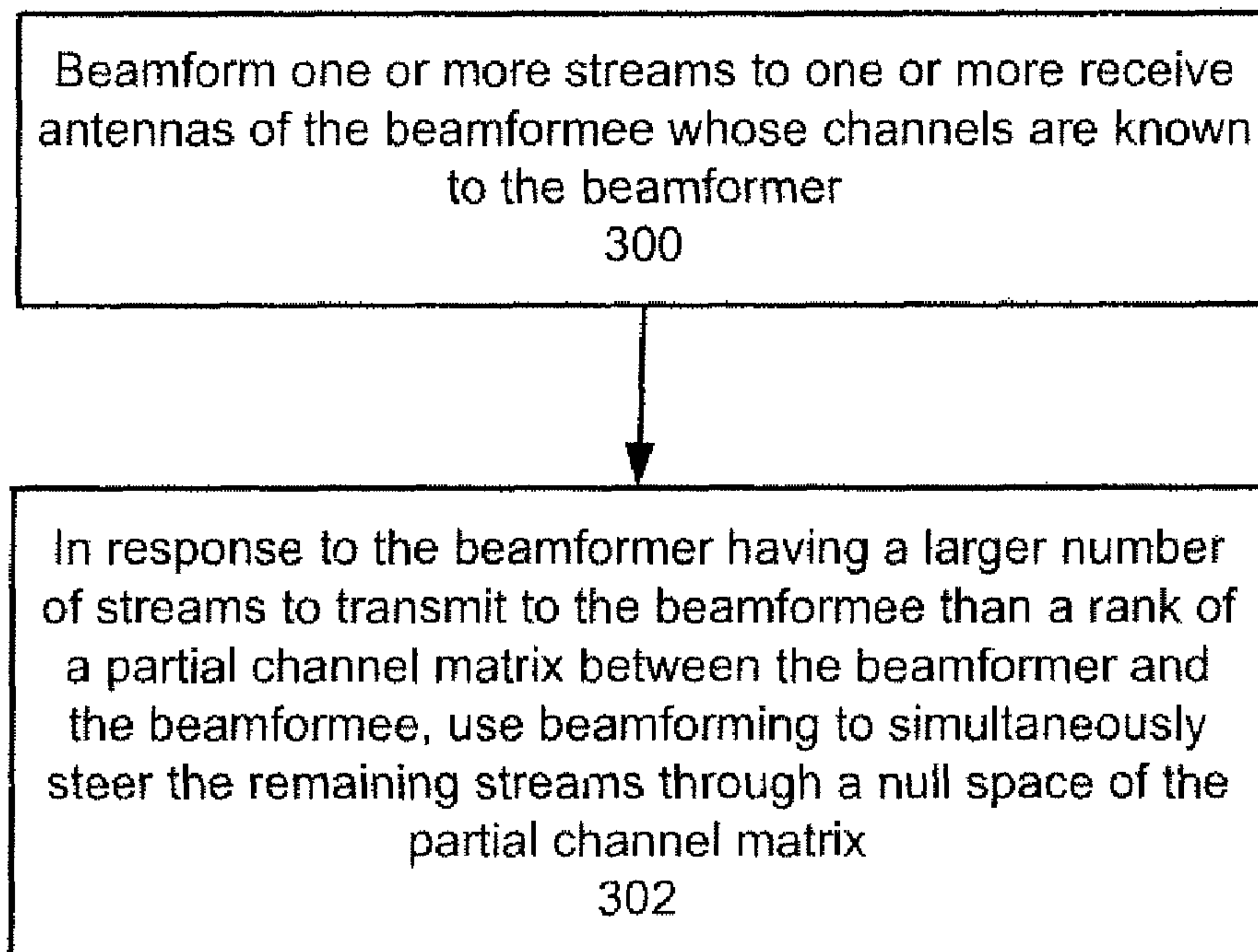


FIG. 3

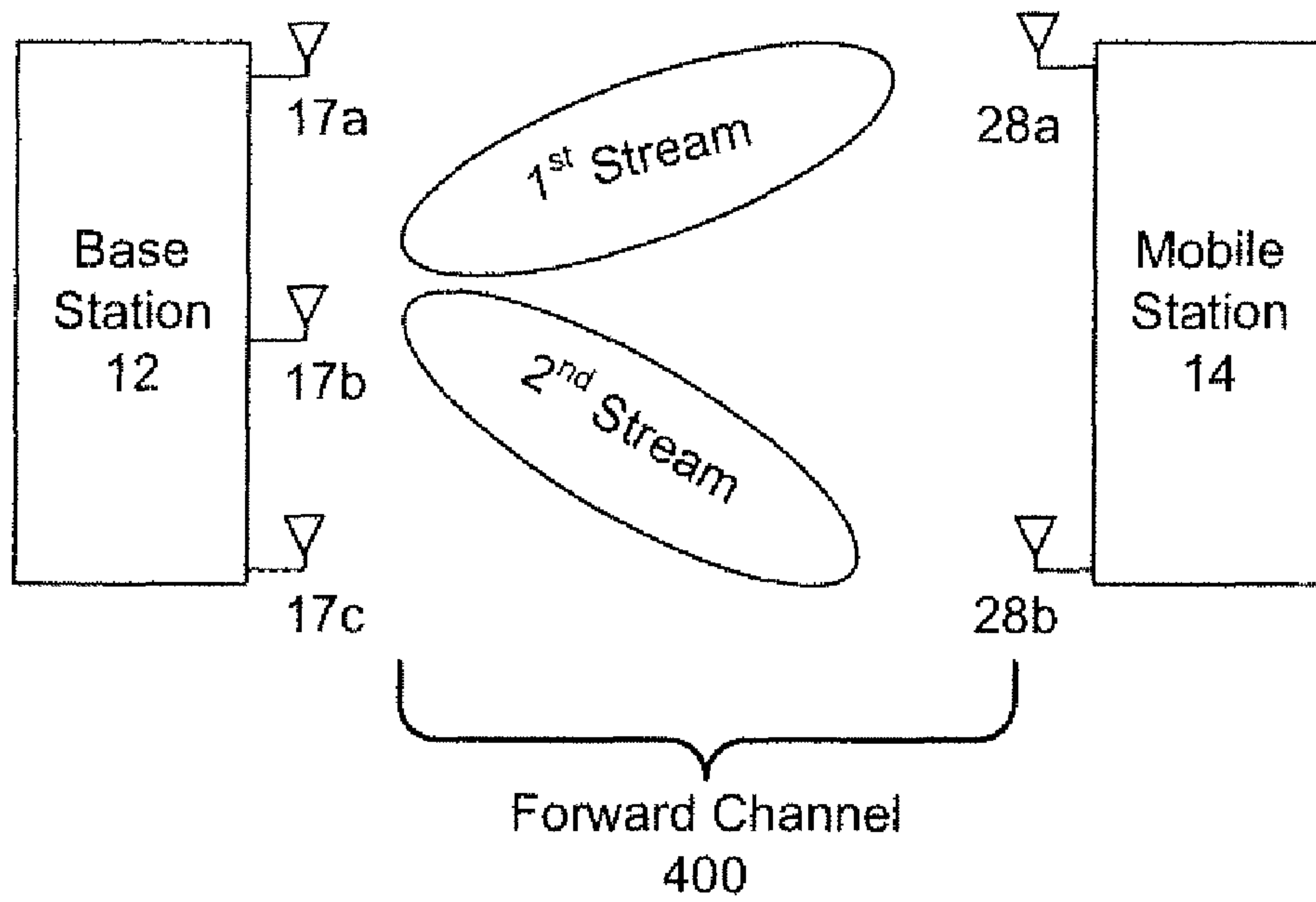


FIG. 4

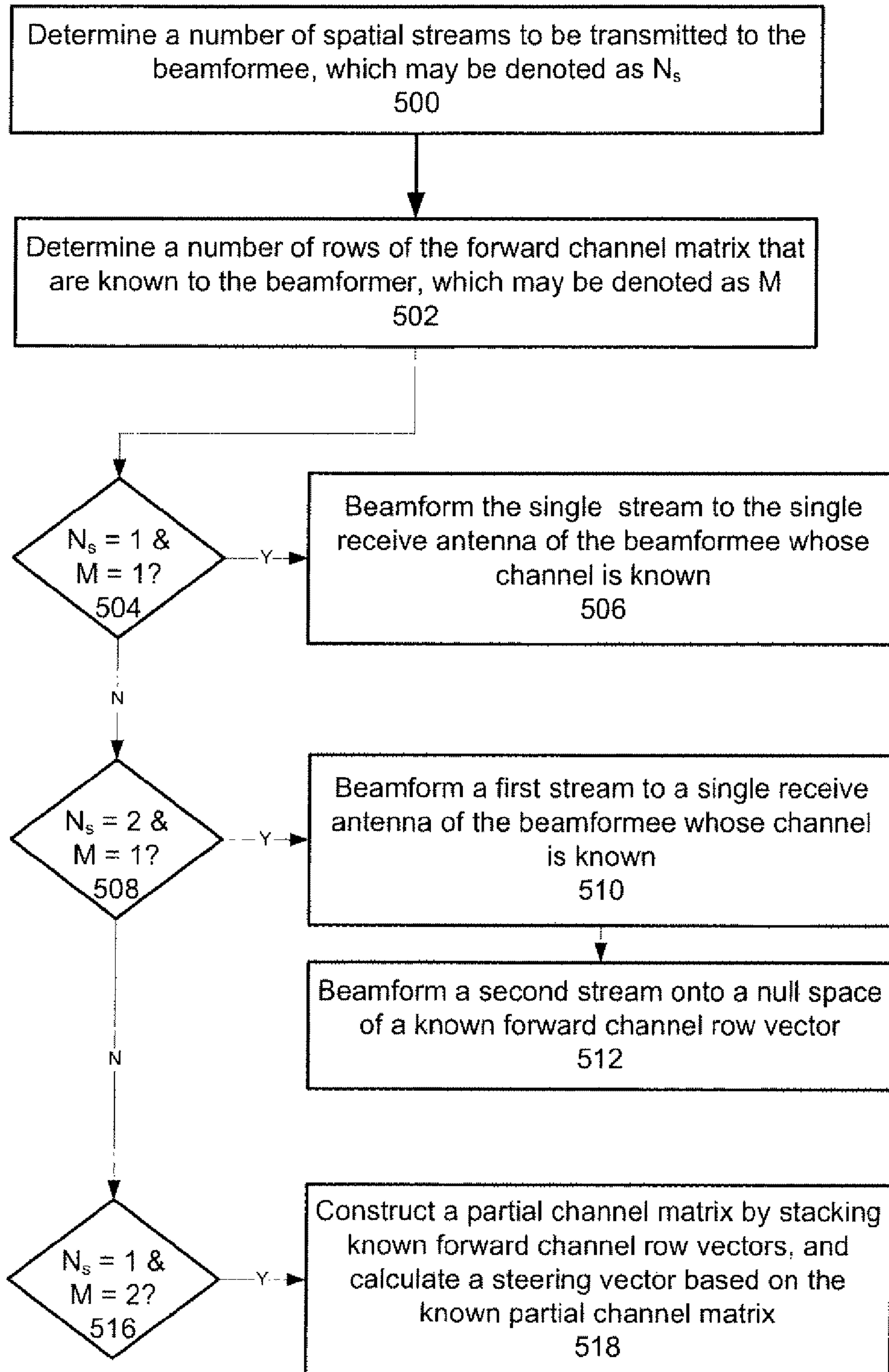


FIG. 5

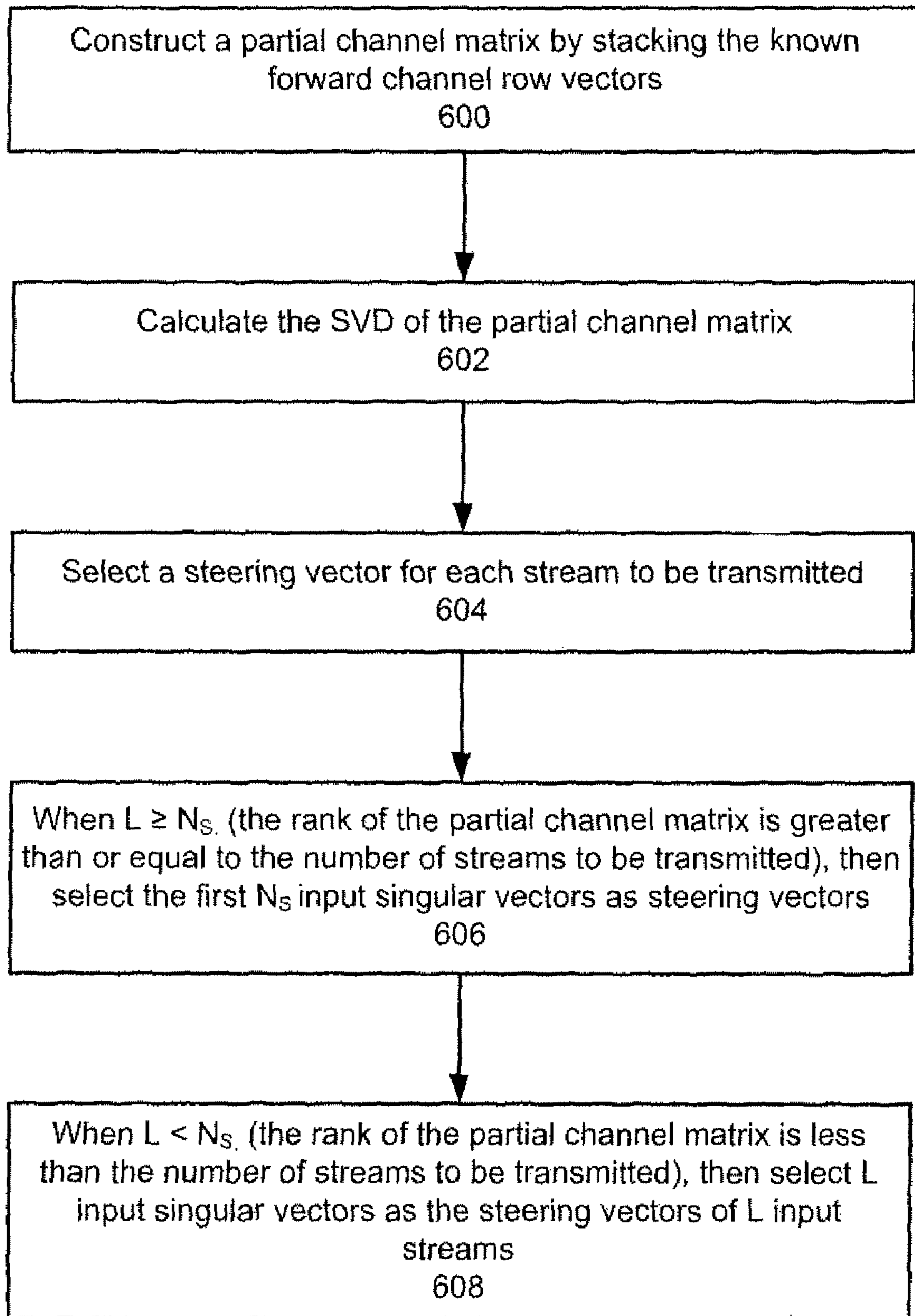


FIG. 6

BEAMFORMING WITH PARTIAL CHANNEL KNOWLEDGE

CROSS-REFERENCE TO RELATED APPLICATIONS

This application is a continuation of U.S. patent application Ser. No. 12/192,264, now U.S. Pat. No. 7,893,871, filed Aug. 15, 2008, that claims the benefit of provisional Patent Application Ser. No. 60/978,942, filed on Oct. 10, 2007. The disclosures of the applications referenced above are incorporated herein by reference.

BACKGROUND

The basis of multiple-input/multiple-output (MIMO) operation is to provide wireless devices with multiple radio interfaces to allow the devices to send data on different channels at the same time in order to achieve greater transmit/receive data rates and with greater reliability. In MIMO systems, a transmitter sends multiple streams of encoded data packets to a receiver by multiple transmit antennas. The streams may be spatially and time encoded and converted into multiple RF signals. The signals are transmitted to the receiver on multiple channels between multiple transmit antennas at the transmitter and multiple receive antennas at the receiver. When the receiver receives the signal vectors from the multiple receive antennas, the receiver decodes the received signal vectors into the original information.

A spatially multiplexed MIMO system that uses multiple transmit and receive antennas not only transmits data between the corresponding transmit and receive antennas but also between adjacent antennas. Thus, data is received in the form of a MIMO channel matrix. Linear algebra techniques such as singular value decomposition (SVD) or matrix inversion may be required to decouple the channel matrix in the spatial domain and recover the transmitted data. The transmitter typically requires some knowledge of the channel state to effectively transmit the streams. One approach for estimating the channel state is to use channel reciprocity, which is generally based on the theory that if a link operates on the same frequency band in both directions, an impulse response of the channel observed between any two antennas may be the same regardless of the direction.

In a MIMO system having a m transmit antennas and n receive antennas, an $(n \times m)$ time varying matrix H is typically denoted as the channel matrix representing the physical propagation channel, where each column represents a channel gain from each transmit antenna of the transmitter to n receive antennas of the receiver.

The channel by which the transmitter transmits the data stream to the receiver is referred to as the forward channel, and may be represented as a channel matrix H_f . The channel from the receiver to the transmitter is referred to as backward channel, and may be represented as channel matrix H_b . Channel reciprocity means that a forward channel and a backward channel are equivalent. Mathematically, channel reciprocity can be defined as:

$$H_b^T = H_f$$

where T is matrix transpose operation.

A forward channel matrix is a transposed version of the backward channel matrix. For example, the forward channel from transmit antenna **1** to receive antenna **2** is the same as the backward channel from receive antenna **2** to transmit antenna **1**.

MIMO performance has been improved through the use of beamforming techniques. Beamforming allows multi-antenna radios to communicate multiple streams of information across a multipath channel such that all streams use the same radio spectrum but do not interfere. Beamforming takes advantage of interference to change the directionality of an antenna array. When transmitting in beamforming, the transmitter is the beamformer and the receiver is the beamformee. The phase and relative amplitude of a signal of beamformer is controlled in order to shape the transmitted beam pattern narrower, such that the energy is transmitted in a particular direction of the beamformee, in contrast to an omnidirectional beam pattern that transmits energy in every direction. When used in a WLAN or cellular environment, beamforming can result in increased received signal power and reduced interference power at the receiver/mobile station.

Several types of beamforming are known, such as beamforming with full channel knowledge and beamforming with no channel knowledge. Beamforming with full channel knowledge can be achieved via two different techniques. One technique for determining full channel knowledge is for beamformer to transmit known training sequences from beamformer transmit antennas to receive antennas of the beamformee to enable the beamformee to estimate channel state information and determine the full channel matrix H_f . Then the beamformee feeds back the forward channel H_f to the beamformer.

Another technique for determining full channel knowledge may be referred to as implicit beamforming. Implicit beamforming calls for the beamformee to “sound the backward channel,” wherein the beamformee sends a known signal to the beamformer. The beamformer then estimates the channel state information for H_b and infers H_f based on channel reciprocity.

Once the beamformer determines full channel knowledge of the forward channel, i.e., the full channel matrix H_f , the beamformer can perform beamforming. In a downlink situation where the beamformer and the beamformee know H_f , they can employ Singular Value Decomposition (SVD) to use input and output singular vectors of H_f to spatially multiplex and demultiplex the transmitted and received vectors to form multiple spatial filters, called beams, with their antenna arrays. In other words, the beams are “steered” in the direction of the receiver. The result of this mux/demux operation is that information symbols in x are communicated through the channel matrix in parallel and without inter-symbol interference. The received symbols are the transmitted symbols scaled by a corresponding singular value, S , but may be corrupted by background noise.

Beamforming may also be performed with no channel knowledge. In beamforming with no channel knowledge, the beamformer randomly generates the steering vector without knowledge of the forward channel to the beamformee. For example, the beamformer may randomly generate a steering vector such that at time 0, a signal is transmitted in a North direction; at time 1, a signal is transmitted in an East direction; at time 2, a signal is transmitted in a South direction; and at time 3, a signal is transmitted in a West direction. Beamformees that receive a strong signal may send a feedback signal reporting that the signal was received and beamformees that received a weak signal may send a feedback signal reporting that the received signal was weak. The beamformer may then decide to which reporting beamformees to allocate the forward channel. Beamforming with no channel knowledge is effective when there are many beamformees associated with a given a station because the beamformer beamforms to an arbitrary direction, and in most cases, the

beamformees will be spread over a whole coverage area in all directions, particularly in cellular systems.

Although beamforming with full channel knowledge and beamforming with no channel knowledge are effective techniques, in some situations, only partial channel knowledge exists. In some MIMO systems, the number of transmit chains in the beamformee can differ from the number of receive chains. For example, in many conventional WiMAX systems, beamformees may have two receive chains, but only one transmit chain, while in WiFi systems, beamformees may have three receive chains and two transmit chains. Typically, the number of transmit chains is smaller than the number of receive chains.

In implicit beamforming based on the beamformee sounding the backward channel, it is assumed that the beamformee sends the known signal to the beamformer on all transmit antennas in order for the beamformer to determine the forward channel. Sometimes, however, the beamformee may not send the known signal to the beamformer using all available transmit antennas. That is, the beamformee may sound only from a subset of available transmit antennas. In this case, only channels from a subset of beamformee transmit antennas may be known to the beamformer. So equivalently, only a partial channel matrix, i.e., a subset of columns of the backward channel H_b , is known. And through channel reciprocity, only a subset of rows of the forward channel H_f will be known to the beamformer. Thus, in this situation, beamforming needs to be done based on the partial channel knowledge, i.e., only a subset of rows of the forward channel H_f .

BRIEF SUMMARY

The exemplary embodiments provide methods and systems for performing beamforming with partial channel knowledge. Aspects of the exemplary embodiment include beamforming one or more streams from a beamformer to one or more receive antennas of a beamformee whose channels are known to the beamformer; and in response to the beamformer having a larger number of streams to transmit to the beamformee than a rank of a partial channel matrix between the beamformer and the beamformee, beamforming is used to steer remaining streams through a null space of the partial channel matrix.

BRIEF DESCRIPTION OF SEVERAL VIEWS OF THE DRAWINGS

FIG. 1 is a block diagram illustrating an exemplary wireless communication system.

FIG. 2 is a diagram graphically representing the modeled MIMO system.

FIG. 3 is flow diagram illustrating a process for beamforming with partial channel knowledge.

FIG. 4 is a block diagram graphically illustrating an example of beamforming with partial channel knowledge.

FIG. 5 is a flow diagram illustrating the process for beamforming with partial channel knowledge in further detail according to an exemplary embodiment.

FIG. 6 is a flow diagram illustrating a process for beamforming streams onto the null space of the known forward channel row vector in further detail.

DETAILED DESCRIPTION OF THE INVENTION

The present invention relates to beamforming with partial channel knowledge. The following description is presented to make and use the invention and is provided in the context of

a patent application and its requirements. Various modifications to the preferred embodiments and the generic principles and features described herein will be readily apparent to those skilled in the art. Thus, the present invention is not intended to be limited to the embodiments shown, but is to be accorded the widest scope consistent with the principles and features described herein.

The preferred embodiment provides methods and systems for beamforming with partial channel knowledge for use in MIMO devices. The exemplary embodiments will be described in terms of MIMO beamforming in the context of an exemplary downlink cellular system comprising a base station and a mobile station. However, the exemplary embodiments are applicable to any MIMO system and other types of wireless communication devices in which beamforming occurs between a beamformer device and a beamformee device. The exemplary embodiments will also be described in the context of particular methods having certain steps. However, the method and system operate effectively for other methods having different and/or additional steps and steps in different orders not inconsistent with the exemplary embodiments.

FIG. 1 is a block diagram illustrating an exemplary wireless communication system. The wireless communication system 10 includes a base station 12 that is in wireless communication with one or more mobile stations 14. The mobile station 14 and the base station 12 communicate through the transmission of signals in the form of streams of encoded data packets over one or more radio frequency (RF) channels. In one exemplary embodiment, both the base station 12 and the mobile station 14 comprise MIMO devices. In one embodiment, the mobile station 14 may comprise a cellular handset, and together with the base station 12, provides cellular services. In another embodiment, the base station 12 may comprise an access point that may be located indoors and the mobile station 14 may be a network device or client station that is used in a desktop/portable computer for communication, for example.

The base station 12 may include two or more independent radio interfaces 16a and 16n (referred herein as radio interface(s) 16) for processing one or more data streams, a controller 18 coupled to the radio interfaces 16, a memory 20 coupled to the controller 18, and a bus interface unit 22 coupled to the controller 18 and to the memory 20 for transmitting data to a host 24 over a host system bus. The mobile station 14 may include a similar architecture.

The radio interfaces 16 are independent from each other because each radio interface 16 has its own antenna 17 and RF chain (not shown). Each RF chain and its corresponding antenna 17 may be capable of transmitting/receiving and processing a data stream. A single frame of data can be broken up and multiplexed across multiple data streams and reassembled at the receiver, which may have the benefits of resolving multipath interference and improving the quality of the received signal. Each of the radio interfaces 16 may be configured as a transceiver, which is capable of operating as both a transmitter and a receiver.

The driver 26 is software or firmware that controls the radio interfaces 16 and can process the data if needed. The driver 26 is executed by the controller 18. The controller 18 may comprise an ASIC, a DSP or other type of processor. The memory 20 stores the incoming and outgoing data packets and any other data needed by the driver 26. The bus interface unit 22 transfers data between the host system 24, and the controller 18 and the memory 20.

In the context of the exemplary cellular system above, a transmission from the base station 12 to the mobile station 14

is known as a downlink transmission, while a transmission from the mobile station **14** to the base station **12** is known as an uplink transmission. During a downlink transmission, the base station **12** transmits information to the mobile station **14** through the transmission of encoded data packets. The data is parallel processed at the base station **12** using a spatial and time encoding function to produce two or more streams of data. Each stream of data is converted into multiple RF signals and transmitted to the mobile station **14** on multiple channels. The transmit streams are transmitted through a channel matrix comprising multiple paths between multiple transmit antennas at the base station **12** and one or more receive antennas at the mobile station **14**. The mobile station **14** receives the multiple RF signals on the multiple channels via receive antennas that recapture the streams of data utilizing a spatial and time decoding function. The mobile station **14** combines and processes/decodes the recaptured streams of data to recover the original data.

In referring to a MIMO system, the concept of beamforming includes a beamformer that transmits a beamformed signal. The receiver of the beamformed signal may be referred to as the beamformee. If beamforming is applied during a downlink transmission in the wireless communication system **10**, then the base station **12** is the beamformer and the mobile station **14** is the beamformee. If beamforming is applied during an uplink transmission, then the mobile station **14** is the beamformer and the base station **12** is the beamformee. The forward channel is the channel over which transmission occurs from the beamformer to the beamformee, and the backward channel is the channel over which transmission occurs from the beamformee to the beamformer.

In one embodiment, the devices in the wireless communication system **10** may have a different number of receive antennas than transmit antennas. In one embodiment, the system **10** may minimally require a 2×2 configuration that has two transmit chains and two receive chains, which allows for two data streams multiplexed across a radio link. Current standards for Worldwide Interoperability for Microwave Access (WiMAX) configuration, as another example, require a minimum of 2 receive antennas and 1 transmit antenna for the mobile station **14**.

Thus, the base station **12** can be described as having a plurality of $N_{B,R}$ receive antennas and $N_{B,T}$ transmit antennas. Similarly, the mobile station **14** can be described as having a plurality of $N_{M,R}$ receive antennas and $N_{M,T}$ transmit antennas. As described above, in beamforming during a downlink, the beamformer (i.e., the base station **12**) beamforms from one or more transmit antennas to one or more receive antennas of the beamformee (i.e., the mobile station **14**). One important factor in beamforming is the number of transmit antennas of the beamformer and the number of receive antennas of the beamformee, which can be designated as, $N_T = N_{B,T}$, and $N_R = N_{M,R}$, respectively.

Beamforming for the MIMO system can be modeled as,

$$y = Hx + n$$

where y represents a $N_R \times 1$ received signal vector, H represents a $N_R \times N_T$ channel matrix, x represents a $N_T \times 1$ transmit signal vector, and n represents a $N_R \times 1$ noise vector.

FIG. **2** is a diagram graphically representing the modeled MIMO system. In the exemplary downlink embodiment where the beamformer transmits to one or more of the receive antennas of the beamformee, the x illustrates the transmit signal vector representing one or more transmit streams transmitted in a channel **200** between the beamformer and the beamformee. The y illustrates the received signal vector received by the beamformee. The channel **200** from the beam-

former to the beamformee is represented by a forward channel matrix H_f that has a dimension of $(N_R \times N_T)$. By transpose, a backward channel H_b from the beamformee to the beamformer has a dimension of $(N_T \times N_R)$. The ranks of the forward and backward channel matrixes are less than or equal to the minimum of the number of transmit antennas N_T and the number of receive antennas N_R .

In implicit beamforming based on the beamformee/mobile station **14** sounding the backward channel, sometimes, the beamformee may sound only from a subset of available transmit antennas. In other words, only a partial channel matrix, i.e., a subset of columns of the backward channel H_b is known. And through channel reciprocity, only a subset of rows of the forward channel H_f will be known to the beamformer/base station **12**. In this situation, the beamformer only has partial channel knowledge, i.e., a partial channel matrix, of the forward channel. The rank of the partial channel matrix will be equal to or less than the number of the beamformee transmit antennas known to the beamformer.

This signal model is applicable to any MIMO system, but for purposes of this disclosure, this MIMO beamforming is described in the context of a downlink cellular system, but may be applied to other types of wireless MIMO systems. This signal model applies to orthogonal frequency division multiplexing (OFDM) system on a per-tone basis. Here just one tone is represented per channel, but if there are multiple subcarriers with OFDM, then there may be parallel channels. As long as the beamforming is done subcarrier by subcarrier, i.e., tone by tone, then this signal model should be sufficient.

FIG. **3** is flow diagram illustrating a process for beamforming with partial channel knowledge in accordance with an exemplary embodiment. In one embodiment, the process may be implemented by the driver **26**. The process may begin by the beamformer beamforming one or more streams to one or more receive antennas of the beamformee whose channels are known to the base station **12** (block **300**). In response to the beamformer having a larger number of streams to transmit to the beamformee than a rank of a partial channel matrix between the beamformer and the beamformee, then the beamformer uses beamforming to simultaneously steer the remaining streams through a null space of the partial channel matrix (block **302**). Each additional stream may be assigned to each orthogonal direction of the null space. This is possible because the dimension of the null space is always larger than or equal to the number of remaining streams.

FIG. **4** is a block diagram graphically illustrating an example of beamforming with partial channel knowledge. In this example, the base station **12** is shown with three transmit antennas **17a**, **17b**, **17e** and the mobile station **14** is shown with two receive antennas **28a**, **28b**. In this example, the base station **12** needs to transmit two streams, and the base station **12** knows the channel for receive antenna **28a**, but not the channel for receive antenna **28b**, and thus only has partial channel knowledge of the forward channel **400**.

Therefore, according to the exemplary embodiment, the base station **12** beamforms a first stream from transmit antenna **17b** to receive antenna **28a** on the channel which is known. The first stream is physically transmitted in the forward channel **400** as a main lobe width (the beam), sidelobes, and null spaces. According to the exemplary embodiment, the base station **12** simultaneously beamforms a second stream from the transmit antenna **17b** through a null space of the partial channel matrix. This is shown graphically as transmitting the second stream through at least a portion of the null spaces in the forward channel **400**, such that the second

stream is not received by the receive antenna **28a** of the mobile station **14** whose channel is known by the base station **12**.

FIG. **5** is a flow diagram illustrating the process for beamforming with partial channel knowledge in further detail according to an exemplary embodiment. The process may begin with the beamformer determining a number of spatial streams N_s to be transmitted to the beamformee (block **500**). In one embodiment, the number of spatial streams N_s to be transmitted to the beamformee is kept less than or equal to a minimum of the number N_R of receive antennas of the beamformee and the number N_T of the transmit antennas of the beamformer,

$$N_s \leq \min\{N_R, N_T\}$$

The beamformer may then determine a number of rows M of the forward channel matrix that are known to the beamformer (block **502**). Conventional techniques, such as sounding the backward channel may be used to determine a number of rows of the forward channel matrix that are known to the beamformer. The number of streams that the beamformer can transmit effectively is limited by the number of receive antennas **28** on the beamformee. By channel reciprocity, the number of transmit antennas **28** on the beamformee indicates the number of rows of the forward channel matrix known to the beamformer.

It is determined if $N_s=1$ and $M=1$ (block **504**). In this case, the number of spatial streams to be transmitted to the beamformee is 1 and the number of rows of the forward channel matrix that are known to the beamformer is 1.

If $N_s=1$ and $M=1$, then the beamformer has channel knowledge of the receive antenna **28a** of the beamformee, and even if the beamformee has more than one receive antenna, the beamformer beamforms the single stream to the single receive antenna **28a** of the beamformee whose channel is known (block **506**).

It is determined if $N_s=2$ and $M=1$ (block **508**), in which case the number of spatial streams to be transmitted to the beamformee is greater than the number of rows of the forward channel matrix that are known to the beamformer. When $N_s=2$ and $M=1$, the beamformer has two streams to transmit, but only has knowledge of one channel to one receive antenna **28a**.

If $N_s=2$ and $M=1$, then the beamformer beamforms a first stream to a single receive antenna **28a** of the beamformee whose channel is known (block **510**). The beamformer also simultaneously beamforms a second stream onto a null space of a known forward channel row vector (block **512**). If the null space has more than one dimension, then a steering vector for the second stream can be randomized within a subspace of the null space as is done in opportunistic beamforming.

It is determined if $N_s=1$ and $M=2$ (block **516**), in which case the number of spatial streams to be transmitted to the beamformee is less than the number of rows of the forward channel matrix that are known to the beamformer. When $N_s=1$ and $M=2$, the beamformer has only one stream to transmit, but has knowledge of two channels to two receive antennas **28a**, **28b**.

If $N_s=1$ and $M=2$, then a partial channel matrix can be constructed by stacking the known forward channel row vectors, and a steering vector can be calculated based on the known partial channel matrix (block **518**). By constructing the partial channel matrix with the known forward channel row vectors, the beamformer is provided with full channel knowledge of the partial channel matrix. In one embodiment, singular value decomposition (SVD) may be used to calculate the steering vector. If SVD is used to calculate a steering

vector, the steering vector may be chosen as the input singular vector having a larger corresponding singular value. In another embodiment, techniques other than SVD may be employed, such as (generalized) co-phasing, and matrix inversion, for example.

FIG. **6** is a flow diagram illustrating a process for beamforming streams onto the null space of the known forward channel row vector (block **512**) in further detail. In one embodiment, beamforming streams onto the null space of the known forward channel row vectors is accomplished by steering the streams to a null space of the partial channel matrix. The process comprises constructing the partial channel matrix by stacking the known forward channel row vectors (block **600**):

$$H_{f,Partial} = \begin{bmatrix} h_{f,1}^T \\ h_{f,2}^T \\ \vdots \\ h_{f,M}^T \end{bmatrix}$$

where $h_{f,1}^T, h_{f,2}^T, \dots, h_{f,M}^T$ represent the known forward channel row vectors. The number of forward channel row vectors in the partial channel matrix is the rank of the partial channel matrix. As stated above, by constructing the partial channel matrix with the known forward channel row vectors, the beamformer is provided with full channel knowledge of the partial channel matrix.

With full channel knowledge of the partial channel matrix and SVD-based steering, the beamformer calculates the SVD of the partial channel matrix $H_{f,Partial}$ (block **602**). The calculation of SVD on the partial channel matrix $H_{f,Partial}$ uses steering vectors to spatially multiplex the known forward channel row vectors to form the transmitted beams,

$$H_{f,Partial} = U \Sigma V^*$$

where

$$\Sigma = \begin{bmatrix} \sigma_1 & 0 & \dots & 0 \\ 0 & \sigma_2 & \dots & 0 \\ \vdots & \vdots & \ddots & \vdots \\ 0 & 0 & \dots & \sigma_L \end{bmatrix}$$

represents a singular value matrix, and singular values are ordered: $\sigma_1 \geq \sigma_2 \geq \dots \geq \sigma_L > 0$;

$V = [v_1 \dots v_L]$ represents an input singular steering matrix; v_1, \dots, v_L : represent input singular vectors; and L represents a rank of the partial channel matrix.

The beamformer selects a steering vector for each stream to be transmitted (block **604**). When $L \geq N_s$ (the rank of the partial channel matrix is greater than or equal to the number of streams to be transmitted), then the beamformer selects the first N_s input singular vectors v_1, \dots, v_L as steering vectors (block **606**). When $L < N_s$ (the rank of the partial channel matrix is less than the number of streams to be transmitted), then the beamformer selects L input singular vectors as the steering vectors of L input streams (block **608**). The beamformer also selects $N_s - L$ orthogonal vectors in a dimension of the null space $N_T - L$. If $N_T > N_s$, the orthogonal vectors can be varied within the $N_T - L$ dimensional null space over time or frequency. If $N_T = N_s$, the assignment of the orthogonal vectors to each stream can be varied over time or frequency.

A method and system for beamforming with partial channel knowledge has been described. The principles herein may be readily expanded. For example if some information about the other receive antennas is available, that information can be utilized. For example, if channel statistics of other receive antennas is known, the beams for those receive antennas can be matched to the statistics. Also, the desired number of streams can be determined based on the effective channel quality dynamically. Furthermore, transmit power may be allocated between the spatial streams that are beamformed onto the known channel and the spatial streams that are beamformed onto the corresponding null space. More generally, power may be allocated across the spatial streams to meet target error rate criteria.

The present invention has been described in accordance with the embodiments shown, and there could be variations to the embodiments, and any variations would be within the spirit and scope of the present invention. For example, the present invention can be implemented using hardware, software, a computer readable medium containing program instructions, or a combination thereof. Software written according to the present invention is to be either stored in some form of computer-readable medium such as memory or CD or DVD-ROM, or is to be transmitted over a network, and is to be executed by a processor. Accordingly, many modifications may be made without departing from the spirit and scope of the appended claims.

We claim:

1. A method in a transmitter for selecting steering vectors for simultaneously transmitting a plurality of streams (N_S) between the transmitter and a receiver, wherein the receiver has N_R receive antennas, wherein the transmitter knows respective channels associated with M receive antennas of the receiver, and wherein M is less than N_R , the method comprising:

constructing a partial channel matrix that describes a multiple input, multiple output (MIMO) channel between the transmitter and the M receive antennas;

generating L independent vectors using the partial channel matrix, wherein L is a rank of the partial channel matrix; and

selecting a respective steering vector for each of the plurality of streams to be transmitted to the receiver, including:

if N_S is less than or equal to L , selecting N_S of the L independent vectors as the steering vectors;

if N_S is greater than L , (i) selecting the L independent vectors as steering vectors to steer L of the plurality of streams; and (ii) selecting N_S-L orthogonal vectors in a null space of the L independent vectors.

2. The method of claim **1**, wherein:

the transmitter has N_T transmit antennas, and a dimensionality of the null space is N_T-L .

3. The method of claim **2**, further comprising, if N_S is greater than L and N_T is greater than N_S , varying the N_S-L orthogonal vectors over time.

4. The method of claim **2**, further comprising, if N_S is greater than L and N_T is greater than N_S , varying the N_S-L orthogonal vectors over frequency.

5. The method of claim **2**, further comprising, if N_S is greater than L and N_T is equal to N_S , varying the N_S-L orthogonal vectors over time.

6. The method of claim **2**, further comprising, if N_S is greater than L and N_T is equal to N_S , varying the N_S-L orthogonal vectors over frequency.

7. The method of claim **1**, wherein generating L independent vectors using the partial channel matrix includes performing singular value decomposition (SVD) of the partial channel matrix.

8. The method of claim **1**, wherein the number of spatial streams N_S to be transmitted to the receiver is less than or equal to a minimum of N_R and a number N_T of the transmit antennas of the transmitter.

9. A beamformer for use with a beamformee having N_R receive antennas, wherein the beamformer knows respective channels associated with M receive antennas of the beamformee, wherein M is less than N_R , and wherein a partial channel matrix describes a multiple input, multiple output (MIMO) channel between the beamformer and the M receive antennas, the beamformer comprising:

multiple (N_T) beamformer antennas;

respective radio interfaces coupled to the multiple beamformer antennas;

a controller coupled to the respective radio interfaces; and

a driver executed by the controller to select steering vectors for simultaneously transmitting a plurality of streams (N_S) to the beamformee, the driver configured to:

construct a partial channel matrix that describes the MIMO channel;

generate L independent vectors using the partial channel matrix, wherein L is a rank of the partial channel matrix;

select a respective steering vector for each of the plurality of streams to be transmitted to the beamformee, including:

if N_S is less than or equal to L , select N_S of the L independent vectors as the steering vectors;

if N_S is greater than L , (i) select the L independent vectors as steering vectors to steer L of the plurality of streams; and (ii) select N_S-L orthogonal vectors in a null space of the L independent vectors.

10. The beamformer of claim **9**, wherein the driver is further configured to vary the N_S-L orthogonal vectors over at least one time and frequency if N_S is greater than L and N_T is greater than N_S .

11. The beamformer of claim **9**, wherein the driver is further configured to vary the N_S-L orthogonal vectors over at least one time and frequency if N_S is greater than L and N_T is equal to N_S .

12. The beamformer of claim **9**, wherein the driver is further configured to generate the L independent vectors using singular value decomposition (SVD) of the partial channel matrix.

13. A communication system comprising:

a beamformee having N_R receive antennas;

a beamformer, wherein the beamformer knows respective channels associated with M receive antennas of the beamformee, wherein M is less than N_R , and wherein a partial channel matrix describes a multiple input, multiple output (MIMO) channel between the beamformer and the M receive antennas, the beamformer including:

multiple (N_T) beamformer antennas,

respective radio interfaces coupled to the multiple beamformer antennas,

a controller coupled to the respective radio interfaces, and

a driver executed by the controller, the driver configured to

steer one or more of streams toward the M receive antennas of the beamformee using the partial channel matrix, wherein the beamformer has N_S streams to transmit to the beamformee, and

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if N_S is greater than a rank of the partial channel matrix between the beamformer and the beamformee, use the partial channel matrix to steer remaining streams through a null space of the partial channel matrix,

wherein the N_S streams are steered simultaneously.

14. The communication system of claim **13**, wherein the driver of the beamformer is further configured to vary the N_S-L orthogonal vectors over at least one of time and frequency.

15. The communication system of claim **13**, wherein the driver of the beamformer is further configured to vary the N_S-L orthogonal vectors over at least one of time and frequency.

16. The communication system of claim **13**, wherein the driver of the beamformer further stacks known forward channel row vectors corresponding to the M receive antennas of the beamformee to construct the partial channel matrix.

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17. The communication system of claim **13**, wherein the driver assigns each remaining stream to each orthogonal dimension of the null space.

18. The communication system of claim **13**, wherein the null space of the partial channel matrix has a plurality of dimensions; wherein

to steer a remaining stream through the null space of the partial channel matrix, the driver randomizes a steering vector for the remaining stream within the subspace of the null space of the partial channel matrix.

19. The communication system of claim **18**, wherein the dimensionality of the null space is N_T-L .

20. The communication system of claim **13**, wherein the number of spatial streams N_S to be transmitted to the beamformee is less than or equal to a minimum of N_R and a number N_T of the transmit antennas of the beamformer.

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